A microcontroller based pacemaker tester

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The design of a microcontroller based pacemaker tester is presented in this paper. Both invasive and noninvasive (transthoracic) pacemakers can be tested using this device. Two test loads are provided; a 500 ohms load for testing invasive pacemakers and a 50 ohms load for testing non-invasive pacemakers. A two line LCD display is used to indicate the test parameters, test results and the test in progress. The test results can be conveniently transferred to a PC or to a printer through the RS232C serial port provided. The unit can be packaged inside a small package and hence can be used as a handheld tester.

1. Introduction

Electrocardiogram (ECG) is one of the common tools used today for the diagnosis of a large number of diseases. The number of heart patients is growing at an enormous rate each year. The use of pacemakers (external and implanted) to correct any malfunctioning of the heart is a very common practice among cardiac patients. Since a pacemaker is a very crucial device, the importance of its testing and calibration need not be overemphasized. The pacemaker tester described in this paper can be used to test a large variety of external pacemakers used today. Since the range of parameters and complexity of operation varies considerably among different manufacturers, it is a very difficult task to integrate all these requirements into a single tester.

The design of pacemaker testers must be made based on easily available components. The selection of a specific microcontroller plays a major role in the design. Since the pacer widths of earlier pacemakers were below 1MS, analog-to-digital converters (ADC) must be carefully selected to provide the required accuracy. The overall cost of the system must also be a major consideration.

2. Terminology

The different parameters related to pacer pulses are given below.

- *Pacer rate* is the number of pacer pulses delivered per minute (PPM) by the pacemaker.
- *Pacer Width* is the duration of a pacer pulse (MS).
- *Pacer Amplitude* is the current delivered by the pacemaker at a particular output setting.
- *Refractory period* is the time interval during which the pacemaker is insensitive to any external inputs. If a QRS is detected during this period, the pacemaker ignores it. On the otherhand, if a QRS is detected outside the refractory interval, then the pacemaker resets its internal timer and the next pacer pulse is generated after a delay of one period from this QRS wave. There are two refractory periods; pacing and sensing.

- *Pacing Refractory period* is the maximum time interval after the generation of a pacer pulse during which time the presence of a QRS wave is ignored by the pacemaker.
- *Sensing Refractory period* is the maximum time interval after a QRS wave is sensed by the pacemaker during which time the presence of a second QRS wave is ignored.
- *Sensitivity* is the minimum QRS amplitude required to cause the pacemaker to operate in the demand mode. (A QRS wave of this amplitude will be sensed by the pacemaker.)
- *A-V delay* is the interval between an atrial stimulus and the subsequent ventricular stimulus.
- *Noise Immunity* is the maximum amplitude of noise signal that will not affect the correct performance of the pacemaker.

A few pacemaker testers are available in the market today [10-11]. These testers are equipped to measure the above mentioned parameters. But none of them have the capability for automatic measurement of any of the above test parameters. For example, the measurement of pacing refractory period is carried out by manually selecting different time intervals on the tester and continuously monitoring the pacemaker until an appropriate response is received. The last tester setting gives the pacing refractory period. On the otherhand the tester described in this paper measures each test parameter automatically without any manual intervention.

3. System Design

A block diagram of the pacemaker tester is shown in Figure 1. The microcontroller houses the program and controls all peripherals. The external static RAM stores all the data collected during the normal test operation and is also used as temporary storage during floating point calculations. The pacemaker output directly feeds into an input selection logic network which selects the test load using a multiplexer (MUX) based on the manufacturer's specifications. The output of this network feeds into an ADC for converting the analog signals into digital data for subsequent processing. The digital data output of the ADC is saved in the RAM. A digital-to-analog converter (DAC) is used to generate different test waveforms for the pacemaker. An LCD display is used for menu display and output indications. The keypad selects the particular test to be performed. The power supply unit provides regulated $\pm 5V$ dc for the whole system and $\pm 2.5V$ dc to the reference inputs of ADC and DAC. This power is derived from a 9V dc battery using a charge pump and a few voltage regulator chips.

3.1 Microcontroller

The microcontroller selected is a derivative of the well-known 80C31: the Philips 80C52 [1]. This is an 8 bit microcontroller which contains an 8kx8 ROM, a 256x8 RAM, four 8-bit ports (P_0 , P_1 , P_2 , P_3), three 16-bit counters/timers, a six-source, two-priority level



Figure 1. Block Diagram of Pacemaker Tester



Figure 2. Microcontroller, Address Decoder and Control Signals

nested interrupt structure, a serial I/O port, and on-chip oscillator and clock circuits. The main program is written in assembly code and the compiled code is stored in the internal ROM. The program code depends on the different measurements to be carried out.

The Ports P_2 and P_0 are used as address and data busses respectively. Port P_1 is used for controlling the keypad and Port P_3 is used for controlling the analog-to-digital converter (ADC). An 8 bit latch L_2 is used for generating extra control signals for the LCD display, digital-to-analog converter (DAC) and the multiplexer (MUX). A simple RC circuit is used for initializing

the microcontroller on power up (RST) as shown in Figure 2. A good choice is R=10k and C=10uF. A crystal of 11.0592 MHz is a good choice for clock frequency. An 8 bit latch 74C373 (L₁) is used for latching the lower order address byte from P₀ bus. The higher order 8 bits of the address is available directly from P₂ port. A 3X8 address decoder is used to generate the chip select (CS) signals for the different devices in the tester. The CMOS decoder 74C138 is used for this purpose. The 3 enable signals together with the 3 address lines in '138 can be carefully selected to generate the proper decoder outputs. The RXD (P_{3.0}) and TXD (P_{3.1}) leads can be used to transfer data to a PC through an RS232C interface.



Figure 3. Input Selection Logic and Analog-to-digital Converter

3.2 Input Selection Logic

The pacemaker output leads feed directly into the input selection logic. For normal testing these outputs are connected to the "Ventricular input" leads of the tester as shown in Figure 3. The "Atrial input" leads are connected to the "Atrial output" leads on the pacemaker during A-V delay measurement. The input network is responsible for the selection of different loads to be connected to the pacemaker and also for amplifying the pacer pulses before feeding to the input of ADC. Two loads are provided; one a 50 Ω and another 500 Ω . Invasive pacemakers are always tested using the 500 Ω load and non-invasive ones using the 50 Ω load. A slide switch S1 is used for selecting the load. When S1 is open, the load seen at the ventricular input leads is 500 Ω , otherwise 50 Ω (for transthoracic pacemakers). In transthoracic pacemakers the current level could go to as high as 200mA. Hence an input attenuation network is included to scale down the voltage. This attenuation network feeds into a dual channel multiplexer (DG409). The scale factors used are 1:10 and 1:100 at X₀ and X₁ inputs of the MUX. Similar attenuation values are provided on the atrial pacer input leads for X₂ and X₃ respectively.

The MUX output is amplified by a differential amplifier (gain $\cong 10$) before feeding to the analog-to-digital converter.

3.3 Analog-to-digital Converter (ADC)

An 8 bit parallel analog-to-digital converter (ADC1205) is used for converting the pacer signal to digital data. The ADC control leads (WR, RD, INT, STS, and CLK) are fed from the microcontroller. The chip select (CS) line is controlled from the address decoder logic. The ALE output of the microcontroller feeds the clock input (CLK) to ADC. The frequency of this signal is 1/6 of the microcontroller clock frequency. The data output lines are tied to the data bus on the microcontroller.



Figure 4. Waveform Generator using Digital-to-analog Converter

3.4 *Digital-to-analog Converter (DAC)*

A dual channel 8 bit DAC (AD7528) is used to generate the test waveforms Sine (SIN), Square (SQR), and Haversine (SSQ) as shown in Figure 4. Three separate look-up tables are used to store these waveforms. The SQR and SSQ waveforms are used to test the sensitivity of the pacemaker as well as its refractory period. The SIN waveform is used for testing the noise immunity. Of the two channels in the DAC one of them is used to control the amplitude of the generated waveform (Channel A). This allows 256 different amplitude selections for the waveform.

4. Test Generation

A comprehensive set of tests for the pacemaker include the measurement of pacer amplitude, pacer width, pacer rate, refractory periods, sensitivity, A-V delay and noise immunity. In fact, the pacemaker must be tested for its full range of parameter values as specified by the manufacturer. Hence for example, pacer amplitude measurement must be done under varying test loads at various current settings, refractory period and sensitivity measurements using signals of varying shapes, pacer rate at various rate settings and so on. To this end, the individual characteristics and parameter ranges of different pacemakers are first tabulated [2-9]. A comprehensive set of such

requirements is given in Table 1. This table lists the pacemaker parameter ranges for both invasive and non-invasive pacemakers. Furthermore Square and Haversine waveforms are recommended by different manufacturers for testing the refractory periods as well as the demand sensitivity. Based on the parameter ranges given in Table 1, and the test waveform requirements, a comprehensive list of tester parameters with appropriate test ranges are identified. This is given in Table 2.

Test Parameters	Invasive Pacemakers	Non-invasive Pacemakers
Rate (PPM)	30-180	30-180
Current (mA)	0.1-20	0-200
Load (ohms)	300-1200	0-1k
Pacer Width (MS)	0.8-1.8	20-40
Refractory Period (MS)	90-335	200-340
Sensitivity (mV)	1-100	
A-V Delay (MS)	0-300	
50/60 Hz Noise Immunity	100mV at max. sensitivity	

Table 2. Parameter Specifications for the Pacemaker Tester

Test Parameters	Invasive Pacemaker	Non-invasive Pacemaker
Test Load (ohms)	500	50
Amplitude (mA)	0.1-25	1-200
Pacer Rate (PPM)	20-1000	20-1000
Pacer Width (MS)	0.8-80	0.8-80
Refractory Period/	10,25,40,100,200MS	10,25,40,100,200MS
Sensitivity	SQ, SSQ Waves	SQ, SSQ Waves

4.1 Algorithm

The first requirement in generating an algorithm is to detect the pacer waveform in the presence of noise signals. This depends on the signal level of the waveform. As the signal level decreases, detecting the waveform becomes more and more difficult. A three point detection method is used to detect the presence of a pacer wave. First a minimum signal threshold value is selected so that when the input level is below (above) this represents the absence (presence) of a pacer wave. This is fixed based on the minimum pacer amplitude. Once the threshold value is fixed, the following criteria is used for the selection of the three points.

1. The first point is selected to be lower than the specified threshold value.

2. The second point identifies the starting of a waveform. The magnitude at this point must be greater than the selected threshold value.

3. The third point is chosen so that the waveform can be separated from transient noise spikes and also to indicate the presence of a sustained pacer wave. The signal level at this point must be higher than that for point 2.

Once the pacer is on, the next step is to detect the termination of the pacer. This is achieved simply by comparing the input signal level with the selected threshold value. Once the signal level falls below this value the pacer is assumed to have ended. Having detected the occurrence of a pacer the next step is to perform various measurements. These tests are selected from the displayed menu using the keypad.

To measure the pacer rate a 1MS timer is started whenever the tester senses the beginning of a pacer pulse. The timer is stopped at the beginning of the next consecutive pacer pulse. Now a count of the number of 1MS timer interrupts is a good measure of the pacer rate. One of the microcontroller timers is used for this purpose.

Pacer width is measured based on the total number of ADC samples while the pacer is on. Hence pacer width accuracy depends on the sampling rate of ADC.

Pacer amplitude measurement is done by measuring the voltage developed across the precision load resistance with the pacemaker output current flowing through it. This voltage when divided by the precision load resistance gives the amplitude.

A-V delay is measured by counting the number of IMS timer interrupts between an atrial pacer and the subsequent ventricular pacer.

The measurement of paced refractory period uses a few cycles of the pacemaker output. First the tester measures the pacer-to-pacer interval T. Then it puts out a SQR/SSQ wave 40MS wide delayed by an interval 'D', which is more than the pacing refractory period, from the last pacer pulse. The pacemaker senses this wave. The delay time 'D' is then decremented in steps during every cycle until the SQR/SSQ waveform is not sensed by the pacemaker. The maximum value of the interval 'D' for which the pacemaker does not sense this wave is the pacing refractory period.

The sensed refractory period is measured in a similar manner, except that the tester now generates two SQR/SSQ waves instead of one. The first wave is generated at a fixed delay from a pacer pulse which is greater than the paced refractory period. The pacemaker always senses this wave. The second one is generated at a delay 'D' from the first wave. Initial value of 'D' is selected to be greater than the sensed refractory period. Hence the first time on, the pacemaker senses both the waves. On subsequent cycles, the delay 'D' is reduced in steps until the pacemaker is unable to sense the second one. The maximum value of 'D' for which the pacemaker does not sense the second wave is the sensed refractory period.

For sensitivity measurement also different waveforms (Square and Haversine) are selectable with widths varying in steps from 10 to 200MS. Once a waveform with a particular width is selected, the tester first generates this waveform with an amplitude of 100mV irrespective of the sensitivity control setting and pacer rate control setting on the pacemaker as long as the pacing rate is set below 150PPM. This waveform is delayed from the pacer pulse so that it is outside the pacing refractory period. The tester then checks whether or not this wave is sensed by the pacemaker. If it is not sensed, then, a message "exceeded" is displayed which means that the pacemaker needs an amplitude more than 100mV for sensing at that setting. If the wave is sensed, the tester then reduces the amplitude in steps until it reaches the lowest value required for

the pacemaker to sense it. (The algorithm used here converges to the lowest value in the least number of cycles.) This lowest value is the sensitivity.

Noise immunity measurements are done by feeding sine waveforms at 50 & 60 Hz to the pacemaker. These waveforms are generated by the waveform generator of Figure 4. During this test the amplitudes of these waveforms are varied until they exceed the rated noise immunity level (100mV). The pacemaker sensitivity must be set to its maximum value during this test.

4.2 Auto Test Mode

This is a very powerful feature provided wherein the tests are performed in a semi-automatic manner without much user intervention. Test and documentation time can be greatly reduced using this feature. For each parameter being tested, a number of different test values are preprogrammed in the tester. The user has the option of running all these tests with the given default values, or running one or more of them with user selectable values, or skipping one or more of these tests. If the user modifies a value, this modified value can be saved so that it can be used for subsequent tests as long as the unit is not switched Off. (The user can also leave the modified value unsaved so that the original default values will be output during subsequent tests.) The test sequence continues until all the parameters are tested.

Once a test is run on the tester, it stores the measured value for future reference. At the completion of the test sequence, the results can either be sent to a printer or to a PC through the RS232C interface. At this time both the set and measured values can be compared and if there is a mismatch of more than a certain percentage, then a comment to that effect can be printed.

5. Conclusion

A simple low cost pacemaker tester is presented here. The design is based on the Philips 80C52 microcontroller. All test results are based on three elements, the crystal oscillator, voltage reference source and the precision load resistance. Time measurements use a IMS timer, controlled by the crystal oscillator running at 11.0592MHz. Voltage and current measurements use a precision load resistance and the reference voltage source. The reference voltage source (Vref) is set at +2.5V dc.

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