Verification of VHDL Designs Using VAL*

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Abstract

VAL (VHDL Annotation Language) uses a small number of new language constructs to annotate VHDL hardware descriptions. VAL annotations, added to the VHDL entity declaration in the form of formal comments, express intended behavior common to all architectural bodies of the entity. Annotations are expressed as parallel processes that accept streams of input signals and generate constraints on output streams. VAL views signals as streams of values ordered by time. Generalized timing expressions allow the designer to refer to relative points on a stream. No concept of preemptive delayed assignment or inertial delay are needed when referring to different relative points in time on a stream. The VAL abstract state model permits abstract data types to be used in specifying history dependent device behavior. Annotations placed inside a VHDL architectural body define detailed correspondences between the behavior specification and architecture. The result is a simple but expressive language extension of VHDL with possible applications to automatic checking of VHDL simulations, hierarchical design, and automatic verification of hardware designs in VHDL.

1.0 Introduction

The VHSIC Hardware Description Language (VHDL) supports the design, description, and simulation of VHSIC components [8] It provides a base language that can be used to describe hardware ranging from simple logic gates to complex digital systems. As an IEEE standard [11], VHDL will provide an important common base language for design tool development and design documentation.

VHSIC designs will incorporate anywhere from a few hundred to perhaps a million components. Managing this complexity requires a powerful hardware design support environment including a library manager, profiler, simulator, and other design tools. A key problem which such environments must address is verifying the correctness of a design. If current practice continues, the VHDL designer will verify designs using a simulator and manually compare huge volumes of simulator output with an informal design specification. For large and complex designs, this is simply not practical.

VHDL Annotation Language (VAL) [1] provides an annotation facility that allows the VHDL designer to apply simple kinds of annotations during the design process. VAL annotations have several possible applications, each of which may be supported by future environment tools. In this paper, we describe VAL and its application to automatic checking of the correctness of a VHDL design during simulation. Other applications of annotations, such as formal verification and optimization of simulation will be discussed in later papers.

In general, annotation languages express information about various aspects of a program in machine readable form that is not normally part of the program itself [5]. They provide facilities for explaining the intended behavior of the program. They are intended to reduce programming errors by making programs more readable and by providing a great deal of error checking at both compile and run time. Readability is improved by enabling the programmer to express design decisions explicitly. Explanations may also serve as specification and thus precede implementation of the program.

VAL allows information about various aspects of a design that may not normally be part of a VHDL description to be expressed explicitly in a machine processable form. Intended behavior, design decisions, and the correspondence between specification and implementation are expressed in a simple but powerful high level language for annotating hardware behavior. Annotations are included in the VHDL text as formal comments. This allows the annotated description to be processed without modification by the VHDL analyzer. A preprocessor, the VAL Transformer, translates VAL annotations into VHDL source code resulting in a self-checking VHDL description.

In the remainder of this paper, we will first give an overview of design checking using VAL. Then we will describe VAL in more detail, showing how VAL annotations are used to generate constraints on a VHDL simulation. A brief overview of the VAL Transformer demonstrates the feasibility of our design. We conclude with some observations made from our experience with VAL to date, and areas for future work.

2.0 Design Checking With VAL

A designer usually verifies a design using some form of simulation. This task often requires the designer to manually compare the simulation result with an informal design specification. Occasionally, the designer also has a high level behavioral description (written in, for example, C or Ada) whose output can be compared to the output of the simulator. The design is simulated using a set of test vectors, the behavioral model is run on the same test vectors, and the results are compared (Figure 1).



Figure 1: Typical Model of Design Checking

While this process of verification is adequate for simple designs, as designs become more complex it becomes less satisfactory. It is limited in the extent to which it allows the designer to debug a new design because it assumes a "black box" view of the design unit (or entity), in which the entity is accessible only through its ports.

VAL's model of design checking is based on generating constraints on the entity's input, internal state, and output (Figure 2). Input constraints allow the simulator to check if an entity is being used cor-

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Paper 4.4 48

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rectly. For example, if the setup or hold time on a signal is not met, the entity can report an input constraint violation. This helps the designer to spot the source of timing errors as opposed to having to trace the source of the error back from the simulation result. Output constraints behave like the post simulation comparison previously described, with the addition that they may be executed dynamically, during the simulation. Mapping constraints allow an additional level of internal checking beyond the checking of ports. For example, if the behavioral description is a state machine, the states in the behavioral description must be somehow encoded within the structural model. (i.e., distributed over the states of the lower level entities in the architecture.) Mapping constraints allow the designer to explicitly describe the encoding and allow the simulator to automatically check the internal state of the structure during simulation, rather than forcing the designer to deduce an incorrect state transition from the simulation result.



Figure 2: VAL Model of Design Checking

3.0 Language Basics

VAL annotations may appear in all three VHDL design units; entities, architectures, and configurations. *Interface annotations* in the entity declaration express intended behavior common to all architectural bodies of the entity. They define an abstract model of the entity's internal state and use it, in conjunction with the inputs to the entity, to define constraints on values carried by the output ports. In VHDL, ports of mode out cannot be read within the entity and therefore assertions cannot normally be made about them. Unlike VHDL, VAL interface annotations have visibility over the contribution of the entity to each of its out ports. This allows the designer to define the entity's contribution to the output port. Constraints may also be defined on input ports.

Annotations within the VHDL architecture (body annotations) can be used to constrain the values of internal signals and ports of components. In addition, VAL annotations within the architecture have visibility over the abstract state of the entity (defined by the interface annotations) as well as the internal states of each component instantiated in the architecture. Annotations in an architecture relating to state are known as mapping annotations. In effect, mapping annotations describe the way in which the abstract state introduced in the interface annotations is mapped into the states of the architecture's components.

Annotations appearing in a configuration (configuration annotations) allow the user to configure the VAL portion of the simulation. For example, the user may want to to select only some of entities in a large simulation for automatic checking. Also, the state model map, similar to a VHDL port or generic map, can be used to map a state model assumed for a component in an architecture into the actual state model of the actual component. This allows a designer to assume an abstract state model for a component during design and provide a type conversion function to translate the assumed state model of the component to the state model of the actual component.

3.1 Interface Annotations

A VAL interface annotation consists of a list of parallel processes that execute continuously. Unlike typical programming languages which execute a process once when it activates, a VAL process executes continuously while active. In many ways this is similar to actual hardware behavior. For example, an AND gate does not behave by watching its inputs for a change and recomputing a new output when a change occurs. Instead, it continuously performs the AND operation. More formally, continuously means that the operation is performed so often that performing it any more frequently would not produce any observable change in behavior.

The following sections summarize the most important kinds of processes in VAL and then show how the VAL concept of relative time, in conjunction with these processes, models hardware behavior.

3.1.1 Assertions

An assertion process generates constraints on the simulation. Consider the VHDL entity interface for a two input AND gate shown in Figure 3. The identifiers input.a and input.b are input ports and result is the output port. Assertions in the form of VAL processes are added to define the behavior of this circuit. The behavior of the AND gate is specified by a single assert process that makes an assertion about the value carried on the output port. The keywords behavior and end behavior delimit the VAL annotations describing the entity's intended behavior. The assert process continuously checks a constraint (in this case (input.a and input.b) = result). It is similar to the assert statement in VHDL. If the constraint ever evaluates to false, the assert process performs the requested action.

-- Annotated VHDL two input AND gate

- entity TwoInputAND is
- port (input_a, input_b : in bit;
- result
- -- VAL Annotations defining the AND gate's behavior
- --| behavior
- --| assert ((input_a and input_b) = result)
- --| severity FAILURE
- -- | report "Error in TwoInputAND" ;
- --| end behavior;
 - end TwoInput_AND;

Figure 3: Annotated VHDL AND Gate Entity Declaration

: out bit);

VAL provides a family of assertion processes for generating constraints. The assert process is the strictest of these, requiring the constraint to be satisfied at every simulation cycle (i.e. at every delta). In this sense it corresponds directly to the VHDL assert statement. Perfectly correct behaviors will often violate this constraint because a zero delay signal assignment in VHDL occurs after a delay of delta, limiting the usefulness of the VHDL assert statement for checking this kind of behavior. Other VAL assertion processes operate by generating constraints only at certain points during the simulation. For example, the finally assertion process allows the user to specify a constraint that must hold only at the last delta in a simulation time point. The constraint generated by the annotation in Figure 4 will report an error for a single delta whenever a change in input_a or input_b causes a change in result because the VHDL simulation of any architecture for this entity does not effect the change until the next delta. Replacing assert by finally checks only after the assignment has been completed, reflecting more closely the intentions of the designer.

Unlike VHDL, VAL provides the capability to hierarchically nest assertions using a guarded process. The keyword when identifies a guarded process that consists of two lists of processes, corresponding to a then part and an else part, and a boolean guard expression. (See Figure 4.) The else part is optional. The guarded process continuously evaluates a boolean expression, and, if the expression is

```
architecture SIMPLE of TwoBitCounter is
    signal Q1, Q2, Q1bar, Q2bar : bit;
    signal D1, D2 : bit;
    component DFlipFlop
      port(Clk, D : in bit;
           Q, Qbar : out bit;
           Reset : in bit);
--| state model is bit: -- local state model declaration
    end component;
    for all: DFlipFlop use
      entity DFlipFlop (SIMPLE);
-- | valentity:
                      -- use the transformed version of this
-- | valarchitecture; -- component for checking
  begin
    DFL1 : DFlipFlop
      port map(Clk, D1, Q1, Q1bar, Reset);
    DFL2 : DflipFlop
     port map(Clk, D2, Q2, Q2bar, Reset);
    D2 <= (Q1 and Q2bar) or (Qibar and Q2);
    D1 \leq C1k:
    Bit0 <= Q1;
    Bit1 <= 02:
-- mapping annotations relate the state of the counter
-- to the states of the components
-- I select state is
     0 => finally(DFL2.state = '0' and DFL1.state = '0')
--|
--1
        report "Counter state does not match flipflop state"
--1
        severity warning;
--| 1 => finally(DFL2.state = '0' and DFL1.state = '1')
--1
        report "Counter state does not match flipflop state"
--1
        severity warning:
--|
     2 => finally(DFL2.state = '1' and DFL1.state = '0')
       report "Counter state does not match flipflop state"
--1
--1
        severity warning;
-- 3 => finally(DFL2.state = '1' and DFL1.state = '1')
--1
        report "Counter state does not match flipflop state"
--1
        severity warning:
--| end select:
end SIMPLE:
```

Figure 7: Two-bit Counter Architecture

4.0 VAL Transformer

The VAL Transformer runs as a pre-processor on an annotated VHDL description to generate a self-checking VHDL description.

4.1 Structure of the Translation

The translation algorithm is based on the generation of an additional architecture called the **MONITOR** that contains an instantiation of the architecture under test and can check its outputs using the assertions in the entity declaration. The concept is similar to plugged a device into a bed of nails for testing. The monitor body has visibility over all signals traveling between the actual entity body and the other components in the simulation. One advantage of this approach is that VAL assertions can also measure the contribution of the entity body's outputs to their environment. A difficult problem in the translation is providing visibility over the component's state needed for the mapping annotations. This is solved by creating an additional port in the entity declaration and passing the state on the port.

The design units involved in the translation are shown in Figure 8. Assume an entity \mathbf{A} exists containing VAL annotations. Three design units are generated; two entity declarations and an architecture. The

Paper 4.4 52 architecture (named MONITOR) contains the VHDL translation of the VAL annotations that appeared in the entity declaration. This includes the annotations which maintain the entity's state model. The ports of architecture MONITOR are the same as for entity A with the addition of an out port of the same type as the entity's state model. This out port is used to provide visibility over the state of components of type A to any annotations within any architecture that instantiates a component of type A. The generated entity A_OUTSTATE describes the interface for MONITOR.

Architecture MONITOR contains a component SOCKET having the same ports as entity A with the addition of an in port of the same type as the entity's state model. A translated version of the original architecture body T of A is plugged into this socket. Because the entity's state is passed into the SOCKET through a port, it is visible to annotations within the architectural body. The translated version of T, T_EXPANDED, contains a translation of the VAL annotations appearing in the architecture into VHDL. Its entity interface is described by A_INSTATE.



Figure 8: Relationship Between Design Units

4.2 Translation of Major Language Constructs

Given the environment for handling scoping and visibility among design units described above, a translation algorithm can be given for each of the language constructs in VAL. The complete details of the VAL to VHDL translation can be found in [7]. The translation mechanism for the **finally** assertion and time qualified assertions are presented here to give the general flavor of the translation process.

4.2.1 Translation of Finally Assertion

The general form of the finally assertion is: finally <test_expression>

- [report <message_expression>]
- [severity <severity_expression>] ;

Within the MONITOR architecture a process is created for each finally assertion. (See Figure 9.) This process is sensitive to all of

```
-- DFlipFlop entity specification
entity DFlipFlop is
  generic (SETUP, HOLD, DELAY : Pos_Integer);
 port (Clk : in bit; -- Clock input
        D
            : in bit; -- Data input
        0
            : out bit; -- Output
        Qbar : out bit);
-- | state model is bit; -- A single bit of memory
--| behavior
-- Assertions about generics
--1
     assert (DELAY >= HOLD)
       report "Error in generic constant" ;
---
-- State maintenance
     when (Clk'Changed('0') then
--1
--1
        when (D'Stable during [-SETUP, HOLD])
--!
         then D -> State[DELAY] ;
--1
          else report "Data not stable" ;
--1
        end when;
--1
     end when:
-- Check outputs
     assert ((State = Q) and (not State = Qbar))
---
       report "Simulator error - D latch" ;
---
--1
     end behavior;
```

end DFlipFlop;

Figure 5: Annotated D Flip-Flop Entity Declaration

Clk'Changed('0') becomes true, the guarded process checking the setup and hold time of the data becomes active. Note that the expression during [-SETUP, HOLD] checks the interval SETUP time units in the past and HOLD time units in the future. If the data remains stable over this interval, the interval state of the D flip-flop is modified after a time DELAY. The assertion processes constrains the ports of the VHDL body to match the state bit, and its negation, at all times.

The constraint DELAY >= HOLD is worth exploring further. Conceptually, this implies that the output can never take on a new value before that new value is latched into the internal state. If DELAY < HOLD were true, then the output could change after DELAY time units, but the hold constraint might not yet be met, in which case the output value should never have changed. In other words, if DELAY < HOLD then the behavior is non-causal. This is more obvious if the VAL description in Figure 5 is rewritten such that the reference point is the point at which the state is assigned a new value. The relevant lines become,

when D'stable during[-SETUP-DELAY,HOLD-DELAY] then

D[-DELAY] -> state[0];

end when;

If HOLD-DELAY > 0, then the assignment to the new value of state depends on an event that hasn't happened yet - the stability of the input during the hold time.

3.2 Body and Mapping Annotations

Any of the VAL processes, with the exception of drive, can appear in the entity body. Body annotations specify implementation details and allow more detailed consistency checking between the interface annotations (the entity's functional description) and the VHDL architecture (implementation). Body annotations have visibility over all VHDL signals and ports normally visible at the point at which the annotation appears, the entity's state model, and the state models of all entities instantiated as components.

The description of the two-bit modulo four counter in Figures 6

and 7 together show how mapping annotations may be used to check the internal state of an entity. The **reset** signal sets the state of the counter. Whenever a transition from '1' to '0' on the clock (Clk) occurs, the counter counts up one. Bit0 represents the least significant bit of the counter and Bit1 the MSB. The VAL state model is an integer and assert processes generate constraints on the output ports based on the VAL state.

entity TwoBitCounter is
port (Clk : in bit;
reset : in bit;
Bit0, Bit1 : out bit);
state model is integer;
behavior
when reset then
state <- 0;
i elsewhen Clk'changed('0') then
state <- (state + 1) mod 4;
! end when;
select state is
0 => finally(Bit0 = '0' and Bit1 = '0')
report "Counter - Dutput error";
severity warning;
1 => finally(Bit0 = '1' and Bit1 = '0')
report "Counter - Output error";
: severity warning;
2 => finally(Bit0 = '0' and Bit1 = '1')
report "Counter ~ Output error";
severity warning;
3 => finally(Bit0 = '1' and Bit1 = '1')
report "Counter - Output error";
severity warning;
end select;
end behavior;
end TwoBitCounter;

Figure 6: Two-bit Counter Entity Declaration

The architecture SIMPLE of the counter contains two D-type flipflops. Each flip-flop is similar to the ones described previously with the exception of a reset signal and the omission of timing information (to keep the examples short enough to fit in this paper). Each flipflop has a state model consisting of a single bit. The states of the flip-flops (DFL1.state and DFL2.state) are related to the state of the counter (state) by mapping annotations.

3.3 Configuration Annotations

Configuration annotations serve two purposes. First, they provide a local state model mapping declaration to map the local state model defined in a component declaration to the actual state model defined by the component's interface annotations. The state model mapping declaration indicates the function to use in mapping between the state model of the actual entity and the state model of the component instance. It appears within a configuration specification at the same point as other binding indications.

Second, they provide configuration information so that VAL generated architectures may be automatically substituted for original component architectures for checking. The user may not want to use a VAL annotated entity in place of the original VHDL entity for all components in a simulation, particularly if the component is a library unit for which no annotated description exists. The valentity construct allows the user to select the components of an architecture to be monitored. The VAL Transformer will only generate code to monitor components marked with valentity. The next section on the VAL Transformer explains how components are monitored.

```
architecture SIMPLE of TwoBitCounter is
    signal Q1, Q2, Q1bar, Q2bar : bit;
    signal D1, D2 : bit;
    component DFlipFlop
      port(Clk, D : in bit:
           Q, Qbar : out bit;
           Reset : in bit):
--| state model is bit; -- local state model declaration
    end component;
    for all: DFlipFlop use
      entity DFlipFlop (SIMPLE);
--1
     valentity;
                       -- use the transformed version of this
-- | valarchitecture: -- component for checking
  begin
   DFL1 : DFlipFlop
     port map(Clk, D1, Q1, Q1bar, Reset);
    DFL2 : DflipFlop
     port map(Clk, D2, Q2, Q2bar, Reset);
    D2 <= (Q1 and Q2bar) or (Q1bar and Q2);
    D1 \leq C1k:
   Bit0 <= Q1;
   Bit1 <= Q2;
-- mapping annotations relate the state of the counter
-- to the states of the components
-- | select state is
    0 => finally(DFL2.state = '0' and DFL1.state = '0')
--1
       report "Counter state does not match flipflop state"
--1
        severity warning;
--1
--t
     1 => finally(DFL2.state = '0' and DFL1.state = '1')
--1
        report "Counter state does not match flipflop state"
--1
        severity warning;
--| 2 => finally(DFL2.state = '1' and DFL1.state = '0')
        report "Counter state does not match flipflop state"
--1
--1
        severity warning;
--| 3 => finally(DFL2.state = '1' and DFL1.state = '1')
--1
       report "Counter state does not match flipflop state"
        severity warning;
--1
--| end select:
```

end SIMPLE:

Figure 7: Two-bit Counter Architecture

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4.2.1 Translation of Finally Assertion

The general form of the finally assertion is:

finally <test_expression>

[report <message_expression>]

[severity <severity_expression>] ;

Within the MONITOR architecture a process is created for each finally assertion. (See Figure 9.) This process is sensitive to all of the signals in the <test_expression> and , in addition, to a wakeup signal VAL_GBE. When the process is activated due to a change in one of the signals in <test_expression>, it remembers the new value of <test_expression> and sets the signal VAL_GBE to wake itself up at the beginning of the next time point to check the remembered value of <test_expression>. The remembered value of <test_expression> will be the value set at the end of all the deltas in the previous simulation point.

process (<test_expression sensitivity list>,VAL_GBE)
variable OLDB : BOOLEAN := TRUE;
variable IB : BOOLEAN := TRUE;
begin
if not VAL_GBE'QUIET then
 assert OLDB
 report <message_expression>
 severity <severity_expression>;
end if;
IB := <test_expression>;
if (IE /= OLDB) then
 OLDB := IB;
 VAL_GBE <= not VAL_GBE after 1fs;
end if;
end process;</pre>

Figure 9: Translation of Finally

4.2.2 Translation of Time Qualified Expressions

In general, a time qualified boolean expression will have the following form:

expr during [T1,T2]

This is translated into VHDL by creating a virtual signal (GBExpr) driven by the boolean expression and then checking the stability of the signal over the requested time interval. The translation looks like:

signal GBExpr : boolean;

 $GBExpr \le expr$;

GBExpr'STABLE(T2 - T1)

Recall that $T2 \ge T1$ in the time qualified boolean expression, and therefore the argument of the attribute 'STABLE must be positive or zero. Nested time qualified expressions generate successive applications of the 'STABLE attribute.

5.0 Experience, Status, and Future Work

The VAL Transformer is currently under development. A prototype transformer for a subset of VAL is currently running with VHDL 7.2. We are currently implementing a VHDL 1076 version of the Transformer. Very preliminary experiments show that annotations in general may slow down the simulation by 20% to 70%, depending of the extent of their use. VAL provides a mechanism (the configuration annotations) for selecting the components that are monitored. This allows the user to select the level of checking necessary for a given application.

VAL has been used in the design and debugging of several benchmarks of moderate size. These include the traffic light controller specified in [6] and described in VHDL in [9], the two-bit counter from which earlier examples have been taken, the ALU in [10], and a simple 16-bit CPU. In all cases the VAL annotation has provided a clean and simple specification of the intended behavior. Perhaps more importantly, the design checking provided by VAL significantly increased our confidence in the correctness of the design. In one instance (the two-bit counter described earlier), an outright design bug missed by the designer in reviewing the VHDL simulation output was flagged and quickly located when the same simulation was automatically checked using VAL. Mapping annotations were particularly useful in isolating the cause of the error. The reason for this is that they allow the subcomponent(s) related to an error to be immediately identified, since an error is detected as soon as an assertion is violated, not just at the outputs of a component.

Currently we are focusing on gaining more experience with annotating larger benchmarks. Language extensions such as additional abstraction mechanisms may be necessary for large and complex entities. Additional kinds of annotations, such as package, type and subtype constraints akin to those in [5] might also be useful. While the current mapping annotations have so far proved adequate, their coarse granularity doesn't provide the detailed level of constraint checking that might be needed. As an aid in debugging, A means of enabling and disabling more detailed assertions would be useful. Finally, VAL's semantics were kept simple to allow the potential application of formal verification methods [2,3]. Formal verification would provide a degree of verification beyond or perhaps in addition to the current model of simulation time constraint checking.

We view VAL as a trend in hardware design languages, and not as a finished project. The next development will probably be constructs for expressing design hierarchy. These are clearly required, even to develop our current VAL checker into a design debugger for use with VHDL simulators. Hierarchy constructs are quite clearly needed to pursue more ambitious applications of design languages such as mathematical verification of designs and (semi-automatic or interactive) synthesis.

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