PCI384

Four-Channel T1/E1/J1
Telecom Adapter

Hardware Manual

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PERFORMANCE TECHNOLOGIES

Document Revision History

Part Number	Date	Explanation of changes
126P0398.10	02/07/02	Initial Release
126P0398.11	04/26/02	Updates to Front Section
126P0398.20	11/02/09	Increased application Flash memory from 16 MB to 32 MB, see pages 25 and 118. Revised QuadFALC from Version 2.1 to Version 3.1, see pages 24 and 73. Moved certifications information and agency approvals to Appendix A, “Agency Approvals” on page 127. Renamed manual to match product name. Reformatted manual.

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Symbols and Conventions in this Manual

An overbar indicates the signal is an Active Low signal.

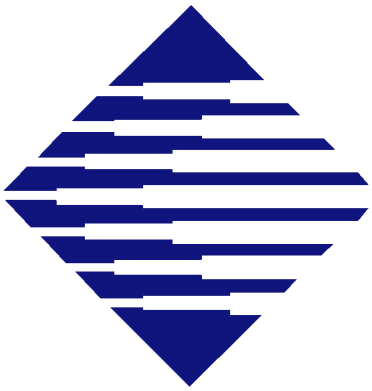
Example: $\overline{\text{ENUM}}$

The following symbols appear in this document.

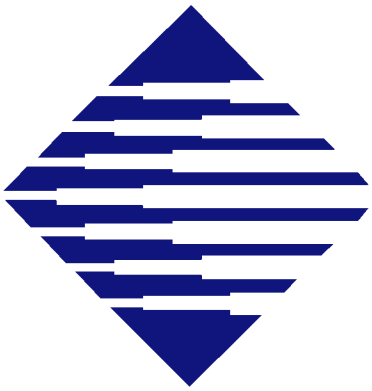
Caution:
There is risk of personal injury or equipment damage. Follow the instructions.

Warning:

Hazardous voltages are present. To reduce the risk of electrical shock and danger to personal health, follow the instructions.



Contents



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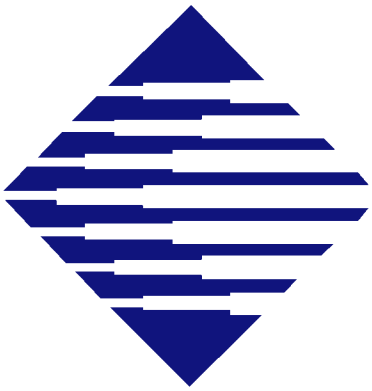
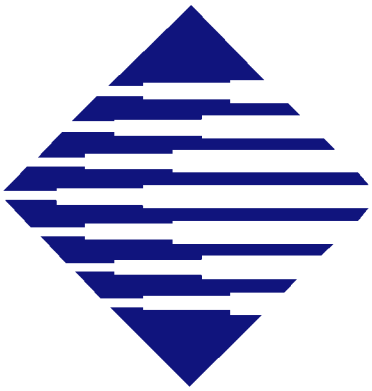


Figures



Chapter

1

About This Guide

Overview

This manual describes the operation and use of the PCI384 Four-Channel T1/E1/J1 Telecom Adapter (referred to as the PCI384 in this manual). The following outline describes the focus of each chapter.

In these chapters you will find installation and configuration information, plus a functional block description intended for the application developer of this board. Here is a brief description of what you will find in this manual:

Chapter 1, “About This Guide,” this chapter, provides links to all other chapters in this manual, customer support and services, and product warranty information for the PCI384.

Chapter 2, “Introduction,” on page 21, introduces users to the PCI384 Four-Channel T1/E1/J1 Telecom Adapter and includes a product description, a list of product specifications, and links to related documents.

Chapter 3, “Installation,” on page 27 describes how to handle, configure the jumpers and install the PCI384.

Chapter 4, “Functional Description,” on page 33 provides an overview of the PCI384 and includes information such as module features, functional block diagram, and a brief description of each block.

Chapter 5, “SDRAM,” on page 49 discusses SDRAM on the PCI384.

Chapter 6, “TDM Clocking,” on page 59 discusses TDM clock distribution, which is centralized about the Zarlink MT90866 H.100 TDM switch on the PCI384.

Chapter 7, “H.100 Bus and Digital Switch,” on page 65 discusses the Zarlink MT90866 switch found on the mezzanine I/O board on the PCI384.

Chapter 8, “T1/E1/J1 Quad Framer and LIU,” on page 73 discusses the T1/E1/J1 quad framer with integral line interface unit found on the mezzanine I/O board on the PCI384.

Chapter 9, “General Purpose Registers,” on page 83 describes the programmable General Purpose Control and Status Registers used to configure and monitor PCI384 functionality.

Chapter 10, “Pinouts,” on page 103 present the pin assignments for the PCI384.

Chapter 11, “Memory,” on page 115 discusses the various memory components and settings on the PCI384.

Appendix A, “Agency Approvals” on page 127 presents agency approval and certification information.

The most current documentation to support any additional components that you purchased from Performance Technologies is available at www.pt.com under the product you are inquiring about.

Text Conventions

This manual uses the following conventions:

Convention	Used For
`Monospace font`	Monospace font represents sample code.
Bold font	Bold font represents: • Paths • File names • UNIX commands • User input
Italic font	Italic font represents: • Notes that supply useful advice • Supplemental information • Referenced documents
$\overline{\text{Overline}}$	An overline is used to represent an Active Low signal.
Regular font	Regular font is used for the ENTER and TAB keys and the SPACEBAR on your keyboard.

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Our offices are open Monday to Friday, 8:00 a.m. to 8:00 p.m. (Eastern Standard Time).

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Phone	+1 (585) 256-0248 (Monday to Friday, 8 a.m. to 8 p.m. Eastern Standard Time)	+1 (585) 256-0248 (Monday to Friday, 8 a.m. to 8 p.m. Eastern Standard Time)

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number. If Performance Technologies determines that the products are not defective, Buyer shall pay Performance Technologies all costs of handling and transportation. This warranty shall not apply to any products Performance Technologies determines to have been subject to testing for other than specified electrical characteristics or to operating and/or environmental conditions in excess of the maximum values established in applicable specifications, or have been subject to mishandling, misuse, static discharge, neglect, improper testing, repair, alteration, parts removal, damage, assembly or processing that alters the physical or electrical properties. This warranty excludes all cost of shipping, customs clearance and related charges outside the United States. Products containing batteries are warranted as above excluding batteries.

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Chapter

2

Introduction

Overview

This chapter introduces users to the PCI384 Four-Channel T1/E1/J1 Telecom Adapter, shown in Figure 2-1, “PCI384 Quad T1/E1/J1 Telecom Adapter,” on page 23.

Key topics in this chapter include:

Scope of this Document

This manual is not intended to be a stand-alone document. See “Additional Documentation,” on page 25 for additional manuals and sources of information. It is necessary to have a complete understanding of all the features and function of the hardware to adequately develop software for this product.

This manual provides information necessary to understand the basic operation and features of the board and answer questions raised by developers as to whether the PCI384 will complement their architecture.

PCI384 Models and Accessories

Table 2-1, “PCI384 Part Numbers,” lists the PTI part numbers for the models and accessories available for the PCI384.

Table 2-1: PCI384 Part Numbers

Item	PTI Part Number
PCI384 Versions	
Quad T1/E1/J1 Communications Adapter	PT-PCI384-11935
Quad T1/E1/J1 Communications Adapter with NexusWare Core and RTU	PT-PCI384N-11928
Quad T1/E1/J1 Communications Adapter with HDLC and RTU	PT-PCI384H-11929
Quad T1/E1/J1 Communications Adapter with X.25 and RTU	PT-PCI384X-11930
Quad T1/E1/J1 Communications Adapter with Frame Relay and RTU	PT-PCI384F-11931
Cable Options	
E1/75 Ω cable (RJ48C to BNC) Ungrounded Shield	PT-ACC384-11938
E1/75 Ω cable (RJ48C to BNC) Grounded Shield	PT-ACC384-11939
RS232 Debug Cable (Console)	PT-ACC324-11977
Two LVDS Clock Input Cables	PT-ACC384-11940
Two-position H.100 Cable	PT-ACC384-11937

Product Summary

The PCI384 is a four-channel T1/E1/J1 telecom adapter for PCI-based systems. See Figure 2-1, “PCI384 Quad T1/E1/J1 Telecom Adapter,” on page 23. It provides OEMs, end users, and integrators with a high performance, fully channelized platform for use in both datacom and telecom applications. Utilizing the Motorola MPC8260 PowerQUICC® II processor running at 200 MHz, the PCI384 can run protocols such as Frame Relay, ISDN, or any protocol using HDLC in Internet/WAN environments. In telecom applications, the PCI384 will support MTP-2 on all four links. The PCI384 supports the PCI Local Bus, Revision 2.2, and the ECTF H.100 bus, revision 1.0.

The heart of the PCI384 is a Motorola MPC8260 PowerQUICC II communications microprocessor running at 200 MHz. The PowerQUICC II includes an embedded communications processor module (CPM) that utilizes a 32-bit RISC controller residing on a separate local bus. The CPM has an instruction set optimized for communications to off-load the PowerPC core CPU from the task of handling lower- level communications and DMA activity. This architecture provides an ideal platform for running Performance Technologies, Inc. Uconx suite of protocol software.

Figure 2-1: PCI384 Quad T1/E1/J1 Telecom Adapter

PCI384 Features

Some of the features of the PCI384 include:

- EC603e PowerPC 200 MHz microprocessor core
- PPC 66 MHz, 60x Bus
- 128 MB SDRAM
- Integrated CSU/DSU
- Supports ATM /TDM inverse muxing (MPC8260)
- PCI-full high, short-card, 64-bit size, 106.68mm x 167.64 mm with I/O mezzanine card that conforms to single-slot PCI requirements
- Zarlink MT90866 H.100 WAN Access switch, ECTF H.100 controller with 1024 timeslots access between local and H.100 Buses
- Local Stratum 4 oscillator reference clock for H.100 and Trunk lines
- Clock distribution logic for TDM devices
- Internal TDM Highway for Framer/ CPU/ switch connections
- Tri-state driver for CT Netref
- Channelizes up to four T1 or E1s (192/256 Timeslots via H.100 controller)
- Dual External LVDS Reverence Clock Input on the front panel micro-DB9 connector
- Four T1/E1/J1 RJ48C Trunk connections for 75/100/120/110 Ohm Connections (75 Ohm BNC requires external cable adapter)
- Integral Line protection and galvanic isolation for trunk interface
- Single QuadFALC® transceiver/framer, Version 3.1
- Software programmable trunk termination options
- Tundra CA91L8260-100CEZ PowerSpan PCI Interface
- Universal signaling 64-bit PCI interface
- 256 KB Communication Buffer Memory on the MPC8260 local bus
- RS232 console port on front panel micro-DB9 connection
- Serial Management Controller (SMC) general-purpose communication UART
- Remote Break detect on the console port provides remote reset operation
- 32 MB of Application Flash
- 512 KB Bytes Boot Flash
- One Status Yellow/Green LED on mezzanine board
- Power-On reset and Control Logic
- CPC388 Compatible Register set
- On-board regulators for CPU and PowerSpan core voltages
- COP Header for Debugger Support
- In-system programmable port header for PAL logic
- Analyzer header positions

PCI384 Specifications

Table 2-2: PCI384 Specifications

Component	Specification
Interface	Four RJ48C interfaces that are independently software programmable on receive and transmit termination. Operating modes supported are: • T1/100 Ohm • E1/75 Ohm • E1/120 Ohm • J1/110 Ohm 1 Micro DB 9 interface supporting: • RS232 Debug (optional cable) • 2 LVDS Clock Inputs (optional cable)
Processor	Motorola MPC8260 PowerQUICC II (MPC603e core) 64 bit data and 32 bit address bus
Framing Standards	D-4, ESF, DS-1, PRI; AMI/B8ZS Line encoding
Memory	128 MB dedicated DRAM 32 MB Application flash PROM 256 KB COM memory (high speed local memory)
Specification Compliance	PCI Revision 2.2 ECTF H.100 Compliant ANSI T1. 102-1993 IEEE 802.3
Physical Interface	T1/E1/J1: Four RJ-48C Connectors Monitors: One Micro DB 9
Protocol Support	SS7, Channel7 MTP-2, HDLC, Frame Relay, LAPD, X.25, PPP Optional NexusWare, Linux-based development environment
Agency Certifications	FCC Class A, CE, UL 1950 (pending), NEBS Level 3 friendly
MTBF	Calculated MTBF: 175,424 POH Calculated FITS: 5700
Power	11 watts maximum (2.2A @ +5.0 V) 9.5 watts idle (1.9A @ +5.0 V)
Dimensions	PCI short form
Temperature	Operating: 0° to 50° C (32° to 122° F) Non-operating: –20° to 80° C (–4° to 176° F)

Additional Documentation

This section provides links to datasheets, standards, and specifications for the technology designed into the PCI384.

Processor

For more information about the Motorola MPC8260 PowerQUICC II Integrated Communications Processor, refer to the Freescale (formerly Motorola) Web site, where you can view the Datasheet and User's Manual for this device:

http://www.freescale.com/webapp/sps/site/prod_summary.jsp?code=MPC8260

PowerSpan PowerPC-to-PCI Bus Switch

For more information about the Tundra CA91L8260-100CEZ PowerSpan PowerPC-to-PCI Bus switch, refer to the Tundra Web site, where you can view the Datasheet and User's Manual for this device:

http://www.tundra.com/products/product-archive/powerspan

QuadFALC Framer and Line Interface Component

For more information about the Infineon Quad T1/E1/J1 Framer and Line Interface Component for Long and Short-Haul Applications, QuadFALC, PEB 22554, refer to the Infineon Web site, where you can view the Datasheet and User's Manual for this device:

http://www.infineon.com/cms/en/product/index.html

H.100 WAN Access Switch

For more information about the Zarlink 4,096 x 2,432 Channels WAN Access Digital switch with H.110 Interface Compliant TDM Backplane, refer to the Zarlink Web site, where you can view the Datasheet and User's Manual for this device:

http://www.zarlink.com/zarlink/hs/82_MT90866.htm

Application Flash Memory

For more information about the Intel StrataFlash memory, refer to the Intel Web site, where you can view the Datasheet and User's Manual for this device:

http://developer.intel.com/

PICMG Specification

The PCI384 is compliant with the PCI Local Bus Specification Revision 2.2. This specification can be purchased from PICMG (PCI Industrial Computers Manufacturers Group). A short form specification in Adobe Acrobat format (PDF) is also available on PICMG's Web site at:

https://www.picmg.org

Chapter

3

Installation

Overview

This chapter summarizes the information you need to make the PCI384 operational. Please read it before attempting to use the board. Key topics in this chapter include:

Installation Overview

Installation consists of the following basic steps:

1. Configure the PCI384 for your application.
2. Install the PCI384.
3. Bring up your system.

This manual presents the installation procedure for the PCI384 hardware only. For instructions on how to install the optionally purchased software, refer to the NexusWare Core manuals:

- *NexusWare Core Installation and Configuration Guide* (106P0055)
- *NexusWare Core Reference Manual* (106P1359)

Before starting, note the following about the mechanical aspects of the PCI384:

- The PCI384 meets PCI Revision 2.1
- The PCI384 has been tested for mechanical compatibility and installs in most systems without issue

Handling Instructions

Caution:
Static electricity can damage the adapter. *Do not open the static-protective package containing the adapter until you are instructed to do so. Follow these handling instructions.*

- Use the static-protective bag (with the adapter still inside) as a grounding device and touch it to a metal expansion slot cover for at least two seconds. This action reduces any static electricity from the package and from your body.
- Do not touch any exposed printed circuitry.
- Do not place the adapter on the computer cover or a metal table.
- Prevent other people from touching the adapter.
- Limit your movement, because your movement can cause static-electricity buildup.
- Always handle the adapter carefully.
- If you put the adapter down after you have removed it from the package, place the adapter on the static-protective package on a flat surface.
- We recommend you use static-protective wrist straps during the installation. Connect them to the computer chassis after you have removed the cover from the computer.

Configuration Jumpers

The configuration jumpers for the PCI384 are at P3 (right angle header on the base board), accessible between the base and mezzanine boards on the side opposite the PCI card edge fingers, as shown in Figure 3-1, “Configuration Jumpers,” below.

Figure 3-1: Configuration Jumpers

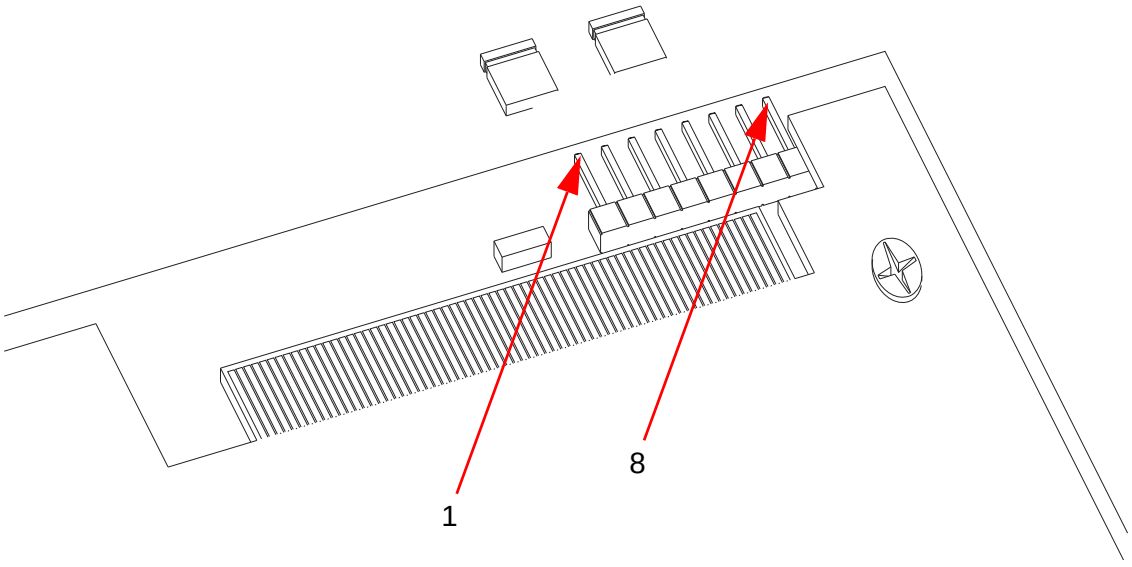


Table 3-1, “Configuration Jumpers,” shows the jumper settings and functions.

Table 3-1: Configuration Jumpers

Jumper Position	Jumper In/Out	Jumper Function
1-2	IN	CPU Reset Confide [1]
2-3	OUT	Power On Reset
4-5	OUT	Non Maskable Interrupt
6-7	OUT	Break Detect Reset (use for Debug only)
7-8	OUT	Autoboot [2]

1. Must be in for normal operation.
2. Do not install this jumper on a PCI384 purchased without the protocol software.

Quick Installation

If you are an experienced user, follow these instructions to install the adapter:

Note: *The Quick Installation is recommended for experienced users only!*

1. Quit all applications. Power down the PC and any attached peripherals. You will probably need a flat blade or Phillips type screwdriver to remove screws securing the computer case. Remove the top cover of the PC.
2. Remove the board from the anti-static bag. Be sure to discharge the static electricity from your body by touching a protective grounding device or the metal chassis of your computer. Always handle the board carefully by its edges and avoid touching any of the components or connectors.
3. Select an available PCI slot and remove the slot filler panel.
4. Slide the board into the PCI connector of the system unit. Make sure the front plate on the board card mounts flush with the chassis panel opening.
5. Install the front plate screw to secure the board into the chassis This also provides a chassis ground connection to the board.
6. Replace the top cover.
7. Install your T1/E1/J1 cable assembly to the board connector.
8. Re-connect any cables from the peripheral devices.

Standard Installation

Follow these instructions to perform a standard installation of the adapter:

Step 1. Unplug the Computer and Remove the Cover

Step 2. Install the Board in an Expansion Slot

If you are installing the board in a floor-standing machine, place the computer on its side so that you can press firmly to install the adapter correctly.

To install the board in an expansion slot, follow these steps:

1. Find an empty expansion slot on the system board.
2. Refer to the section of the computer's reference manual that describes board installation.
3. Remove the expansion slot cover(s). Follow the steps in the computer's reference manual that describe how to remove the expansion slot cover(s).
4. Remove the board from its static-protective package.
5. Follow the instructions in the computer's reference manual for placing the board in an expansion slot.

Note: *Ensure the board fits properly into its expansion slot connector at the bottom of the computer's chassis. Then press down the board vertically until it snaps into place.*

Step 3. Reinstall the Computer Cover and Reconnect the Cables

1. Refer to the section of your computer's reference manual that describes the computer installation and follow those instructions.
2. To reconnect the cables, follow the instructions given in your computer's reference manual.

Step 4. Master/Slave Installation

Two dual-port PCI384 boards may be configured to operate as a Master/Slave pair.

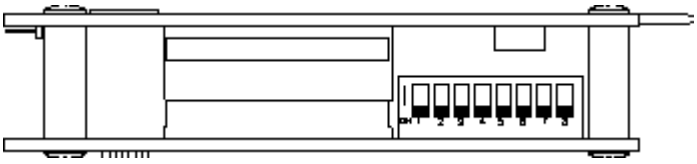
Installation of a master slave pair must be done as follows:

1. Install the Master/Slave pair into adjacent slots in your system following the procedures outlined above.
2. Install Master/Slave PCM Expansion cable to P1 of the Master and P2 of the Slave across the edges of the PCI384 pair. (Consult your Performance Technologies sales representative for PCM Expansion cable ordering information.)
3. The pin assignments for the Master/Slave PCM Expansion cable can be found in "Master/Slave PCM Expansion Connector Pin Assignments," on page 108.

Step 5. Restart and Configure the Computer

This completes the hardware installation. At this point, turn the power back on to the computer system and proceed to any software installation instructions that may have been provided.

Figure 3-2: Stacked PCI384 Boards



There are no hardware functions assigned to DIP switches 1-2. Switches used have assignments for specific software packages. Consult with the software manual for appropriate settings.

Chapter

4

Functional Description

Overview

The PCI384 is a full-featured four-port T1/E1/J1 communications processor and time division access controller with a quad T1/E1/J1 transceiver, ECTF H.100 time division multiplex interface, with PCI and TDM, High Speed I/O.

See Figure 4-1, “PCI384 Block Diagram,” on page 35.

Key topics in this chapter include:

PCI384 Product Description

The PCI384 is composed of a CompactPCI main card and a mezzanine I/O card. The main card contains all of the active components relating to MPC8260 core operations. This includes the MPC8260 processor; SDRAM and SSRAM memory, Boot Flash, Application Flash, the PCI-bus interface and general control and option register logic. The mezzanine I/O card provides the T1/E1 Framers and Line Interface Unit (LIU), ECTF H.100 controller, the receiver/transmitter metallic connectors (RJ48C) and passive components including the telephony line protection.

The on-board T1/E1 framers and LIU receives and transmits framed serialized T1/E1/J1 bit streams, recover timing information, and demultiplex the T1/E1 data directly to the DS-0 level. The result is that 192/256 T1/E1/J1 DS-0 timeslots that may be terminated in different manners. The data may be passed to the PCI bus as terminated data or to the H.100 bus as DS-0s in a TDM stream.

In the receive direction, the LIU and Framers accept T1/E1 or J1 bipolar signals, perform equalization, convert the signals into binary digital data streams, receive frame synchronization, data recovery, frame overhead extraction, and provide data connection paths to the various on board terminating interfaces, via the H.100 controller.

In the transmit direction, the LIU and Framers accept digital data from one of the terminating interfaces, via the H.100 controller. The LIU and Framers perform frame generation overhead insertion, and convert the digital clock and data to the appropriate wave shapes for transmission as bipolar T1/E1 or J1 streams.

The H.100 controller interfaces the T1/E1 LIU and Framers the MPC8260 processor's TDM. It is fully compliant with the ECTF H.100 specification and has no restrictions to the possible types of switching connections that can be made. Each local timeslot can be assigned to any of the possible 4096 H.100 external timeslots or reassigned to any other local timeslot.

The PCI interface provides bus master capability of 32/64 bit data accesses at 33/66 MHz bus rates.

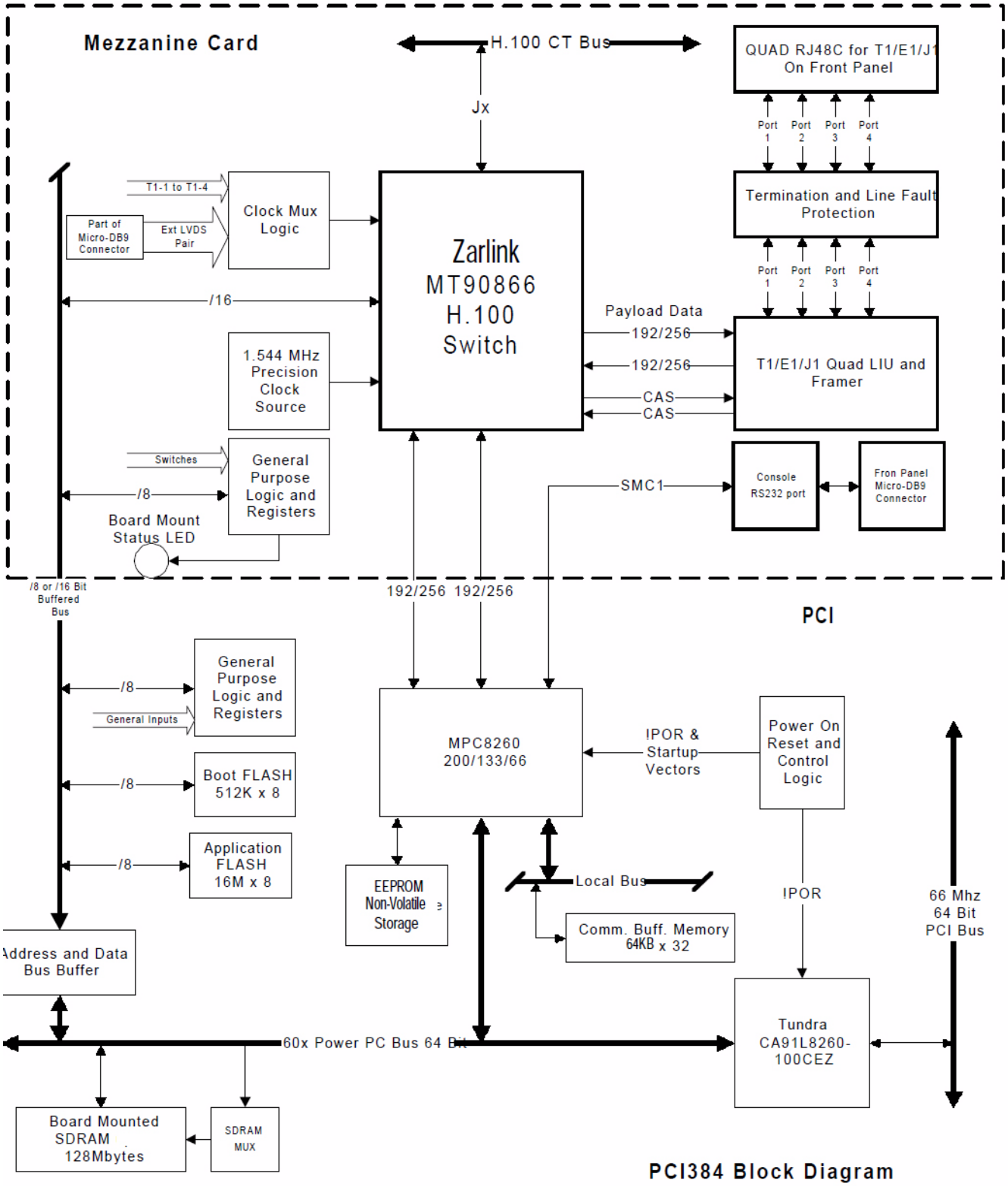
The MPC8260 PowerQUICC II CPM implements channel routing and time division multiplexing (TDM) to its internal communications controllers, via its internal time slot assigner (TSA). Data can be routed to and from the various CPM communications controllers including:

- Two multi-channel communications controllers (MCC)
- Two fast communications controllers (FCC)
- Two serial-management controllers (SMC)

PCI384 Functional Block Diagram

Figure 4-1, “PCI384 Block Diagram,” on page 35 is a functional block diagram of the PCI384. The topics that follow provide overviews of the functional blocks.

Figure 4-1: PCI384 Block Diagram



Reset Logic

The reset logic for the board has many components. There are power-on, hard, and soft resets and there are resets caused by the external PCI bus. Each of these will be discussed in the following sections.

Power-on Reset

A Dallas DS1233A initiates the Power-on reset sequence for the board. The device senses the 3.3 V power and issues a PORESET to the MPC8260 and the Tundra PCI bridge. The DS1233A provides a 350ms reset pulse after the 3.3 V power settles to an *intolerance* condition. It will also signal PORESET if there is a brownout condition. In addition to the CPU and bridge, the signal is provided to all of the FALC JTAG resets, the PSM519 PAL for the PAL reset logic and it has a pin on the P4 Debug for connection to an analyzer.

The device also will de-bounce a push button or jumper on it's output. This feature is supported in the design with jumper. Shorting P3-3 to P3-2 will cause a debounced PORESET to be issued from the part.

Hard Reset

The hard reset signal, PQ_HRESET, is generated by the MPC6820 and the PSM519 PAL Logic. The MPC8260 generates PQ_HRESET in response to a power on reset, a software watchdog reset (if enabled), a bus monitor reset (if enabled) or a checkstop reset (if enabled).

The PSM519 PAL will generate a PQ_HRESET under the following condition. The configuration of a few jumpers that feed into the PSM519 PAL, regulate which signals generate PQ_HRESET.

The P1_RST signal is generated by the PCI bus and is the PCI reset. The signal is buffered and fed to the PAL logic and the PCI side of the Tundra chip as P1_RST_IN. The signal will always drive a PQ_HRESET.

The PQ_HRESET signal is generated when the falling edge of P1_RST_IN is detected and held low for approximately 11, 50 ns. clock cycles. The PQ_HRESET will then be negated and not reasserted until the P1_RST_IN has been negated for at least 50 ns.

Soft Reset

The soft reset for the board is accomplished by asserting the PQ_SRESET signal. The signal is distributed to the MPC8260, PSM517 PAL, PSM519 PAL, the P7 Analyzer Connector and the COP8 JTAG header for the debug port. The signal sources for PQ_SRESET include the MPC8260 and the PSM519 PAL logic.

The MPC8260 will assert this signal in response to any power on reset or hard reset condition. The effect of soft reset on the processor is different than power on reset or hard reset and is outlined in the Motorola MPC8260 User's Manual in the RESET chapter.

Logic inside the PSM519 PAL will also assert the PQ_SRESET signal when the Tundra PowerSpan chip asserts a level 5 interrupt. This mechanism is used to let the system host cause a soft reset to the board under the control of the system software. This feature is only active when the PQ_HRESET signal is inactive and the Reset Configuration jumper P3, is in the Normal Position
P3-1 to P3-2.

Tundra PowerSpan Resets

The Tundra PowerSpan has two sources of Reset. It can both receive and source reset on the PCI bus and the local processor or PB bus. The setup on each source is a follows.

Primary PCI Reset

In the PCI384 implementation, the chip is set to receive a reset on the primary PCI 1 bus and to not be a source. This is accomplished by setting the P1_RST_DIR pin low. The P1_RST signal from the PCI bus will cause logic within the chip to be reset as outlined in the Tundra users manual. The P1_RST also causes a PQ_HRESET to the board as described in the Hard Reset section.

Processor Bus Reset

The Power Span processor Bus reset pin, PB_RST, is configured to be an input. This is accomplished by setting the PB_RST_DIR pin low. The source of the signal is the PSM519 logic. The logic drives the PB_RST pin with a standard driver output and it drives the signal true when PQ_HRESET is true or whenever the Reset Configuration jumper P3, is in the PROM Programming Position P3-2 to P3-3.

Peripheral Resets

Each peripheral chip on the PCI384 has an individual reset line. The peripheral resets are driven by the PSM519 PAL logic. The resets are all set to a reset state by anything that creates a PQ_SRESET on the board. The devices will be held in reset until the appropriate bit in the general-purpose registers is set. Table 4-1, “Peripheral Resets,” describes the individual peripheral reset signals, the power-up reset state and the bit that controls the device in the general-purpose registers.

Table 4-1: Peripheral Resets

Peripheral	Reset Signal Name	PQ_SRESET State	Register Bit Name
H.100 Switch	H100_RST	0	h100_rst_n
QuadFALC 1	QFALC1_RST	0	qfalc1_rst_n

The bit actions are described further in the Chapter 9, “General Purpose Registers,” on page 83.

Central Processing Unit

The CPU that is used on the board is a Motorola MPC8260. It is located on the base-board portion of the assembly.

The MPC8260 is the primary controller on the PCI384. It supplies the PowerPC CPU core, the Communications Processor Module (CPM), and the 60X bus processor bus controller. It has direct connections to, and is the controller for the Synchronous DRAM (SDRAM), the local bus Synchronous Static RAM (SSRAM), and the Tundra PowerSpan PCI Bridge. It also controls all of the on board peripheral chips via a buffered data and address bus.

For more information on the internal portions of the chip, see “Processor,” on page 25 for a link to the Freescale (formerly Motorola) Web site. There you can search for supporting documentation for this device.

MPC8260 Initial Configuration

A great deal of the initialization of the MPC8260 occurs during the hard reset process. When the PQ_HRESET signal is active the CPU selects its CPU PLL settings and several other initial configuration settings. The following sections describe these settings.

Core Clock and PLL

The initial implementation of the board will be with a unit that has a 200 MHz Power PC Core (PPC), a 133 MHz CPM and a 66 MHz external 60x, multi-master bus. The chip is supplied with a 66 MHz primary clock input (CLK1) on the CLKIN input. The clock PLL settings to select the core speed are driven into the part by the MODCLOCK lines from the PSM517 PAL logic while the PQ_HRESET signal is active. The clock settings are as follows:

200 MHz core/133 MHz CPM/66 MHz 60x Bus (Initial Release Product) MODCLOCK[2:0] 111

Table 4-2, “System Clock Control Register Bit Field Definitions,” presents suggested core clock configurations settings for operation with the Performance Technologies default software settings for system clocks, refresh timers, and baud rate generators.

Table 4-2: System Clock Control Register Bit Field Definitions

Bits	Name	Recommended Setting	Description
29	CLPD	0	CPM does not enter low power mode when the core enters low power mode.
30-31	DFBRG	01	Division factor of BRGCLK relative to VCO_OUT (twice the CPM clock). Defines the BRGCLK frequency. The BRGCLK is divided from the CPM clock. 01 set the divisor to 16.

Note: *When the MPC8260 does not have the RSTCONF jumper in the Boot PROM program mode, P3-1 to P3-2, the clock default modes for the PLL are invoked. The MPC8260 will function in this mode, it will be able to use the SDRAM and program the Boot Flash PROM but no other operations, CPM or peripherals are guaranteed or supported.*

Reset Configuration

The reset configuration has two modes. If the RSTCONF is high (removed jumper in the P3-1 to P3-2 Boot Flash programming mode) the MPC8260 will take the default settings and there will be no Reset Configuration cycles on the Bus because the MPC8260 will be set to be a RESET Slave and the PowerSpan will be held in reset.

If the RSTCONT jumper is in the Normal mode (P3-1 to P3-2) the MPC8260 will be setup as the Reset Configuration Master. The PowerSpan device will be the 3rd configuration slave.

The MPC8260 will read its configuration from the Boot Flash device beginning at location 0. The format of the boot words is listed in the MPC8260's user manual. The information read from the PROM is loaded into the Hard Reset Configuration Word Register. The address for this device is located in the memory map in the MPC8260s user manual. The standard configuration supplied with the Boot Flash is:

- No external arbitration
- Internal memory controller selected for the flash
- Core is active
- External Bus Mode is 60X compatible
- Boot port size is 8 bit
- Core initial exception prefix is zero. Exceptions are vectored to Physical address 0xFFFn_nnnn
- Internal space port size is set to 32 bit
- Bus address pins selected for Layer2 cache pin configuration. (BADDR Pins selected)
- No data bus parity, IRQ pins selected for the multifunction pins
- Base address of the Internal Memory space is 0xFF00_0000
- Boot memory space 0xFE00_0000 to 0xFFFF_FFFF
- Address Bus Busy and Data Bus Busy selected on multifunction pins
- MMR Register initial value is no masking on Bus Request lines
- Local Bus Pin Configuration (LBPC), pins function as a local bus
- Address Parity Pin configuration, no address parity, Bank Select Function Selected on Multifunction pins
- CS10 pin configuration is set to be $\overline{\text{BCTL1}}$, buffer control for byte lane 1
- MODCK_H Clock Reset Configuration high order bits set to 111

Internal Initial Configuration

The internal Initial configuration is obtained from the Boot Code running out of the Flash PROM. The initial settings for the System Integration Unit module configuration register (SIUMCR) because its register fields are most closely tied to the hardware configuration.

Table 4-3: SIUMCR Register Settings

Field	Recommended Setting	Description
BBD	0	$\overline{\text{ABB}}$/$\overline{\text{IRQ2}}$ pin is $\overline{\text{ABB}}$, $\overline{\text{DBB}}$/$\overline{\text{IRQ3}}$ pin is $\overline{\text{DBB}}$
ESE	1	External Snooping is enabled, $\overline{\text{GBL}}$/$\overline{\text{IRQ1}}$ pin is $\overline{\text{GBL}}$.
PBSE	0	Parity byte select is disabled, GPL4 is available for UPM use.
CDIS	0	MPC8260 core is enabled.

Table 4-3: SIUMCR Register Settings (Continued)

Field	Recommended Setting	Description
DPPC	00	Pins set to $\overline{\text{IRQx}}$
L2CPC	10	Level 2 cache pins set to BADDR[29-31]
LBPC	00	Local Bus pins function as local bus
APPC	10	MODCLK Pins function as BNKSEL[0-2] and $\overline{\text{IRQ7}}$ out and $\overline{\text{CS11}}$
CS10PC	01	$\overline{\text{CS10}}$ is $\overline{\text{BCTL1}}$
BCYLC	00	$\overline{\text{BCTL0}}$ is used as W/R!, $\overline{\text{BCTL1}}$ is $\overline{\text{OE}}$ control.
MMR	00	No masking on bus request lines.
LPBSE	0	Parity Byte disabled. LGPL4 output of UPM is available for memory control.

Memory Map

The memory map is defined by the PCI384 Address Decode Scheme. There are different levels of Address Decode built into the PCI384. The first level and primary decode is done by the MPC8260s System Interface Unit (SIU). One of the SIUs subsections is the memory controller. The memory controller is responsible for controlling a maximum of twelve memory banks shared by a high performance SDRAM machine, a general-purpose chip-select machine (GPCM), and three user-programmable machines (UPMs). It supports a glueless interface to synchronous DRAM (SDRAM), SRAM, EPROM, Flash EPROM, burstable RAM, regular DRAM devices, extended data output DRAM devices, and other peripherals. This flexible memory controller allows the implementation of memory systems with very specific timing requirements.

- The SDRAM machine provides an interface to synchronous DRAMs, using SDRAM pipelining, bank interleaving, and back-to-back page mode to achieve the highest performance.
- The GPCM provides interfacing for simpler, lower-performance memory resources and memory-mapped devices. The GPCM has inherently lower performance because it does not support bursting. For this reason, GPCM-controlled banks are used primarily for boot loading and access to low-performance memory-mapped peripherals.
- The UPM supports address multiplexing of the external bus, refresh timers, and generation of programmable control signals for row address and column address strobes to allow for a glueless interface to DRAMs, burstable SRAMs, and almost any other kind of peripheral. The refresh timers allow refresh cycles to be initiated. The UPM can be used to generate different timing patterns for the control signals that govern a memory device. These patterns define how the external control signals behave during a read, write, burst-read, or burst- write access request. Refresh timers are also available to periodically generate user-defined refresh cycles.

The primary control of the devices served by the memory controller machines is through the MPC8260s external Chip select lines. The specific memory controller setups will be defined for each type of device in the section of the specification that describes the device.

Table 4-4, “MPC8269 Chip Select Assignments,” represents the primary address decode of the External chip select lines.

Table 4-4: MPC8269 Chip Select Assignments

Chip Select Line	Controlled Device	Address Range
$\overline{\text{CS0}}$	Flash Boot PROM	FFF0_0000 to FFF7_FFFF h
$\overline{\text{CS1}}$	SDRAM	0000_0000 to 07FF_FFFF h
$\overline{\text{CS2}}$	Local Bus SSRAM	2080_0000 to 2083_FFFF h
$\overline{\text{CS3}}$	Application Flash	1000_0000 to 10FF_FFFF h
$\overline{\text{CS4}}$	PAL General Purpose Registers	2000_0000 to 2000_00ff h
$\overline{\text{CS5}}$	QuadFALC	2010_0000 to 2010_7FFF h
$\overline{\text{CS6}}$	H.100 Switch Chip	2020_0000 to 2020_7FFF h
$\overline{\text{CS7}}$	Not Used	
$\overline{\text{CS8}}$	Not Used	2030_0000 to 2030_7FFF h
$\overline{\text{CS9}}$	Not Used	2040_0000 to 2040_7FFF h
$\overline{\text{CS11}}$	Not Used	
None	Power Span mapped space	3000_0000 to 3000_0fff

60x Bus Connections

The MPC8260 has its external bus configured in the Power PC 60x mode. The bus features a 32-bit non-parity address bus. The address bus is pulled up by 10K resistor packs to V3V to prevent unknown bus conditions during no drive periods on the bus. The 64-bit non-parity data bus is also pulled up to V3V with 10K resistor packs to prevent unknown bus conditions during no drive periods on the bus. The bus runs at a 66 MHz clock rate and because of this the length of the bus is restricted to 6in. or less with a very limited number of peripheral connections.

There are some multifunction pins that are used for bus control. These assignments have been made as follows:

- For multiple beat transfers, the MPC8260 is configured to supply the BADDRXX lines as the lower order address bits.
- The PBSx/$\overline{\text{PSDDQMx}}$/$\overline{\text{PWEx}}$ lines are configured as PSDDQMx.
- PGPL5/PSDAMUX is configured as PSDAMUX for SDRAM control.
- MODCLKx/Apx/TCx/BANKSELx are used as the MODCLKx inputs during hard reset and as BNKSELx for the SDRAM.
- PGPL3/$\overline{\text{PSDCAS}}$ is used as the $\overline{\text{PSDCAS}}$ for the SDRAM.
- PGPL2/$\overline{\text{PSDRAS}}$/$\overline{\text{POE}}$ is used as the $\overline{\text{PSDRAS}}$ for the SDRAM.
- PGPL1/$\overline{\text{PSDWE}}$ is used as the $\overline{\text{PSDWE}}$ for the SDRAM.
- PGPL0/PSDA10 is used as PSDA10 for the SDRAM.

The external cache lines are not used and are therefore not connected. The internal CPU arbiter is used as the bus arbiter. All of the normal 60x bus transfer size (TSIZE0-3) and transfer type (TT0-4) lines are supported. The rest of the transfer control signals are used in their normal mode. All of the control signals are pulled to V3V with 10K (min.) resistors to prevent accidental activation during no drive periods on the bus.

MPC8260 Interrupt Sources

Several multifunction pins are used to supply the MPC8260 with the direct connect interrupts from the various board peripherals. The IRQ0 to IRQ7 lines are used along with some of the PORT C interrupt capable pins. The interrupt sources, for the most part, have multiple interrupt conditions. Refer to the individual component subsections or the component's user manual for the complete breakdown of interrupt causes. Table 4-5, “MPC8260 Interrupt Source,” shows the connections from these devices to the MPC8260:

Table 4-5: MPC8260 Interrupt Source

IRQ Level	Pin Number	Controlled Device
IRQ0	T1	NMI Interrupt from the Abort Jumper P3-4 to P3-5
IRQ1	A22	QuadFALC 1 General Interrupt
IRQ2	E21	Not Used
IRQ3	D21	Not Used
IRQ4	C21	PowerSpan General Interrupt
IRQ5	B21	H.100 Switch FAIL_A, FAIL_B, LREF0 and LREF1 Fail
IRQ6	A21	Not Used
IRQ7	E20	Not Used
PC0	AB26	Not Used
PC1	AD29	Not Used
PC2	AE29	Not Used
PC3	AE27	Not Used

Local Bus Connections

The MPC8260 Local Bus is configurable to be a PCI interface or a slave device, local memory bus. This bus is closely coupled to the MPC8260s CPM processor and its most effective use is as memory for the CPM. In the PCI384 implementation, the bus is configured as a slave device, local memory bus for a bank of Synchronous Static RAM (SSRAM). The local bus provides address, data and control for the SSRAM. The cycle control for the SSRAM is provided by the UPMB and the CS2s address space decode. The Local Bus multifunction pins are configured for the SSRAM as follows:

- LGPL0/LSDA10 is configured as LGPL0 and controls the SSRAM output enable signal (LCL_OE).
- LGPL1/LSDWE is configured as LGPL1 and controls the SSRAM Address Control Strobe signal (LCL_ADSC).
- LGPL2/LSDRAS/LOE is configured as LGPL2 and controls the SSRAM Address Advance Strobe signal (LCL_ADV)
- LGPL3/LSDCAS is configured as LGPL3 and controls the SSRAM Chip Enable signal (LCL_CE).
- The LBSx/LSDDQMx/LWEx are configured as LBSx and control the SSRAM byte strobes (LCL_LBSx).
- The PCI_C/BEx/LCL_DPx are configured as DPX and control the SSRAM data parity bits (LCL_DPx).
- LPL5 and LWR are not used in this design.

MPC8260 Parallel I/O Ports

The CPM supports four general-purpose I/O-ports A, B, C, and D. Each pin in the I/O ports can be configured as a general-purpose I/O signal or as a dedicated peripheral interface signal. Port C is unique in that 16 of its pins can generate interrupts to the internal interrupt controller.

Each pin can be configured as an input or output and has a latch for data output, read or written at any time, and configured as general-purpose I/O or a dedicated peripheral pin. Some of the pins can be configured as open-drain (the pin can be configured in a wired-OR configuration on the board). The pin drives a zero voltage but three-states when driving a high voltage. Note that port pins do not have internal pull-up resistors. Any unused I/O pins are programmed to dedicated outputs and are programmed to be a Logic 1. Due to the CPMs significant flexibility, many dedicated peripheral functions are multiplexed onto the ports. The functions are grouped to maximize the pins' usefulness in the greatest number of MPC8260 applications. Refer to the MPC8260 user's manual for more information on the various peripheral setups.

Serial Management Controllers SMC 1 and SMC 2

The MPC8260 features two, general-purpose serial management controllers (SMCs), that may be used as general purpose RS232 communications interfaces. The two SMCs are full-duplex ports that can be configured independently to support one of three operating or modes:

- UART
- Transparent
- General Circuit Interface (GCI)

The PCI384 is designed to support UART operation on one port. A mini-DB9 connector, located on the PCI384 faceplate provides connectivity to the SMC which is configured as the Console port. The console port can be used to provide a debug, monitor or download function. The console port also has the ability to generate a PQ_HRESET if it receives a RS232 break signal that lasts for longer than 3.2ms. The logic that watches the console receive line is located in the PSM519 PAL. The reset on break detect can be enabled by setting jumper P3-5 to P3-6. Removing the jumper will disable the reset action.

The secondary Serial I/O channel is not used.

PCI384 TDM for T1/E1 channels on TDM_B1 and TDM B2

In general, the MPC8260 can support eight full duplex TDMs on two SIs (1, 2). Each SI supports four TDMs (A, B, C, D). Each SI has an SIRAM with 256 entries for Transmit, 256 entries for Receive. Each of these 256 entries is divided into four banks of 64 entries each. Only one entry is needed to route a timeslot to a particular MCC channel. Allocation the SIRAM for each TDM is done in 32 entries segments. The user can define the size of each SI RAM that is related to a certain TDM channel by programming the starting bank of that TDM.

On the PCI384 both Serial Interfaces (SI) in the MPC8260 are configured to support up to 256 data or voice channels on TDM_B1 and TDM_B2.

MPC8260 Parallel Port Pin Assignments

The program settings for the parallel port pins can be found in the files `init.inc` and `hardware.c`. Note that all unassigned channels on all ports should be programmed to be data register outputs wherever no conflicts or limitations in the part exist. This will prevent unwanted current draw on the unused channels.

MPC8260 Port A Pin Assignments

Table 4-6, "MPC8260 Port A Pin Assignments," shows the MPC8260 Port A Pin assignments.

Table 4-6: MPC8260 Port A Pin Assignments

Pin	Pin Function					
	PPARA = 1				PPARA = 0	
	PSORA = 0		PSORA = 1			
	PDIRA =1 Out	PDIRA = 0 In	PDIRA = 1 Out	PDIRA = 0 In	PDIRA = 1 Out	PDIRA = 0 In
PA31	Default =Unassigned					
PA30	Default = Unassigned					
PA29	Default = Unassigned					
PA28	Default = Unassigned					
PA27	Default = Unassigned					
PA26	Default = Unassigned					
PA25	Default = Unassigned					
PA24	Default = Unassigned					
PA23	Default = Unassigned					
PA22	Default = Unassigned					
PA21	Default = Unassigned					
PA20	Default = Unassigned					
PA19	Default = Unassigned					
PA18	Default = Unassigned					
PA17	Default = Unassigned					
PA16	Default = Unassigned					
PA15	Default = Unassigned					
PA14	Default = Unassigned					
PA13	Default = Unassigned					
PA12	Default = Unassigned					
PA11	Default = Unassigned					
PA10	Default = Unassigned					
PA9	Default = Unassigned					
PA8	Default = Unassigned					
PA7	Default = Unassigned					
PA6	Default = Unassigned					
PA5	Default = Unassigned					
PA4	Default = Unassigned					
PA3					EE_CS	
PA2					EE_DO	
PA1						EE_DI
PA0					EE_SCL	

MPC8260 Port B Pin Assignments

Table 4-7, “MPC8260 Port B Pin Assignments,” shows the MPC8260 Port B Pin assignments.

Table 4-7: MPC8260 Port B Pin Assignments

Pin	Pin Function					
	PPARB = 1				PPARB = 0	
	PSORB = 0		PSORB = 1			
	PDIRB = 1 Out	PDIRB = 0 In	PDIRB = 1 Out	PDIRB = 0 In	PDIRB = 1 Out	PDIRB = 0 In
PB31				TDM_B2: L1TXD		
PB30				TDM_B2: L1RXD		
PB29				TDM_B2: L1RSYNC		
PB28				TDM_B2: L1TSYNC		
PB27	Default = Unassigned					
PB26	Default = Unassigned					
PB25	Default = Unassigned					
PB24	Default = Unassigned					
PB23	Default = Unassigned					
PB22	Default = Unassigned					
PB21	Default = Unassigned					
PB20	Default = Unassigned					
PB19	Default = Unassigned					
PB18	Default = Unassigned					
PB17	Default = Unassigned					
PB16	Default = Unassigned					
PB15	Default = Unassigned					
PB14	Default = Unassigned					
PB13	Default = Unassigned					
PB12	Default = Unassigned					
PB11	Default = Unassigned					
PB10	Default = Unassigned					
PB9	Default = Unassigned					
PB8	Default = Unassigned					
PB7	Default = Unassigned					
PB6	Default = Unassigned					
PB5	Default = Unassigned					
PB4	Default = Unassigned					

MPC8260 Port C Pin Assignments

Table 4-8, “MPC8260 Port C Pin Assignments,” shows the MPC8260 Port C Pin assignments.

Table 4-8: MPC8260 Port C Pin Assignments

Pin	Pin Function					
	PPARC = 1				PPARC = 0	
	PSORC = 0		PSORC = 1			
	PDIRC = 1 Out	PDIRC = 0 In	PDIRC = 1 Out	PDIRC = 0 In	PDIRC = 1 Out	PDIRC = 0 In
PC31	Default = Unassigned					
PC30	Default = Unassigned					
PC29		Assign CLK3 to TDMB1 RxCLK				
PC28		Assign CLK4 to TDMB1 TxCLK				
PC27	Default = Unassigned					
PC26	Default = Unassigned					
PC25	Default = Unassigned					
PC24	Default = Unassigned					
PC23	Default = Unassigned					
PC22	Default = Unassigned					
PC21	Default = Unassigned					
PC20	Default = Unassigned					
PC19	Default = Unassigned					
PC18	Default = Unassigned					
PC17		Assign CLK15 to TDMB2 RxCLK				
PC16		Assign CLK16 to TDMB2 TxCLK				
PC15	Default = Unassigned					
PC14	Default = Unassigned					
PC13	Default = Unassigned					
PC12	Default = Unassigned					
PC11	Default = Unassigned					
PC10	Default = Unassigned					
PC9	Default = Unassigned					
PC8	Default = Unassigned					
PC7	Default = Unassigned					
PC6	Default = Unassigned					
PC5	Default = Unassigned					
PC4	Default = Unassigned					
PC3	Default = Unassigned					
PC2	Default = Unassigned					
PC1	Default = Unassigned					
PC0	Default = Unassigned					

MPC8260 Port D Pin Assignments

Table 4-9, “MPC8260 Port D Pin Assignments,” shows is the MPC8260 Port D Pin assignments.

Table 4-9: MPC8260 Port D Pin Assignments

Pin	Pin Function					
	PPARD = 1				PPARD = 0	
	PSORD = 0		PSORD = 1			
	PDIRD = 1 Out	PDIRD = 0 In	PDIRD = 1 Out	PDIRD = 0 In	PDIRD = 1 Out	PDIRD = 0 In
PD31	Default = Unassigned					
PD30	Default = Unassigned					
PD29	Default = Unassigned					
PD28	Default = Unassigned					
PD27	Default = Unassigned					
PD26	Default = Unassigned					
PD25	Default = Unassigned					
PD24	Default = Unassigned					
PD23	Default = Unassigned					
PD22	Default = Unassigned					
PD21	Default = Unassigned					
PD20	Default = Unassigned					
PD19	Default = Unassigned					
PD18	Default = Unassigned					
PD17	Default = Unassigned					
PD16	Default = Unassigned					
PD15	Default = Unassigned					
PD14	Default = Unassigned					
PD13				TDM_B1: L1TXD		
PD12				TDM_B1: L1RXD		
PD11				TDM_B1: L1TSYNC		
PD10				TDM_B1: L1RSYNC		
PD9	SMC1: SMTXD					
PD8		SMC1: SMRXD				
PD7	Default = Unassigned					
PD6	Default = Unassigned					
PD5	Default = Unassigned					
PD4	Default = Unassigned					

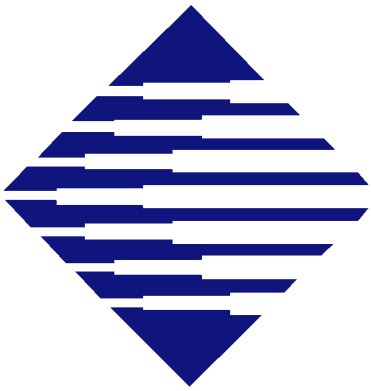
JTAG Support

The JTAG testing port on the MPC8260 is used to support the EST COP debugger on the P4 connector. The connector is setup to support the extended 16-pin COP debugger signaling, but the basic 10 pin signaling devices may be used with an interposing adapter. The P4 pinout is shown in Table 4-10, “JTAG Pinouts,” below.

Table 4-10: JTAG Pinouts

Pin Number	Signal Name
1	PQ_TDO - JTAG Test Data Out signal
2	$\overline{\text{PQ\_QACK}}$- Quiescent State Acknowledge, not supported
3	PQ_TDI- JTAG Test Data In signal
4	$\overline{\text{PQ\_TRST}}$- JTAG Reset and Tri-state signal
5	$\overline{\text{PQ\_QREQ}}$- Quiescent State Request
6	V3V
7	PQ_TCK - JTAG Test Clock
8	No Connection
9	PQ_TMS - JTAG Test Mode Select
10	No Connection
11	$\overline{\text{PQ\_SRESET}}$ - MPC8260 Soft Reset
12	Ground
13	$\overline{\text{PQ\_HRESET}}$ - MPC8260 Hard Reset
14	No Connection
15	$\overline{\text{CHKSTPO}}$ - Checkstop output, not supported
16	Ground

All of the control signals are pulled to V3V with a 10K resistor to prevent false assertion when no JTAG controller is connected.



Chapter

5

SDRAM

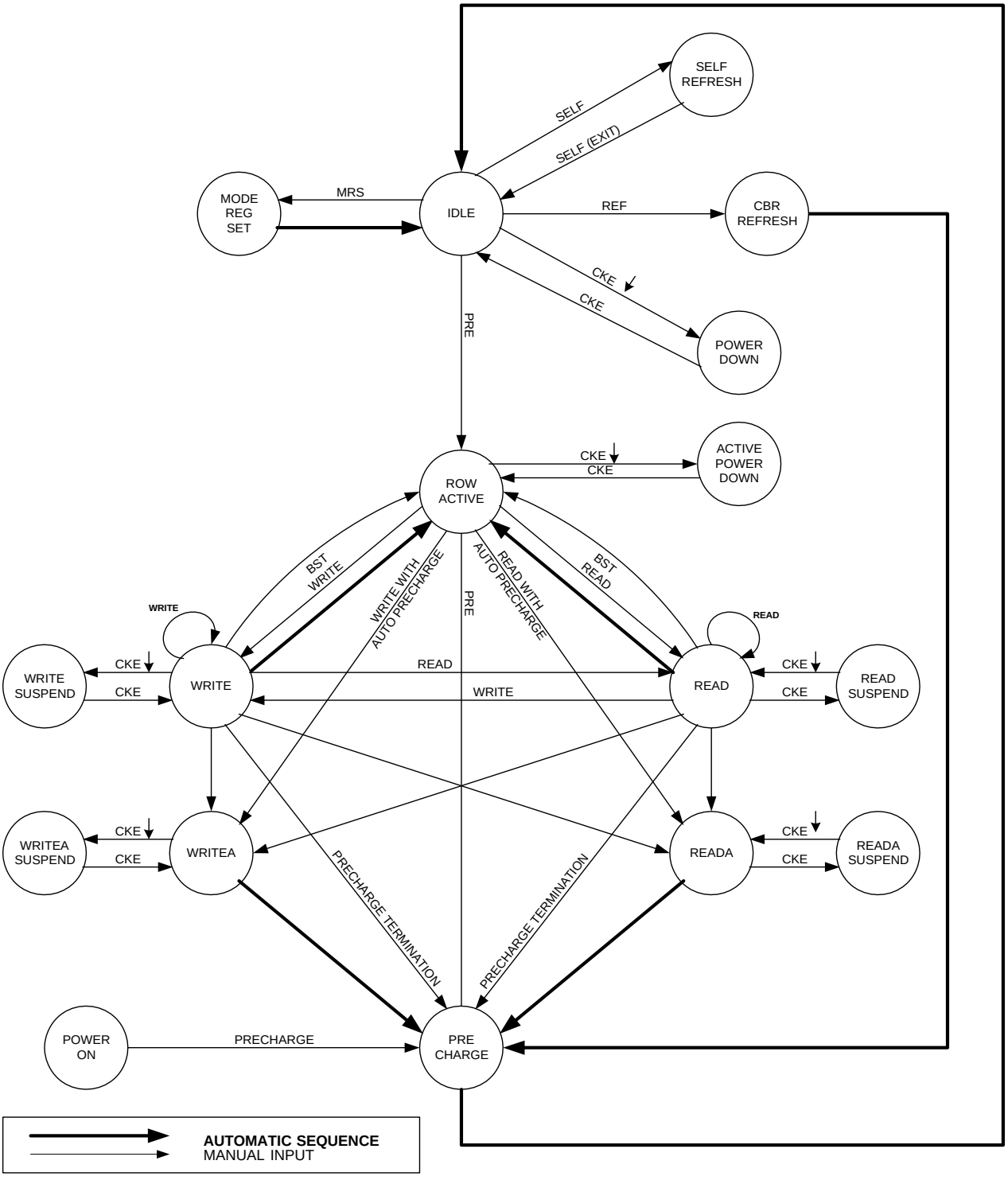
Overview

The SDRAM memory bank is comprised of individual chips mounted on the component and circuit sides of the base board. The SDRAM architecture provides the ability to synchronously burst data at a high data rate with automatic column address generation, the ability to interleave between internal banks in order to hide PRECHARGE time, and the capability to randomly change column addresses on each clock cycle during a burst access.

Figure 5-1, “Partial FSDRAM State Diagram,” on page 50 shows a Partial FSDRAM State Diagram.

Key topics in this chapter include:

Figure 5-1: Partial FSDRAM State Diagram



This synchronous DRAM design provides support for a bank of 3.3 V SDRAM. A density of 64 MB or 128 MB is supported at 66 MHz. The data port size is 64-bits. Clock drive to the module is provided from the same low-skew clock driver that provides 66 MHz to the MPC8260. The clock enable signals (CKE0 and CKE1) are permanently enabled, thus the SELF-REFRESH command and operation are unavailable. The MPC8260 can be programmed to provide refresh as needed. The 64-bit data bus, eight data read/write mask signals, and the three command signals (WE#, CAS#, and RAS#), and the chip select (CS#) is provided directly from the MPC8260. The MPC8260 will automatically assert CS# during refresh cycles. Multiplexed row/column address bits (MA0 through MA12) and bank select signals (BA0 and BA1) are driven to the SDRAM bank from PSM517 (see Figure 5-2, “External Master SDRAM Configuration,” on page 53). The PSM517 PLD creates these from the MPC8260 address and bank select bits that have been latched with the processors ALE signal and multiplexed with the processors SDAMUX signal. Burst address bits (BADDR27 and BADDR28) are utilized directly to provide the least significant bits to the SDRAM during column address presentation. The least significant three address lines 29-31 are unused due to the 8 byte wide data path. The SDRAM uses BADDR27 and BADDR28 to select one of four sequential locations as the first to be burst. Bursts occur up to a maximum of four transfers. The remaining three locations are addressed sequentially, after a mod 4 operation is applied to BADDR27 and BADDR28. SDA10 passes through PSM517 and always drives MA10 on the SDRAM. During the activate command, this pin provides the row address for A10. During a PRECHARGE command this pin selects PRECHARGE all-banks (PSDA10 = 1), or PRECHARGE-SELECTED-BANK (PSDA10 = 0) using BANKSEL[1-2]. Four banks are supported and BANKSEL[0] is unused. During READ and WRITE commands this pin selects AUTO-PRECHARGE (PSDA10 = 1) or no AUTO-PRECHARGE (PSDA10 = 0). Address bit ordering conventions are opposite between the MPC8260 (A0 is most significant bit) and the SDRAM (A0 is least significant bit) these are resolved within PSM517.

Memory Organization

The MPC8260 SDRAM register programming, described in subsequent sections, is in reference to a 64 MB memory organized as 4 banks x 4096 rows x 512 columns x 16 data bits x 4 words and a 128 MB memory organized as 4 banks x 8192 rows x 512 columns x 16 data bits x 4 words. The 60x address bus is programmed to utilize bank interleaving and is partitioned as shown in Table 5-1, “60x Address Bus Partitioning,” below.

Table 5-1: 60x Address Bus Partitioning

	MSB of Bank Address	Bank Select	Row	Column	LSB
128 MB	A0-A4	A5-A6	A7-A19	A20-A28	A29-A31
64 MB	A0-A5	A6-A7	A8-A19	A20-A28	A29-A31

Following initialization, access to the SDRAM is composed of an ACTIVATE and READ/WRITE command. During the ACTIVATE command the bank select and row selection are provided by PSM517 on the SDRAM address lines as shown in Table 5-2, “SDRAM Address Port During Activate Command,” on page 52.

Table 5-2: SDRAM Address Port During Activate Command

SDRAM PINS	NA	BA1-BA0	MA12-MA2	MA1-MA0	NA	
	MPC8260 PINS	A0-A14	BNKSEL [1-2]	A17-A26	BADDR27-BADDR28	A29-A31
128 MB	PHYSICAL ADDRESS	NA	A5-A6	ROW (A7-A17)	A18-A19	NA
64 MB	PHYSICAL ADDRESS	NA	A6-A7	ROW (A8-A17)	A18-A19	NA

During the row-addressing phase of SDRAM access, PSM517 provides bank select and address to the SDRAM as shown in Table 5-3, “SDRAM Address Port During Row Addressing,” below.

Table 5-3: SDRAM Address Port During Row Addressing

SDRAM PINS	NA	BA1-BA0	MA12	MA11	MA10	MA9-MA2	MA1-MA0
128 MB	MPC8260 PINS	BNKSEL[1-2]	A7	A8	SDA10	A10-A17	A18-A19
64 MB	MPC8260 PINS	BNKSEL[1-2]	A7	A8	SDA10	A10-A17	A18-A19
128 MB	PHYSICAL ADDRESS	A5-A6	A7	A8	A9	A10-A17	A18-A19
64 MB	PHYSICAL ADDRESS	A6-A7	A7	A8	A9	A10-A17	A18-A19

During the READ/WRITE command the bank select and column selection are provided by PSM517 on the SDRAM Bank address lines as shown in Table 5-4, “SDRAM Address Port During Read/Write Command,” below.

Table 5-4: SDRAM Address Port During Read/Write Command

SDRAM PINS	NA	BA1-BA0	MA12	MA11	MA10	MA9	MA8	MA7-MA2	MA1-MA0
	MPC8260 PINS	BNKSEL[1-2]	0	0	SDA10	A19	A20	A21-A26	BADDR27-BADDR28
128 MB	PHYSICAL ADDRESS	A5-A6	-	-	AP	-	COLUMN (A20)	COLUMN(A21-A26)	COLUMN(A27-A28)
64 MB	PHYSICAL ADDRESS	A6-A7	-	-	AP	-	COLUMN (A20)	COLUMN(A21-A26)	COLUMN(A27-A28)

During the column-addressing phase of SDRAM access, PSM517 provides bank select and address to the SDRAM as shown in Table 5-5, “SDRAM Address Port During Column Addressing,” below.

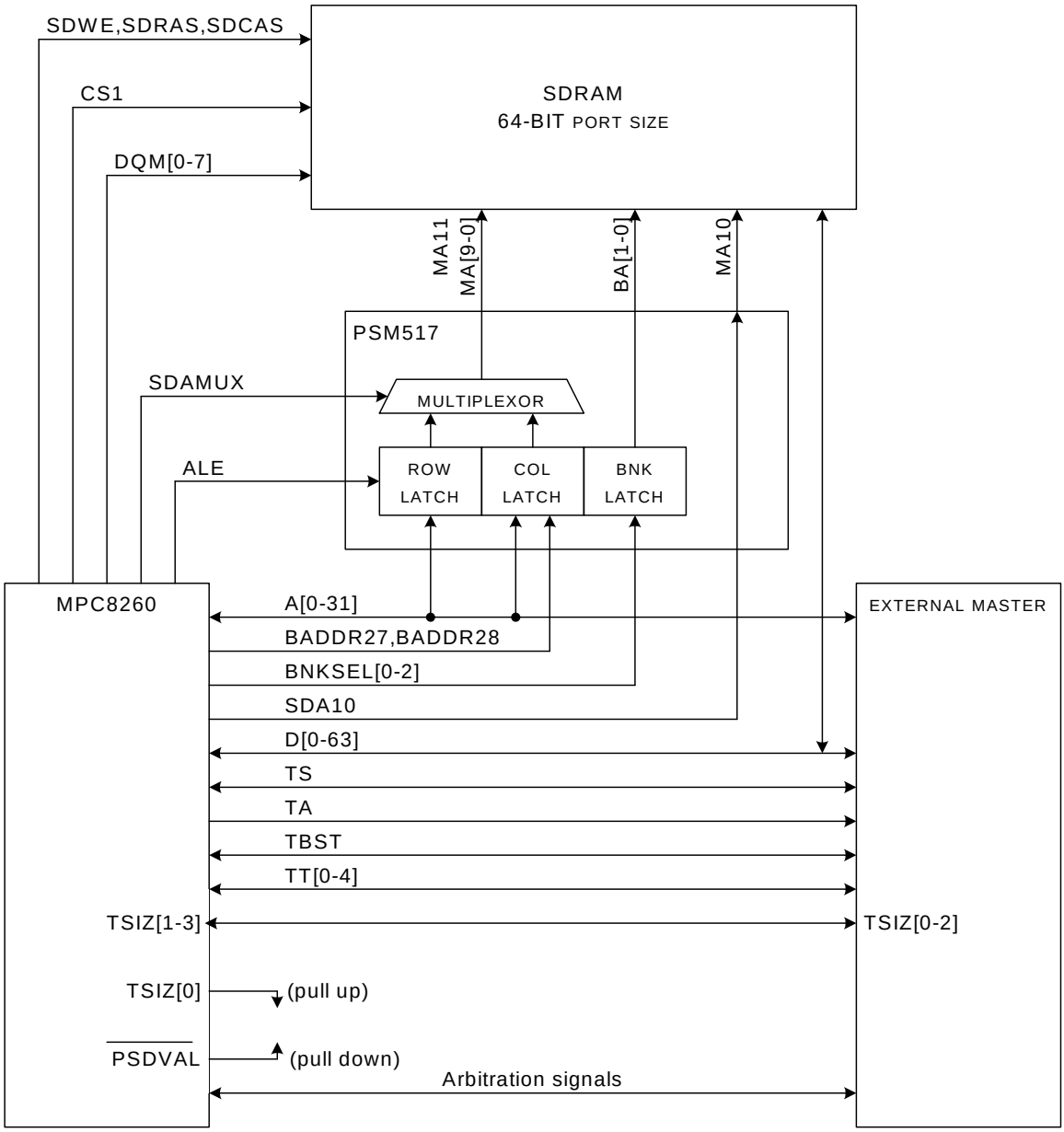
Table 5-5: SDRAM Address Port During Column Addressing

SDRAM PINS	NA	BA1-BA0	MA11	MA10	MA9	MA8	MA7MA2	MA1-MA0
	MPC8260 PINS	BNKSEL[1-2]	0	SDA10	0	A20	A21 A26	BADDR27-BADDR28
128 MB	PHYSICAL ADDRESS	A5-A6	-	AP	-	A20	A21-A26	A27-A28
64 MB	PHYSICAL ADDRESS	A6-A7	-	AP	-	A20	A21-A26	A27-A28

External Master Support

When the MPC8260 is placed in 60x compatible mode, external masters can access the SDRAM as shown in Figure 5-2, “External Master SDRAM Configuration,” below. External masters will not necessarily multiplex row and column addresses for the SDRAM. Therefore, an SDRAM interface (PSM517) has been located between the 60x address bus and SDRAM address bus. 60x addresses are latched with ALE at the start of each memory cycle. The MPC8260 SDAMUX signal is used to multiplex address bits during row and column address presentation to the SDRAM. The MPC8260 monitors the 60x address bus and asserts ALE, SDAMUX, BADDR, and SDA10 appropriately for external masters.

Figure 5-2: External Master SDRAM Configuration



SDRAM Controller Initialization

Prior to SDRAM Controller Initialization the SCMR, BSR, SIUMCR, and SWSR registers are initialized. The MPC8260 documentation recommends the following register order for SDRAM controller initialization: MPTPR, PSRT, PSDMR, OR, BR. A PRECHARGE-ALL-BANKS command is issued next, followed by eight CBR REFRESH commands and the SDRAM load mode register command. The SDRAM and MPC8260 SDRAM Controller will now be completely initialized. The SDRAM Controller can now be placed in normal operating mode by setting the PSDMR[OP] field to 000.

MPC8260 MPTPR and PSRT Register Configuration

The SDRAM must receive 4096 refresh cycles in 64 ms, or a periodic refresh rate of 15.625 us. The Memory Periodic Timer Prescaler Register (MPTPR) and SDRAM Refresh Timer (PSRT) together configure the refresh rate for SDRAM on the 60x bus. Each time the MPC8260 SDRAM refresh timer expires, all banks that qualify (e.g., CS1#) generate a bank staggering auto-refresh request using the SDRAM machine. Memory banks internal to SDRAM devices have common row addresses and are refreshed simultaneously. The product of PSRT and MPTPR must be less than the product of the refresh rate and the BUS clock frequency (66.6 MHz). Therefore, the product of MPTPR and PSRT must be less than 15.625 us x 66.6 MHz = 1041. MPTPR is programmed to 0x32 = 50 and PSRT is programmed to 0xE = 14 to produce a product of 700.

MPC8260 PSDMR Register Configuration

The 60x SDRAM Mode Register (PSDMR) contains PBI, RFEN, OP, SDAM, BSMA, SDA10, RFRC, PRETOACT, ACTTORW, BL, LDOTOPRE, WRC, EAMUX, BUFCMD, and CL fields, as shown in Table 5-6, “PSDMR Register Configuration,” below.

Table 5-6: PSDMR Register Configuration

FIELD	64 MB	128 MB	Description
PBI	0	0	Bank Interleaving
RFEN	1	1	Refresh required
SDAM	001	001	Address Multiplex Size (10.4.5.1)
BSMA	010 (A14-A16)	001 (A13-A15)	Bank select uses right most bits
SDA10	011	011	A10 control = A9
RFRC	100	100	Refresh recovery 6 clocks
PRETOACT	010	010	Precharge to Activate 2 clocks
ACTTORW	010	010	Activate to R/W 2 clocks
BL	0	0	Burst Length = 4
LDOTOPRE	10	10	Last data to precharge = 2 clocks
WRC	10	10	Write recovery = 2 clocks
EAMUX	1	1	Ext address muxing
BUFCMD	1	1	Control strobes = 2 clocks
CL	10	10	CAS Latency = 2 clocks
PSDMR Value	0x414e24ae	0x412e24ae	SDRAM Mode Register Value

PBI is set to 0 to enable bank-based interleaving. RFEN is set to 1 to enable refresh services in the MPC8260. OP defines the command presented to the SDRAM. SDAM is set to specify that external bus address pins A[14-31] will present address signals A[5-22] during SDRAM accesses. The value of BSMA selects which address line are present on the BANKSEL[0-2] pins. The PCI384 uses 4 bank devices so only BANKSEL[1-2] pins are used. SDA10 is set to 011 to select A9 to be presented on the SDA10 pin during an ACTIVATE command to the SDRAM. The value of SDMR[PBI] does/does not need to be programmed prior to programming PSDMR[SDA10]. RFRC is set to 100 to select 6 clocks for the refresh recovery time after an ACTIVATE command and before a REFRESH command. The minimum value of 5 is required for the MICRON 70 ns trc of 70 ns and clock frequency of 66.6 MHz. This discrepancy is not yet resolved. PRETOACT is set to 010 to set the minimum interval (following a PRECHARGE command and preceding an activate/refresh command) to 2 clocks. This should be 2 for the MICRON tRP of 20 nS and clock frequency of 66.6 MHz. ACTTORW is set to 010 to set the minimum interval (following an activate command and preceding a read/write command) to 2 clocks. This should be 2 for the MICRON tRCD of 20 nS and clock frequency of 66.6 MHz. BL is set to 0 to select a burst length of 4, which is required for a port size of 64 bits. LDOTOPRE is set to 10 to set the minimum interval (following the last data READ and preceding a PRECHARGE command) to 1 clock. This is less than the ap-note recommended value of tDPL, but consistent with the SDRAM datasheet recommendation of CAS-latency - 1 = 1. WRC is set to 10 to set the minimum interval (following the last data WRITE and preceding a PRECHARGE command) to 2 clocks. This is greater than the ap-note recommended, MICRON tCDL value of 1, but consistent with the WRITE recovery specification of 1CLK + 7 nS. EAMUX is set to 1 to cause the memory controller to add another cycle for each address phase. This compensates for delays through the external address multiplexing circuitry. BUFCMD is set to 1 to select relaxed timing, which is recommended when address or command lines are buffered. CL is set to 10 to select a CAS latency of 2. This matches the SDRAM mode configuration word setting of CAS latency.

MPC8260 OR Register Configuration

The Option Register (OR) contains SDAM, LSDAM, BPD, ROWST, NUMR, PMSEL, and IBID fields, as shown in Table 5-7, “MPC8260 OR Register Configuration,” below.

Table 5-7: MPC8260 OR Register Configuration

FIELD	64 MB	128 MB	Description
SDAM	1111_1100_0000_0000_0	1111_1000_0000_0000_0	SDRAM Address Mask
LSDAM	0000_0	0000_0	Lower Address Mask
BPD	01 (4 banks)	01 (4 banks)	Banks per device
ROWST	010 (A8)	001 (A7)	Row start address
NUMR	011 (12 Rows)	100 (13 Rows)	Number of Row Address
PMSEL	0	0	Page mode select
IBID	1	1	Internal Interleaving

SDAM is set to specify the mask for the expected address range. BPD is set to 01 to specify four internal banks per device. ROWST sets the demultiplexed row start address bit. NUMR sets the number of row address lines (i.e., 12 or 13). PMSEL is set to 0 to select the back-to-back page mode (normal operation). IBID is set to 1 to disable bank interleaving between internal banks of an SDRAM device connected to the chip-select line.

MPC8260 BR Register Configuration

The Base Register (BR) contains the BA, PS, DECC, WP, MS, EMEMC, ATOM, DR, and V fields, as shown in Table 5-8, “MPC8260 BR Register Configuration,” below.

Table 5-8: MPC8260 BR Register Configuration

FIELD	64 MB	128 MB	Description
BA	0000_0000_0	0000_0000_0	Base Address
PS	00	00	Port Size
DECC	00	00	Data ECC
WP	0	0	Write Protect
MS	010	010	Machine Select
EMEMC	0	0	Ext MEMC Cntl
ATOM	00	00	Atomic Operation
DR	0	0	Data Pipelining
V	1	1	Valid Bit

BA is set to 0000_0000_0 to specify a base address of 0 for SDRAM memory. PS is set to 00 to specify a data port size of 64-bits. DECC is set to 00 to disable data error checking and correction. WP is set to 0 in order to enable read and write accesses. MS is set to 010 in order specify that SDRAM is located on the 60x bus, as opposed to the local bus. EMEMC is set to 0 to specify that accesses will be handled according to MSEL, as opposed to an external memory controller. ATOM is set to 00 to indicate that the SDRAM memory space is not used for atomic operations. DR is set to 0 to specify that no data pipe lining will be performed. V is set to 1 to indicate that this memory bank is valid, which allows the chip select to be asserted.

SDRAM Initialization Requirements

The SDRAM must be powered up and initialized in a predefined manner. After a stable clock and power are provided to the device, a 100 us period must elapse before a command is issued (other than COMMAND INHIBIT or NOP). COMMAND INHIBIT or NOP should be applied during this 100 uS period until the end of the period. Once the 100 us period has been satisfied with COMMAND INHIBIT or NOP, a PRECHARGE command should be applied to PRECHARGE all banks and transition to the IDLE state. Once in the IDLE state, two AUTO REFRESH cycles must be performed. The mode register powers up in an undefined state and must now be loaded prior to applying any operational command. Assuming the refresh rate has been programmed in the MPC8260 refresh controller, refresh may now be enabled.

COMMAND INHIBIT

After power-up, a 66 MHz clock is driven to the SDRAM. Chip select from the processor is inactive and thus provides the COMMAND INHIBIT command.

PRECHARGE

Precharging all banks will place the SDRAM into the IDLE state. In order to precharge all banks the MPC8260 requires the PSDMR[OP] field be set to 101. The SDRAM requires A10 to be high in order to precharge all banks. This is accomplished by connecting SDA10 from the MPC8260 to A10 of the SDRAM during row-address presentation. PRECHARGE of individual banks is supported in the same way. SDA10 from the MPC8260 is provided to A10 of the SDRAM during column-address presentation to support the AUTO-PRECHARGE command.

AUTO REFRESH

The AUTO REFRESH access occurs when the MPC8260 PSDMR register is accessed at 0xFF01_0190 with the OP field set to 001. This causes a CAS# before RAS# (CBR) refresh command to be presented to the SDRAM.

SDRAM Mode Register Configuration

The mode register must be loaded when all banks are idle, and the controller must wait the specified time before initiating the subsequent operation. The mode register is used to define the specific mode of operation of the SDRAM. This definition includes the selection of a burst length, a burst type, CAS latency, an operating mode and a write burst mode. The LOAD MODE REGISTER command is selected when CS#, RAS#, CAS#, and WE# are low. This access occurs when the MPC8260 PSDMR register is accessed at 0xFF01_0190 with the OP field set to 011. The SDRAM initialization routine sets burst length to 4, burst type to SEQUENTIAL, CAS latency to 2, operating mode to standard, and the write burst mode to PROGRAMMED. A burst length of 4 is required by the MPC8260 for 64-bit port size. A burst type of SEQUENTIAL is required for 60x-cache-wrap compatibility.

Chapter

6

TDM Clocking

Overview

The TDM clock distribution is centralized about the Zarlink MT90866 H.100 TDM switch located on the mezzanine I/O board. This device generates all of the local TDM Frame Pulses and data clocks. It uses a variety of clock sources to lockup its internal digital phase locked loop (DPLL) to a system reference. This ability ensures that the data transferred on the board stays in synchronization with the system and no timing slips occur.

See Figure 6-1, “TDM Clock Tree,” on page 63, for a visual representation of the circuit.

Key topics in this chapter include:

H.100 DPLL

The core of the part is the DPLL. It has a 20 MHz 25 ppm signal source as its primary clock input. This clock directly drives the DPLL. The DPLL accuracy and the 25 ppm input clock, allows the H.100 switch to function as a Stratum 4 Enhanced clock source when it runs in a free run mode. In a normal integration into a Telephony system, synchronized timing is maintained by locking the PLL to a system-supplied reference. Some of the important PLL specifications are listed in Table 6-1, “H.100 DPLL Operating Specifications,” below.

Table 6-1: H.100 DPLL Operating Specifications

Parameter	Value
Acceptable Reference Clock Input Frequencies	1.544 MHz, 2.048 MHz or 8 kHz per AT&T TR62411
Skew Control	Eight taps @ 1.9 ns each, allows 0 to13.3 ns total skew adjustment between reference input and clock outputs. This translates to static phase offset of 1.28 us to 1.293 us for reference clock input of 1.544 MHz. For a reference clock input of 8 kHz or 2.048 MHz, the static phase offset is 960 ns to 973 ns.
Phase Offset, dependent on reference clock frequency	For 1.544 MHz reference clock, maximum phase offset +/- 1.28 us, minimum phase adjustment 10 ns. For 8 kHz or 2.048 MHz reference clock, maximum phase offset +/- 0.96 us, minimum phase adjustment 7.5 ns.
Phase Slope Limiter	Maximum phase slope response for input transient 4.6 ns per 125 us. Meets AT&T TR62411 standard.
Loop Filter	Equivalent to a first order low pass filter with 1.52 Hz cutoff.
Intrinsic Jitter	6.25 nspp for the 80 MHz Master clock.
Jitter Transfer	Rate limited by Phase Slope Limiter to 4.6 ns/125 us. Jitter Transfer Function Cutoff frequency 1.52 Hz with slope of 20 db/decade. Refer to the datasheet for the actual curves and the UI calculations.
Frequency Accuracy	Master Clock/ppm + DPLL/ppm= 25+.03=25.03 ppm of selected frequency.
Holdover accuracy	Same as above calculation
Locking Range	+/- 273 ppm
Maximum Time Interval Error (MTIE)	21 ns for every reference switch.
Phase Continuity	Maintained to within 4.6 ns at the instance (over one frame) of all reference switches.
Phase Lock Time	Less than 25 seconds

As mentioned above, the DPLL can choose from several input sources to obtain the signal it needs to lock to. This mechanism is what supplies the system synchronization. These clock sources can be selected either internally via the reference mux internal to the part or externally by the PSM521 PAL logic. Refer to the *Zarlink MT90866 WAN Access Switch Datasheet* for further information.

Internal Reference Clock Sources

The internal reference selector mux for the DPLL has a redundant configuration. The primary and secondary muxes both have access to the same set of clock sources. These sources include the any of the six local reference inputs, the A and B CTbus clocks or the CT Netreferences from the CTbus.

Internal Reference Monitors

There are two Reference Monitor Circuits: one for the primary reference (PRI_REF) and one for the secondary reference (SEC_REF). These two circuits monitor the selected input reference signals and detect failures by setting up the adequate internal fail outputs (FAIL_PRI and FAIL_SEC). These fail signals are used in the Autodetect mode as the internal LOS_PRI and LOS_SEC signals to indicate when the reference is failed. Method of generating failure depends on the selected reference:

- For all references, the “minimum 90 ns” check is done. This is the requirement by the H.100 specifications. Both, low level and high level of the reference must last for minimum 90 ns each.
- The “period in the specified range” check is done for all references. The length of the period of the selected input reference is checked if it is in the specified range. For the E1 (2.048 MHz clock) or the T1 (1.544 MHz clock) reference, the period of the clock can vary within the range of 1 +/- 1/4 of the defined clock period which is 488 ns for the E1 clock and 648 ns for T1 clock. For the 8 KHz reference, the variation is from 1 +/- 1/32 period.
- If the selected reference is E1 or T1, “64 periods in the specified range” check is done. The selected reference is observed for long period (64 reference clock cycles) and checked if it is within the specified range - from 62 to 66 clock periods.

These reference signal verifications include a complete loss or a large frequency shift of the selected reference signal. When the reference signal returns to normal, the LOS_PRI and LOS_SEC signals will return to logic low.

CT Clock and Frame Monitor Circuits

These monitor circuits check the period of the C8_A and the C8_B clocks and the FRAME_A and FRAME_B frame pulses. According to the H.100 signal specification, the C8 period is 122 ns with a tolerance of +/-35 ns measured between rising edges. If C8 falls outside the range of [87 ns,157 ns], the clock is rejected and the fail signal (FAIL_A or FAIL_B) becomes high. The Frame pulse period is measured with respect to the C8 clock. The frame pulse period must have exactly 1024 C8 cycles. Otherwise, the fail signal (FAIL_A or FAIL_B) becomes high. These two signals are connected to the PSM521 PAL logic and are two of the inputs that form the H.100 Interrupt to the MPC8260.

External Reference Clock Sources

These sources include any of the recovered clocks from the QuadFALC or the Dual External LVDS Clock receivers. The recovered clocks are connected to the PSM521 PAL logic, which is internally programmed to select one of 4 recovered T1 clocks from the framers or the Dual External LVDS Clock receivers. There are two of these selectors and they feed the H100_LREF0 and H100_LREF1 signals to the LREF0 and LREF1 inputs on the H.100 chip.

External Reference Clock Source Monitor

There is a set of clock monitors in the PSM521 logic that will check the external clock sources for validity. The Clock Timer Timeout Period Register contains a clock count value that is used to check for a valid time period for the local reference clocks fed from the PSM521 select multiplexer to the H.100 switch chip. The timer value is the maximum number of 20 MHz clocks that are allowed to elapse during the reference clock period. If the 20 MHz count is exceeded without seeing at least one valid local reference clock the internal lrefxsts bit will be set. This signal is one of the sources for the H.100 interrupt to the MPC8260.

H.100 Clock Operating Modes

Clock distribution and control circuitry support both master and slave modes of H.100 operation. The H.100 switch provides primary and secondary clock and frame synchronization sources for the local TDM streams. These are distributed to the MPC8260, T1/E1/J1 framers, and clock monitoring circuitry. H.100 backplane clocks CT_C8_A and CT_C8_B are supported at 8 MHz operation exclusively. Local clocks are supported at 8 MHz operation exclusively

In master mode, the H.100 switch accepts two of several local clock references that may be recovered by the T1/E1/J1 framers or generated by an optional local precision oscillator. These may be divided down to create an 8 KHz clock reference. The reference clock can be used to drive CTREF to the H.100 backplane for primary or secondary master operation. They are also provided to a PLL that generates and drives the CT_FRAME_A or CT_FRAME_B synchronization, and CT_C8_A or CT_C8_B clock, to the H.100 backplane. The selected local reference clocks are monitored and cause an interrupt upon failure. Local status and masking is provided for these interrupt sources.

In slave mode, the H.100 switch accepts CTREF, CT_C8_A and CT_C8_B clocks. One CTNR1 and one CT_C8 clock are selected for primary operation, while the alternate pair stand-by as secondary clocks. Support is included to allow A-side and B-side source termination resistors on CT_C8_A/B and CT_FRAME_A/B to be bypassed.

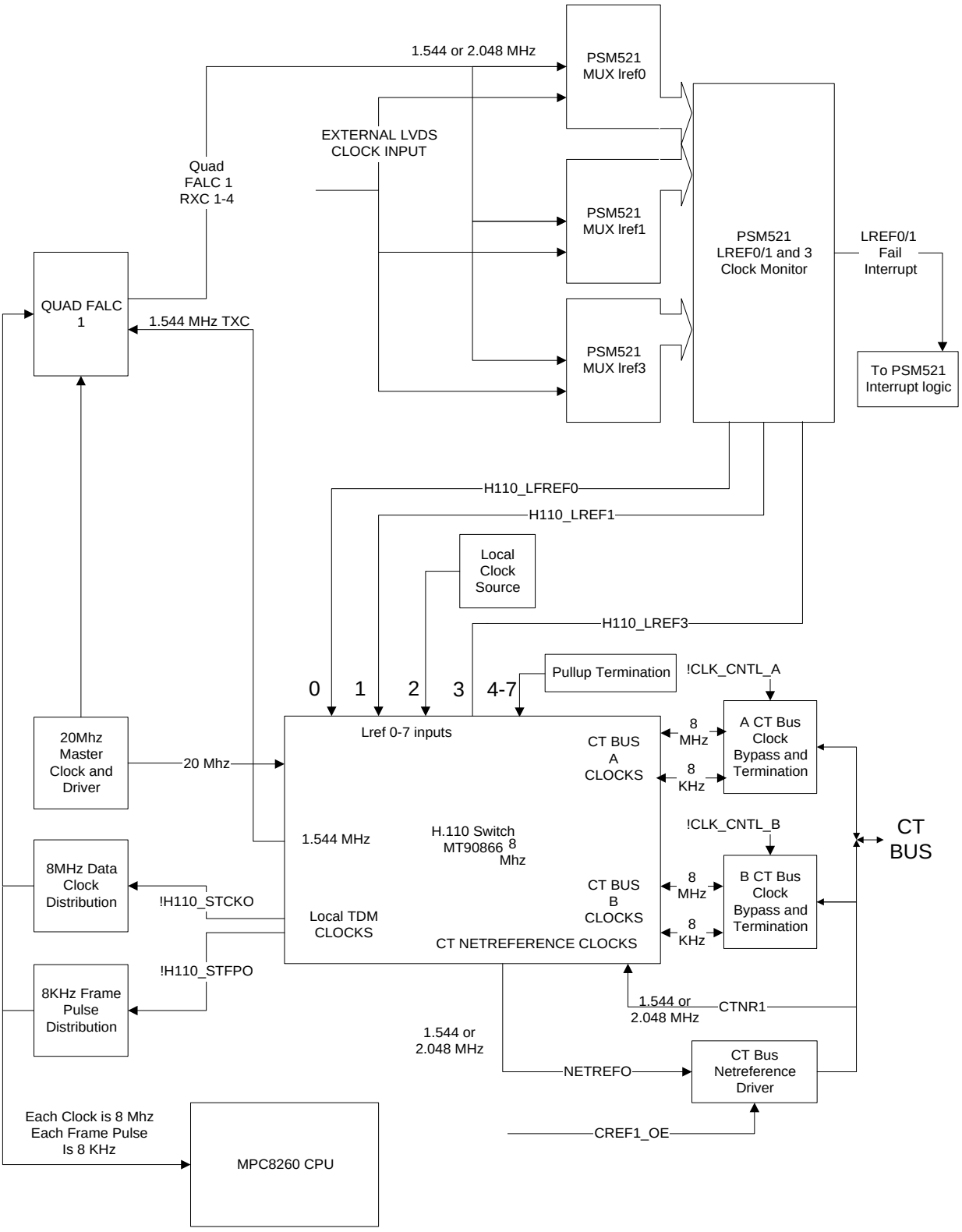
A and B side clock and frame synchronization are monitored by the H.100 switch and failure causes the corresponding FAIL_A or FAIL_B signal to be asserted. This signal is intercepted and provided as an interrupt to the MPC8260. Local status and masking is provided for this interrupt source. A 20 MHz crystal oscillator is distributed to provide a 20 MHz clock for each of the H.100 switch, T1/E1/J1 framers, and general-purpose logic components. These are dedicated clocks without control options.

CT Netreference Sources

The switch can also source the CTbus Netreference signal. The H100_LREF3 signal is supplied to the switch as the reference input for this function. This signal source includes any of the recovered clocks from the QuadFALC or the Dual External LVDS Clock receivers. The recovered clocks are connected to the PSM521 PAL logic, which is internally programmed to select one of four recovered T1 clocks from the framers or the Dual External LVDS Clock receivers. Bits in the Clock Control A and B and Netref Control register control the output of the NETREFO signal onto the CTbus to drive netreference.

For more information on the switch setup and control, refer to the *Zarlink MT90866 WAN Access Switch Datasheet*.

Figure 6-1: TDM Clock Tree



Chapter

7

H.100 Bus and Digital Switch

Overview

This chapter discusses the Zarlink MT90866 switch found on the on the mezzanine I/O board on the PCI384.

Key topics in this chapter include:

H.100 Bus and Digital Switch Features

The PCI384 uses the Zarlink MT90866 switch to connect the various Time Domain Multiplexed interface streams together. The switch resides on the mezzanine I/O board. It has local side connections to the MPC8260s TDM I/O and the QuadFALCs TDM bus. Its Backplane connection is to the 32-bit H.100 CTbus. The switch features:

- 3.3 V operation with 5 V tolerant inputs and I/Os
- 5 V tolerant PCI drivers on CT-Bus I/Os
- 2,432 x 2,432 non-blocking switching among local streams
- 4,096 x 2,432 blocking switching between backplane and local streams
- 2,048 x 2,048 non-blocking switching among backplane streams
- Rate conversion between backplane and local streams
- Rate conversion among local streams
- Backplane interface accepts data rate of 8.192 Mb/s

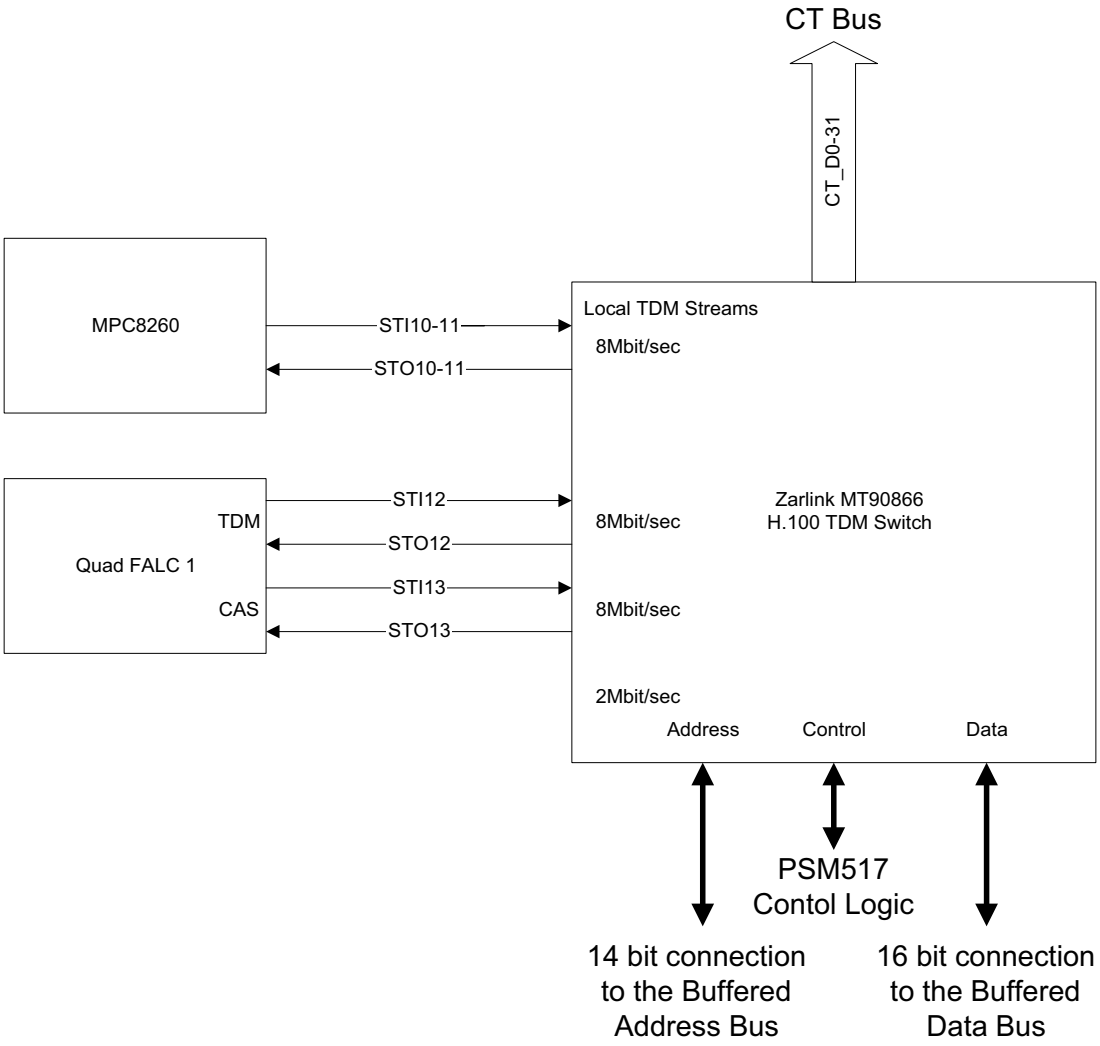
- Local interface accepts data rates of 2.048 Mb/s, 4.096 Mb/s, or 8.192 Mb/s
- Sub-rate switching (2 or 4 bits) configuration for local streams at a data rate of 2.048 Mb/s
- Per-channel variable or constant throughput delay
- Fully compliant to H.100 timing specification
- Per-stream input delay, programmable for local streams on a per bit basis
- Per-stream output advancement, programmable for backplane and local streams
- Per-channel direction control for backplane streams
- Per-channel message mode for backplane and local streams
- Compatible to Stratum 4 Enhanced clock switching standard
- Integrated PLL conforms to Bellcore Stratum 4 Enhanced switching standard
- Holdover mode with hold over frequency of 0.07 ppm
- Wander attenuation from 1.5 Hz.
- Time interval error (TIE) correction
- H.100 Bus Master and Slave mode operation
- Connection memory block programming for fast device initialization
- Tristate-control outputs for external drivers
- Pseudo-Random Binary Sequence (PRBS) pattern generation and testing for backplane and local streams

General Description

The MT90866 Digital Switch provides switching capacities of 4,096 x 2,432 channels between backplane and local streams, 2,432 x 2,432 channels among local streams and 2,048 x 2,048 channels among backplane streams. The local connected serial inputs and outputs have 32, 64 and 128 64kb/s channels per frame with data rates of 2.048, 4.096 and 8.192 Mb/s respectively. The backplane connected serial inputs and outputs have 128 64kb/s channels per frame with data rate of 8.192 Mb/s. The MT90866 also offers a sub-rate switching configuration, which allows 2-bit wide 16kb/s or 4-bit wide 32kb/s data channels to be switched within the device. The device has features that are programmable on a per-stream or per-channel basis including message mode, input delay offset, output advancement offset, direction control and high impedance output control.

On the PCI384, the device is connected to the TDM peripherals, the CTbus and the MPC8260 as shown in Figure 7-1, “H.100 Data and Control,” on page 67.

Figure 7-1: H.100 Data and Control



The device is accessed by the MPC8260 for setup and control purposes through the Buffered Data bus. The Buffered data bus supplies a buffered MPC8260 address and a bidirectional 16 bit data bus. The CS_H100A is supplied by the MPC8260 on CS6 and acts as the device chip enable. The BUF_DS is supplied by the PSM517 PAL logic and acts as the device data strobe in Motorola signaling mode. The PSM517 also supplies the BUF_RW# strobe, which is the read/write, enable.

The H.100 device uses a transfer acknowledge signal, (DTA_H100) to signal completion of data transfers on the Buffered Data Bus. This signal is synchronized to the MPC8260 clock in the PSM517 PAL and reissued as PQ_PGTA. This signal is used by the GCPM to recognize the completion of the access and either terminate the cycle in the case of a write or latch the data on the 60x bus in the case of a read.

Data transfers for this device are 16 bit in nature. The device does not support eight-bit transfers, even though the Buffered Data Bus does support them. Therefore sequential addresses in the device are all offset by 2 Hex, 0000, 0002, 0004, etc…. Be aware that an eight-bit write is still allowed to happen but any byte write will still cause a 16-bit write to the device.

MPC8260 Related H.100 Settings

Table 7-1, “CS6 Base Register Settings,” shows the CS6 base register settings.

Table 7-1: CS6 Base Register Settings

Field	Recommended Setting
PS	10 16-bit
DECC	00 Data Errors checking disabled
WP	0 Read and write accesses are allowed
MSEL	000 Machine Select GPCM-60x bus
EMEMC	0 Accesses are handled by the MSEL selected controller. No external memory controller allowed in this space.
ATOM	00 The address space is controlled by the memory controller bank and is not used for atomic operations
DR	0 No data pipelining is done.
V	1 This bank is valid

The Option Register Settings follow the form outlined for the GPCM Machine, as shown in Table 7-2, “CS6 Option Register Settings,” below.

Table 7-2: CS6 Option Register Settings

Field	Recommended Setting
BCTLD	0 $\overline{\text{BCTLx}}$ is asserted upon access to the current memory bank
CSNT	1 $\overline{\text{CS}}$/WE is negated a quarter of a clock early.
ACS	11 $\overline{\text{CS}}$ is output half a clock after the address lines.
SCY	6 Clock Wait states
SETA	1 $\overline{\text{PSDVAL}}$ is generated after external logic asserts $\overline{\text{GTA}}$
TRLX	1 Relaxed timing is generated by the GCPM for this memory region.
EHTR	00 Normal Idle Timing

The address range for the different internal sections of the part is shown in Table 7-3, “H.100 Digital Switch Register Map,” below.

Table 7-3: H.100 Digital Switch Register Map

Address	Register	Function	Type	Initial Value
CS6+0000h	CR	Control Register	Read/Write, 16 Bit	0x0000
CS6+0002h	DMS	Device Mode Selection Register	Read/Write, 16 Bit	0x0000
CS6+0004h	BPM	Block Programming Mode Register	Read/Write, 16 Bit	0x0000
CS6+0008h	LIDR0	Local Input Delay Register 0	Read/Write, 16 Bit	0x0000
CS6+000Ah	LIDR1	Local Input Delay Register 1	Read/Write, 16 Bit	0x0000
CS6+000Ch	LIDR2	Local Input Delay Register 2	Read/Write, 16 Bit	0x0000
CS6+000Eh	LIDR3	Local Input Delay Register 3	Read/Write, 16 Bit	0x0000
CS6+0010h	LIDR4	Local Input Delay Register 4	Read/Write, 16 Bit	0x0000
CS6+0012h	LIDR5	Local Input Delay Register 5	Read/Write, 16 Bit	0x0000
CS6+0014h	LIDR6	Local Input Delay Register 6	Read/Write, 16 Bit	0x0000
CS6+0018h	LIDR7	Local Input Delay Register 7	Read/Write, 16 Bit	0x0000
CS6+001Ah	LIDR8	Local Input Delay Register 8	Read/Write, 16 Bit	0x0000
CS6+001Ch	LIDR9	Local Input Delay Register 90	Read/Write, 16 Bit	0x0000
CS6+0038h	BOAR0	Backplane Output Advancement Register 0	Read/Write, 16 Bit	0x0000
CS6+003Ah	BOAR1	Backplane Output Advancement Register 1	Read/Write, 16 Bit	0x0000
CS6+003Ch	BOAR2	Backplane Output Advancement Register 2	Read/Write, 16 Bit	0x0000
CS6+003Eh	BOAR3	Backplane Output Advancement Register 3	Read/Write, 16 Bit	0x0000
CS6+0040h	LOAR0	Local Output Advancement Register 0	Read/Write, 16 Bit	0x0000
CS6+0042h	LOAR1	Local Output Advancement Register 1	Read/Write, 16 Bit	0x0000
CS6+0044h	LOAR2	Local Output Advancement Register 2	Read/Write, 16 Bit	0x0000
CS6+0046h	LOAR3	Local Output Advancement Register 3	Read/Write, 16 Bit	0x0000
CS6+004Eh	LBIS	Local Bit Error Rate Input Selection Register	Read/Write, 16 Bit	0x0000
CS6+0050h	LBERR	Local Bit Error Rate Register	Read/Write, 16 Bit	0x0000
CS6+0052h	BBIS	Backplane Bit Error Rate Input Selection Register	Read/Write, 16 Bit	0x0000
CS6+0054h	BBERR	Backplane Bit Error Rate Register	Read/Write, 16 Bit	0x0000
CS6+0056h	DOM1	DPLL Operation Mode 1 Register	Read/Write, 16 Bit	0x0000
CS6+0058h	DOM2	DPLL Operation Mode 2 Register	Read/Write, 16 Bit	0x0000
CS6+005Ah	DPOA	DPLL Output Adjustment Register	Read/Write, 16 Bit	0x0000
CS6+005Ch	DHKR	DPLL House Keeping Register	Read/Write, 16 Bit	0x0000

Table 7-3: H.100 Digital Switch Register Map (Continued)

Address	Register	Function	Type	Initial Value
CS6+0400h to 7FFFh		Connection Memory Setup and Connection Memory Data depending on Memory Selection Bits in Control Register		
CS6+0400ht o 16FEh		MS=0 Local Connection Memory Low Bits	Read/Write, 16 Bit	Undefined
CS6+040h0t o 16FEh		MS=1 Local Connection Memory High Bits	Read/Write, 16 Bit	Undefined
CS6+0400ht o 23FEh		MS=2 Backplane Connection Memory	Read/Write, 16 Bit	Undefined
CS6+0400ht o 16FEh		MS=3 Local Data Memory Read Bits	Read/Write, 16 Bit	Undefined
CS6+0400ht o 23FEh		MS=4 Backplane Data Memory Read	Read/Write, 16 Bit	Undefined

For information on how to use the registers in the H.100 TDM switch, refer to the *Zarlink Semiconductor MT90866 WAN Access Switch Datasheet*.

Digital Switch Local TDM Streams Connection

The H.100 switch serves as the focal point for all of the TDM streams from the various on board peripherals. See Figure 7-1, “H.100 Data and Control,” on page 67, for a complete view of the peripheral connections. The switch local I/O connections are grouped as inputs and outputs. The inputs and outputs are also sub-grouped in fours. This allows each sub-group to be run at one of the available data rates, 2.048, 4.096 or 8.192 Mb/s. Some of the sub-groups will also support 2- and 4-bit sub-rate switching when the group is run at the 2.048 Mb/s data rate. The Local TDM clock is sourced from the switch's DPLL and is fixed at 8.192 MHz. The DPLL is locked to a system reference, which is programmable and chosen by the system architect. The Local Frame Pulse signal is 8 KHz and is also supplied by the DPLL locked to the system reference.

Table 7-4, “Digital Switch Local TDM Stream Connection,” lists the peripheral connections for the PCI384 and the data rate at which the link operates.

Table 7-4: Digital Switch Local TDM Stream Connection

Group	Device	Switch Connection	TDM Data Rate	Interleaving Format
Group 1	Input Pulled up	STi0/STo0	8.192 Mb/s	
	Input Pulled up	STi1/STo1	8.192 Mb/s	
	Input Pulled up	STi2/STo2	8.192 Mb/s	
	Input Pulled up	STi3/STo3	8.192 Mb/s	
Group 2	Input Pulled up	Sti4/Sto4	8.192 Mb/s	
	Input Pulled up	Sti5/Sto5	8.192 Mb/s	
	Input Pulled up	Sti6/Sto6	8.192 Mb/s	
	Input Pulled up	Sti7/Sto7	8.192 Mb/s	

Table 7-4: Digital Switch Local TDM Stream Connection (Continued)

Group	Device	Switch Connection	TDM Data Rate	Interleaving Format
Group 3	Input Pulled up	Sti8/Sto8	8.192 Mb/s	
	Input Pulled up	Sti9/Sto9	8.192 Mb/s	
	MPC8260 TDMA via the B1 TDM Port	Sti10/Sto10	8.192 Mb/s	Each DS-0 in the input stream of 127 channels is defined by the MPC8260. Each DS-0 of the output stream is defined by the connection memory.
	MPC8260 TDMB via the B2 TDM Port	Sti11/Sto11	8.192 Mb/s	Each DS-0 in the input stream of 127 channels is defined by the MPC8260. Each DS-0 of the output stream is defined by the connection memory
Group 3	QuadFALC Data Port 0-3	Sti12/ Sto12	8.192 Mb/s	128 DS-0s Bidir
	QuadFALC CAS Port 0-3	Sti13/Sto13	8.192 Mb/s	128 DS-0s Bidir
	Input Pulled up	Sti14/Sto14	8.192 Mb/s	
	Input Pulled up	Sti15/Sto15	8.192 Mb/s	
Group 4	Input Pulled up	Sti16/Sto16	8.192 Mb/s	
	Input Pulled up	Sti17/Sto17	8.192 Mb/s	
	Input Pulled up	Sti18/Sto18	8.192 Mb/s	
	Input Pulled up	Sti19/Sto19	8.192 Mb/s	
	Input Pulled up	Sti20/Sto20	8.192 Mb/s	
	Input Pulled up	Sti21/Sto21	8.192 Mb/s	
	Input Pulled up	Sti22/Sto22	8.192 Mb/s	
	Input Pulled up	Sti23/Sto23	8.192 Mb/s	
	Input Pulled up	Sti24/Sto24	8.192 Mb/s	
	Input Pulled up	Sti25/Sto25	8.192 Mb/s	
	Input Pulled up	Sti26/Sto26	8.192 Mb/s	
	Input Pulled up	Sti27/Sto27	8.192 Mb/s	

Chapter

8

T1/E1/J1 Quad Framer and LIU

Overview

This chapter discusses the T1/E1/J1 quad framer with integral line interface unit (LIU) found on the on the mezzanine I/O board on the PCI384.

Key topics in this chapter include:

QuadFALC Features

The PCI384 has an T1/E1/J1 quad framer with integral line interface unit (four ports total) located on the mezzanine I/O board. The device is the Infineon (Siemens) PEB22554 (QuadFALC), version 3.1. See Figure 8-1, "Quad Framer Block Diagram," on page 76. The QuadFALC supports a multitude of features. Among them are:

- High density, generic interface for all T1/E1/J1 applications
- Four analog receive and transmit circuits for long/short haul applications
- E1 or T1/J1 mode selectable for each channel individually
- Data and clock recovery using an integrated digital phase locked loop
- Maximum line attenuation up to -43 dB at 1024 kHz (E1) and up to -36 dB at 772 kHz (T1/J1)
- Programmable transmit pulse shapes for E1 and T1/J1 pulse masks

- Programmable Line Build-Out for CSU signals according to ANSI T1. 403 and FCC68: 0 dB, -7.5 dB, -15 dB, -22.5 dB (T1/J1)
- Low transmitter output impedances for high transmit return loss
- Tri-state function of the analog transmit line outputs
- Transmit line monitor protecting the device from damage
- Receive line monitor mode
- Jitter specifications of ITU-T I.431, G.703, G.736 (E1), G.823 (E1) and AT&T TR62411 (T1/J1) are met
- Crystal-less wander and jitter attenuation/compensation
- Common master clock reference for E1 and T1/J1 (any frequency within 1.02 and 20 MHz)
- Power down function per channel
- Support of automatic protection switching
- Selectable line codes (E1: HDB3, AMI/T1: B8ZS, AMI with ZCS)
- Loss of signal indication with programmable thresholds according to ITU-T G.775, ETS300233 (E1) and ANSI T1.403 (T1/J1)
- Optional data stream muting upon LOS detection
- Programmable receive slicer threshold
- Clock generator for jitter free system/transmit clocks per channel
- Local loop and remote loop for diagnostic purposes
- Low power device, single power supply: 3.3 V with 5 V tolerant digital inputs

Frame Aligner Features

- Frame alignment/synthesis for 2048 kbit/s according to ITU-T G.704 (E1) and for1544 kbit/s according to ITU-T G.704 and JT G.704 (T1/J1)
- Programmable frame formats: E1: Doubleframe, CRC Multiframe (E1)T1: 4-Frame Multiframe (F4,FT), 12-Frame Multiframe (F12, D3/4), Extended Superframe (F24, ESF), Remote Switch Mode (F72, SLC96)
- Selectable conditions for recover/loss of frame alignment
- CRC4 to Non-CRC4 Interworking according to ITU-T G. 706 Annex B (E1)
- Error checking via CRC4 procedures according to ITU-T G. 706 (E1)
- Error checking via CRC6 procedures according to ITU-T G. 706 and JT G.706 (T1/J1)
- Performs synchronization in ESF format according to NTT requirements (J1)
- Alarm and performance monitoring per second: 16 bit counter for CRC-, framing errors, code violations, error monitoring via E bit and SA6 bit (E1), errored blocks, PRBS bit errors
- Remote Alarm generation/checking according to ITU JT-G.704 in ESF-format (J1)
- IDLE code insertion for selectable channels
- Single bit defect insertion
- Flexible system clock frequency for receiver and transmitter
- Supports programmable system data rates with independent receive/transmit shifts: E1: 2.048, 4.096, 8.192 and 16.384 Mb/s (according to H.100 bus). T1/J1: 2.048, 4.096, 8.192, 16.384 Mb/s and 1.544, 3.088, 6.176, 12.352 Mb/s
- System interface multiplex mode; multiplexing of four channels into an 8.192 Mb/s data stream and vice versa, bit- or byte-interleaved
- Elastic store for receive and transmit route clock wander and jitter compensation; controlled slip capability and slip indication
- Programmable elastic buffer size: 2 frames/1 frame/short buffer/bypass

- Provides different time slot mapping modes
- Supports fractional E1 or T1/J1 access
- Flexible transparent modes
- Programmable In-Band Loop Code detection and generation (TR62411)
- Channel loop back, line loop back or payload loop back capabilities (TR54016)
- Pseudo Random Bit Sequence (PRBS) generator and monitor (framed or unframed)
- Clear channel capabilities (T1/J1)
- Loop-timed mode

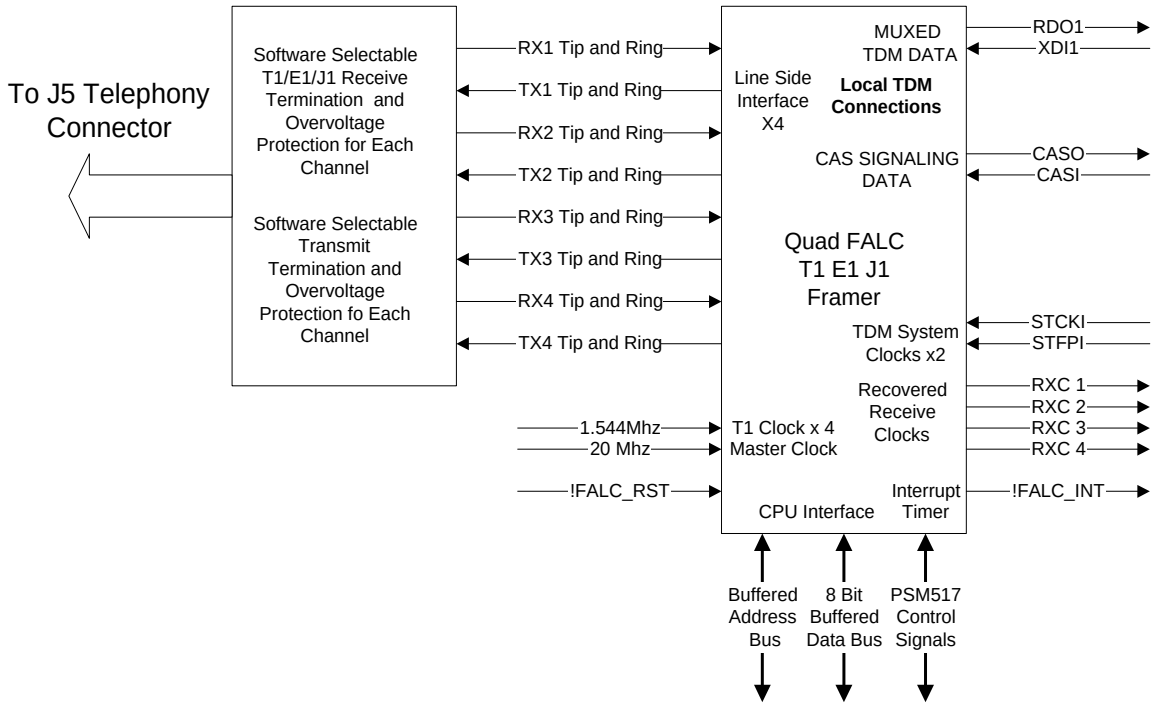
Signaling Controller Features

- HDLC controller Bit stuffing, CRC check and generation, flag generation, flag and address recognition, handling of bit oriented functions
- Supports Signaling System #7 delimitation, alignment and error detection according to ITU-Q.703 processing of fill in signaling units, processing of errored signaling units
- CAS/CAS-BR controller with last look capability, enhanced CAS- register access and freeze signaling indication
- DL-channel protocol for ESF format according to ANSI T1.403 specification or according to AT&T TR54016 (T1/J1)
- DL-bit access for F72 (SLC96) format (T1/J1)
- Generates periodic performance report according to ANSI T1. 403
- Provides access to serial signaling data streams
- Multiframe synchronization and synthesis according to ITU-T G.732
- Alarm insertion and detection (AIS and LOS in time slot 16)
- Transparent Mode
- FIFO buffers (64 bytes deep) for efficient transfer of data packets
- Time slot assignment: Any combination of time slots selectable for data transfer independent of signaling mode (useful for fractional T1/J1 applications)
- Time-slot0S a 8...4-bit handling via FIFOs (E1)
- HDLC access to any S a -bit combination (E1)

General Features

- Extended interrupt capabilities
- One-second timer (internal or external timing reference)

Figure 8-1: Quad Framer Block Diagram



The FALC is accessed by the MPC8260 for setup and control purposes through the Buffered Data bus. The Buffered data bus supplies a buffered MPC8260 address and a bidirectional 16-bit data bus. The QuadFALC is set up to be an 8 bit device on the bus. The chip selects are sub-decoded from the CS_INTERFACE signal in the PSM519 PAL logic. The CS_INTERFACE signal is supplied by the MPC8260 on CS5. Table 8-1, “CS5 Address Decode and Sub-decode,” shows the primary chip select address and the PSM519 sub-decodes:

Table 8-1: CS5 Address Decode and Sub-decode

Address	Device
For Base Address, see “Memory Map,” on page 40	CS_INTERFACE
CS_INTERFACE+0h-CS_INTERFACE+03ffh	QuadFALC 1

The cycle control portion of the buffered data bus for these devices is supplied by the BUF_DS and BUF_RW# signals. The BUF_DS is supplied by the PSM517 PAL logic and acts as the device data strobe in Motorola signaling mode. The PSM517 also supplies the BUF_RW# strobe, which is the read/write, enable. The cycle timing including the cycle termination is handled by the GCPM in the MPC8260. The settings for the CS5 (CS_INTERFACE) address space and the GPCM are as follows.

MPC8260 Related FALC Settings

Table 8-2, “CS5 Base Register Settings,” shows the CS5 base register settings.

Table 8-2: CS5 Base Register Settings

Field	Recommended Setting	Description
PS	01	8-bit
DECC	00	Data Errors checking disabled
WP	0	Read and write accesses are allowed
MSEL	000	Machine Select GPCM-60x bus
EMEMC	0	Accesses are handled by the MSEL selected controller. No external memory controller allowed in this space.
ATOM	00	The address space is controlled by the memory controller bank and is not used for atomic operations
DR	0	No data pipelining is done.
V	1	This bank is valid

The Option Register Settings follow the form outlined for the GPCM Machine, as shown in Table 8-3, “CS5 Option Register Settings,” below.

Table 8-3: CS5 Option Register Settings

Field	Recommended Setting	Description
BCTLD	0	$\overline{\text{BCTLx}}$ is asserted upon access to the current memory bank.
CSNT	1	$\overline{\text{CS}}$/WE is negated a quarter of a clock early.
ACS	11	$\overline{\text{CS}}$ is output half a clock after the address lines.
SCY	5	Clock Wait states
SETA	0	$\overline{\text{PSDVAL}}$ is generated internally.
TRLX	1	Relaxed timing is generated by the GCPM for this memory region.
EHTR	00	Normal Idle Timing

The address range for the different internal sections of the part is as follows.

Quad Framers and LIUs Register Map

The internal register map for the Quad Framers is quite large and is not reproduced here. The map in its entirety can be found in the *Infineon Technologies ICs for Communications PEB22554 Datasheet*. The detailed register definitions are also found in this document.

Quad Framers and LIUs Physical Configuration

There are several functional sections in each device. Several of the sections have different pinout options for system connection to provide unique modes of operation. The following sections describe how the QuadFALC is configured for use in the PCI384 environment.

Line Interface Receive Configuration

The line receive interface option is set to analog input with a transformer and external line conditioning. The terminated and conditioned analog input signal is received on the RL1_X and RL2_X (X=1-4) for each framer. This mode is selected in the LIM1 register by setting the DRS bit to 0 (LIM1.DRS=0).

Line Interface Transmit Configuration

The line transmit interface option is set to analog output with a transformer and external line conditioning. The terminated and conditioned analog output signal is presented to the system on the XL1_X and XL2_X (X=1-4) for each framer. This mode is selected in the LIM1 register by setting the DRS bit to 0 (LIM1.DRS=0). After RESET the transmit lines are in high impedance mode until FMR0.C1 is set to a 1.

Master Clock Configuration

The master clock input can be from 1.02 to 20 MHz at +/- 32 ppm min. The PCI384 supplies the Master Clock input to the QuadFALC with a 25 ppm, 20 MHz frequency input clock. The internal settings for the clock setup and control (refer to registers GCM1 through 6 in the datasheet) can be determined by the calculations given in the datasheet or by a PC calculator tool available on the Infineon Web site.

External DCO-R Clock SYNC Pin Configuration

The external DCO-R SYNC input pin is not supported and has no connection other then a pullup.

One Second Timer Pin Configuration

The SEC pin can transmit or receive an input pulse on a periodic basis (typically one second) for use as an event timer within the part. This line is available on the QuadFALC and can be source of the signal. The default power up condition is set as a pulled up input.

Line Receive Clock Output Pin Configuration

The framer in the QuadFALC has an output pin that will output the recovered line clock or the synchronized DCO-R signal. These pins are connected the reference clock mux. The output of the mux is used to drive the H.100 switch local reference input. Refer to register PC5.CRP, CMR1.RS1, and GPC1.R1S1 in the datasheet for the appropriate configuration settings.

Local TDM Receive Data Pin Configuration

Each of the received data streams is actually sourced from an internal elastic data store device which rate compensates between the trunk input rate and the local TDM stream. Each of the four elastic stores multiplexes its receive data onto the RDO1 output pin for termination in the connection memory of the H.100 TDM switch. The RDO1 data stream is synchronized to the local TDM timing via the H.100 switch's local clock and frame sync. The local streams are all setup to use an 8.192 Mb/s transfer rate regardless of the line side operating rate. Refer to the datasheet for the proper control registers and their settings to achieve the multiplexed configuration.

Local TDM Receive System Clock Pin Configuration

The framer has an elastic data store on the receive data output. The elastic store is used to compensate between the receiver trunk line rate and the local TDM data rate. In the multiplexed data mode, the clock input pin SCLKR1 is used to supply the elastic data stores with a clock that is synchronized to the local TDM. The clock has a fixed input frequency of 8.192 MHz. The SIC2.SSC2 bits are used to scale the clocks input so that each of the four elastic stores can comply to the 8 Mb/s data rate. Refer to the datasheet for the proper control registers and their settings.

Local TDM Receive Frame Pulse Input Pin Configuration

The framers' receive data output is connected to the local TDM bus. The data output stream is sourced from the respective framer's elastic data store. In order to properly synchronize the operation of elastic data store to the local system timing a Frame Pulse is required. The H.100 switch sources the 8 KHz frame pulse. The signal is input to each of the framers on the Receive Multifunction Port A pin 1. This function is called the $\overline{\text{SYPR}}$ or Synchronous Pulse receive mode for the Receive Multifunction pin 1 on each port. Refer to the datasheet for the proper control registers and their settings.

Receive Channel Associated Signaling

Each of the four framers' Channel Associated Signaling (CAS) is multiplexed onto the RPB1 (programmed as RSIG) pin for connection to the local TDM bus. The CAS is terminated in the H.100 connection memory and from there it can be routed out onto the H.100 bus or terminated in the MPC8260. The CAS is transmitted the local TDM bus at the 8.192 Mb/s transfer rate. Refer to the datasheet for the proper control registers and their settings to achieve the multiplexed configuration on RPBI at the 8.192 MHz rate.

Local TDM Transmit Data Pin Configuration

The transmit data source for all four framers is the H.100 switch’s connection memory. Data for each framer is multiplexed onto one TDM stream by the switch and delivered to the framer at an 8.192 Mb/s rate. The QuadFALC receives the multiplexed data stream on the XDI1 pin. The XDI1 pin is synchronized to the board TDM streams by the H.100's 8.192 MHz clock and 8 KHz frame pulse.

The framer demultiplexes the data internally and routes it to each framers elastic store which rate compensates between the trunk output rate and the local TDM stream. Refer to the QuadFALC and H.100 datasheets for the proper control registers and their settings for the demultiplexing configuration.

Local TDM Transmit Channel Associated Signaling Pin Configuration

The transmit Channel Associated Signaling (CAS) source for all four framers is the H.100 switch's connection memory. CAS data for each framer is multiplexed onto one TDM stream by the switch and delivered to the framer at an 8.192 Mb/s rate. The QuadFALC receives the multiplexed CAS data stream on the XPC1 (programmed as the XSIG) pin. The XPC1 pin is synchronized to the board TDM streams by the H.100s 8.192 MHz clock and 8 KHz frame pulse.

The framer demultiplexes the data internally and routes it to each framer for insertion into the outbound data stream. Refer to the QuadFALC sheet for the proper control registers and their settings for this configuration.

Local TDM Transmit System Clock Pin Configuration

The transmit clock input pin, SCLKX1 is used to supply the QuadFALC with a 8.192 MHz bit clock for data input clocking from the TDM bus. The clock is routed and scaled internally to the QuadFALC, so that each elastic data store gets the clock when its data is valid on the TDM bus. The multiplexing control registers along with the SIC2.SSC2 bits are used configure the device for correct operation. Refer to the QuadFALC datasheet for the proper control registers and their settings.

Local TDM Transmit Frame Pulse Input Pin Configuration

The framers' transmit data input is received by the XDI1 pin and demultiplexed internally by the QuadFALC. The data input stream is typically loaded into the respective framer's elastic data store. In order to properly synchronize the operation of elastic data store to the local system timing a Frame Pulse is required. The H.100 switch sources the 8 KHz frame pulse. The signal is input to each of the framers from the Transmit Multifunction Port XPA1 pin. This function is called the $\overline{\text{SYPX}}$ or Synchronous Pulse transmit mode for the Transmit Multifunction pin. Refer to the H.100 datasheet for the proper control registers and their settings.

Line Side T1 Transmit Clock Source

The framer can be setup to source its transmitter clock from a number of sources. Most of the clock sources are derived from internal sources on the chip. For the user's convenience, a 1.544 MHz clock for T1/J1 operation is supplied on the Transmit Multifunction XPB1..4. Refer to the datasheet for the proper control registers and their settings.

QuadFALC Interrupts

The QuadFALC has an interrupt pin. Registers internal to the part define status and control of the pin. Refer to the datasheet for the proper control registers and their settings. The QuadFALCs interrupt is connected directly to the MPC8260s $\overline{\text{IRQ1}}$ input.

Transmitter Line Termination and Conditioning

The Trunk side transmit lines from the QuadFALC are terminated and conditioned by a series of passive components located on the PCI384 mezzanine I/O board. There is an Isolation transformer that divides the circuit into a local side and a line side for galvanic isolation purposes. The Isolation Transformer is a quad multi core device that handles the receive and transmit isolation of all of the quad framer's channels.

On the local side of the Isolation transformer there is a set of series line termination resistors and overvoltage protection diodes. On the line side of the isolation transformer there is lightning protection for each set of transmit lines. The lightning protection consists of a set of fuses and a semiconductor transient absorption device for each set of transmit lines.

The only adjustments to the circuit are a set of programmable switches for the series termination resistors for each leg of transmit line pair.

Receiver Line Termination and Conditioning

The Trunk side receive lines from the QuadFALC are terminated and conditioned by a series of passive components located on the PCI384 mezzanine I/O board. There is an Isolation transformer that divides the circuit into a local side and a line side for galvanic isolation purposes. The Isolation Transformer is a quad multi core device that handles the receive and transmit isolation of all of the quad framer's channels.

The receiver termination consists of parallel line termination resistors and overvoltage protection diodes on the local side of the Isolation Transformer. On the Line side if the Isolation Transformer there is a set of lightning protection devices consisting of fuses and a semiconductor transient absorption device for each set of receive lines.

The only adjustments to the circuit are a set of programmable switches that set the value for the parallel termination resistor of each receive line pair. These termination values can be programmed to support 100 Ohm T1, 110 Ohm J1, 120 Ohm E1 and 75 Ohm E1. The default configuration value of the resistors is set to 100 Ohms for a T1 termination.

Chapter

9

General Purpose Registers

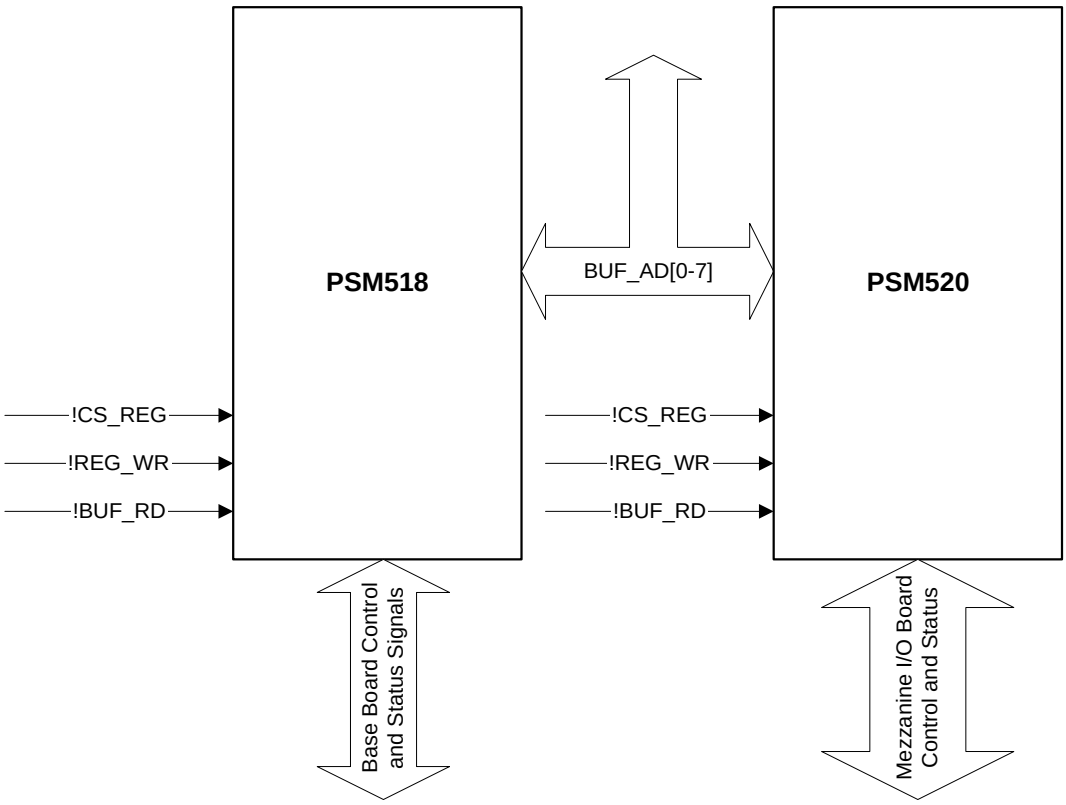
Overview

The General Purpose Control and Status registers control and provide the status of the unique functions of the PCI384 that are not provided by the various peripheral devices on the board. All of the registers are eight bit and are located on the buffered data bus. The PSM518 is located on the base board and the PSM520 is located on the mezzanine I/O board.

Figure 9-1, “Miscellaneous Registers Block Diagram,” on page 84 shows a block diagram for these miscellaneous registers.

Key topics in this chapter include:

Figure 9-1: Miscellaneous Registers Block Diagram



The necessary control and timing for each cycle comes initially from the 60x bus and the BR4 and OR4 GPCM registers. Table 9-1, “CS4 Base Register Settings,” and Table 9-2, “CS4 Option Register Settings,” on page 85 reflect the required settings for this space.

Table 9-1: CS4 Base Register Settings

Field	Recommended Setting	Recommended Setting
PS	01	8-bit
DECC	00	Data Errors checking disabled
WP	0	Read and write accesses are allowed
MSEL	000	Machine Select GPCM-60x bus
EMEMC	0	Accesses are handled by the MSEL selected controller. No external memory controller allowed in this space.
ATOM	00	The address space is controlled by the memory controller bank and is not used for atomic operations
DR	0	No data pipelining is done.
V	1	This bank is valid

The Option Register Settings follow the form outlined for the GPCM Machine, as shown in Table 9-2, “CS4 Option Register Settings,” below.

Table 9-2: CS4 Option Register Settings

Field	Recommended Setting	Description
BCTLD	0	BCTLx is asserted upon access to the current memory bank
CSNT	1	CS/WE is negated a quarter of a clock early.
ACS	11	CS is output half a clock after the address lines.
SCY	3	Clock Wait states
SETA	0	PSDVAL is generated by the GCPM.
TRLX	1	Relaxed timing is generated by the GCPM for this memory region.
EHTR	00	Normal Idle Timing

Once the GCPM initiates the cycle on the 60x bus, the logic in the PSM517 pal then takes the 60x bus input and provides the BUF_ALE, BUF_RD and REG_WR signals. The logic in the register devices, all use the multiplexed address feature of the buffered data bus for address decode. The buffered data bus provides the requested address on the data bits during the first portion of an access cycle and the address is latched and decoded on the falling edge of the BUF_ALE signal. The PSM517 logic then configures the buffered bus to either drive write data or receive read data from the register PAL. The BUF_RD and REG_WR strobes are supplied as required during the data phase of the cycle. Data is written into the registers on the rising edge of REG_WR. BUF_RD is essentially asserted during the entire data phase of the read cycle causing the data from the registers to be driven onto the data lines as long as the strobe is true. Read and Write cycles are completed by the GCPMs wait state timing. There is no cycle complete signal generated by the register logic.

The logic for the registers is contained the PSM518 PAL and PSM520 PAL devices. All of the registers are mapped into the 2000_0000H to 2000_00FF memory space on the MPC8260 chip select 4. The PCI384 select signal is CS_REG. The explicit address of each register and its associated function are contained in the following sections.

PSM518 and PSM520 General Purpose Registers

The registers contained in the PSM518 and PSM520 General Purpose PAL logic exist in address space beginning at offset 0000_004xH to 0000_005xH in the CS_REG space. These registers are shown in Table 9-3, “PSM518 and PSM520 General Purpose Registers,” below.

Table 9-3: PSM518 and PSM520 General Purpose Registers

Register Name	Address Offset	Page Number
LREF0 and LREF1 Select Register	40H	page 87
Clock Control A and B and Netref Control Register	41H	page 88
Peripheral Reset Control Register	42H	page 89
H.100 Peripheral Interrupt Enable Control Register	43H	page 90
H.100 Interrupt Status Register	44H	page 91
Miscellaneous Status and Control Register	46H	page 92
Clock Timer Timeout Period Register	47H	page 93
General Purpose Latch Register	48H	page 94
General Purpose Switch and P1 Reset Register	49H	page 94
LREF3 Select Register	4AH	page 95
Board ID Register	4CH	page 96
Board Option and Revision Register	4DH	page 96
General Purpose Switch and H.100 Termination	4EH	page 97
H.100 Master/Slave and Clock Select	4FH	page 98
Board Status LED Control Register	50H	page 99
General Purpose Switch	5DH	page 100
Receiver Termination Control Register	5EH	page 101
Transmitter Termination Control Register	5FH	page 102

Note: *In the following tables, RSVD means Reserved and RO means Read Only.*

LREF0 and LREF1 Select Register

The LREF0 and LREF1 Select Register, controls the selection of the LREF0 and LREF1 clock input for the H.100 TDM switch. These clock inputs can used as the primary and secondary reference for the switch's internal DPLL. The DPLL supplies all of the timing for the local TDM bus. This logic resides in the PSM521 PAL. The outputted clocks are H100_LREF0 and H100_LREF1.

See Table 9-4 and Table 9-5.

Table 9-4: LREF0 and LREF1 Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	lref1_sel [3:0]				Lref0_sel[3:0]			
Reset Value	00H							
R/W	R/W							
Address	Base+40H							

Table 9-5: LREF0 and LREF1 Register Bit Field Definitions

Bits	Name	Description
7-4	lref1_sel	Setting these bits to the value indicated, selects the indicated reference clock output on the H100_LREF1 output line. The field is mapped as follows: 0 QuadFALC1 receive clock 1 1 QuadFALC1 receive clock 2 2 QuadFALC1 receive clock 3 3 QuadFALC1 receive clock 4 4-A Reserved B External LVDS Clock Source 1 C External LVDS Clock Source 2 D-F Reserved
3-0	lref0_sel	Setting these bits to the value indicated, selects the indicated reference clock output on the H100_LREF0 output line. The field is mapped as follows: 0 QuadFALC1 receive clock 1 1 QuadFALC1 receive clock 2 2 QuadFALC1 receive clock 3 3 QuadFALC1 receive clock 4 4-A Reserved B External LVDS Clock Source 1 C External LVDS Clock Source 2 D-F Reserved

Clock Control A and B and Netref Control Register

This register controls three sets of functions. The first is the A and B CTbus clock bypass device and the second is the output enables for the CTbus netref signals. The third function is the control of the CTbus Clock and Frame Pulse termination. The CTbus termination is enabled when the PCI384 is the last board on either end of the CTbus. The logic resides in the PSM521 Pal and the signals outputted are CLK_CNTL_A, CLK_CNTL_B, CREF1_OE, CREF2_OE, H.100 frame and clocks and FALC frame and clocks.

See Table 9-6 and Table 9-6.

Table 9-6: Clock Control A and B and Netref Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RESERVED				crefen2	crefen1	clkcntb	clkcnta
Reset Value	0				0	0	1	1
R/W	READ ONLY					R/W		
Address	Base+41H							

Table 9-7: Clock Control A and B and Netref Control Register Bit Definitions

Bits	Name	Description
0	clkcnta	When set to a 0, the CTbus A Clock Series Termination resistor is bypassed allowing full drive on the CTbus. This mode is used when the local H.100 switch is the A Clock master on the CTbus. When set to 1 the series Termination resistor is in the circuit. This mode is used when the local H.100 chip is an A Clock Slave.
1	clkcntb	When set to a 0, the CTbus B Clock Series Termination resistor is bypassed allowing full drive on the CTbus. This mode is used when the local H.100 switch is the B Clock master on the CTbus. When set to 1 the series Termination resistor is in the circuit. This mode is used when the local H.100 chip is an B Clock Slave
2	crefen1	When set to a 1, this bit will enable the Zarlink MT90866s CT Netreference 1 signal to be driven onto the CTbus. When set to a 0 the CT Netreference 1 signals drive is tri-stated.
3	crefen2	Reserved
4-7	RSVD	Reserved

Peripheral Reset Control Register

The Peripheral Reset Control Register reflects the current state of all of the peripheral chip resets as well as allowing the user to enable/reset the chip under program control.

See Table 9-8 and Table 9-9.

Table 9-8: Peripheral Reset Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	qf1rst	h100rst
Reset Value	1	1	0	0	0	0	0	0
R/W	READ ONLY						R/W	
Address	Base+42H							

Table 9-9: Peripheral Reset Control Register Bit Field Definitions

Bits	Name	Description
0	h100rst	When this bit is set to a 0, the H.100 switch chip is held in reset. The bit will always be set to 0 during a PQ_SRESET. Setting the bit to a 1 will release the H.100 switch chip from the reset state.
1	qf1rst	When this bit is set to a 0, the QuadFALC 1 chip is held in reset. The bit will always be set to 0 during a PQ_SRESET. Setting the bit to a 1 will release the QuadFALC 1 chip from the reset state
2-7	RSVD	Reserved

H.100 Peripheral Interrupt Enable Control Register

The H.100 Peripheral Interrupt Enable control register is used to enable one or more of the TDM peripheral devices as an interrupting source. The H.100 interrupt is mapped to the MPC8260 IRQ5. The individual interrupt sources and their enables are listed in the sections following tables.

See Table 9-10 and Table 9-11.

Table 9-10: H.100 Peripheral Interrupt Enable Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RESERVED				lref1enb	lref0enb	failbenb	failaenb
Reset Value	0				0	0	0	0
R/W	READ ONLY				R/W			
Address	Base+43H							

Table 9-11: H.100 Peripheral Interrupt Enable Control Register Bit Field Definitions

Bits	Name	Description
0	failaenb	When the failaenb bit is set to a 1, the A side CT clock fail interrupt is enabled. A 0 in this bit position will disable the condition as an interrupting source. A PQ_SRESET will cause the bit to be disabled.
1	failbenb	When the failbenb bit is set to a 1, the B side CT clock fail interrupt is enabled. A 0 in this bit position will disable the condition as an interrupting source. A PQ_SRESET will cause the bit to be disabled.
2	lref0enb	When the lref0enb bit is set to a 1, the local reference 0 clock fail interrupt is enabled. A 0 in this bit position will disable the condition as an interrupting source. A PQ_SRESET will cause the bit to be disabled.
3	lref1enb	When the lref1enb bit is set to a 1, the local reference 0 clock fail interrupt is enabled. A 0 in this bit position will disable the condition as an interrupting source. A PQ_SRESET will cause the bit to be disabled.

H.100 Interrupt Status Register

The H.100 Interrupt Status Register reflects the current state of the different H.100 interrupt sources. When one or more of these sources is enabled in the H.100 Interrupt Enable Control Register, the source bit(s) will create an interrupt on $\overline{\text{IRQ5}}$ on the MPC8260. The register is also used to clear a latched bit when the interrupt source has been cleared. Their bit definitions are reflected in the following tables.

See Table 9-12 and Table 9-13.

Table 9-12: H.100 Interrupt Status Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RESERVED				lref1sts	lref0sts	failbsts	failasts
Reset Value	0				Current Value			
R/W	READ ONLY				R/W			
Address	Base+44H							

Table 9-13: H.100 Interrupt Status Register Bit Field Definitions

Bits	Name	Description
0	failasts	If the Fail A output from the H.100 switch chip is enabled, this bit is set to a 1 when the CTbus A clock has failed. It is set to a 0 if the CTbus A clock is normal. If the Fail A output from the H.100 switch chip is disabled, read-back of this bit is undefined and should be disabled as an interrupt source. If a 1 is written to this bit and the interrupt source is cleared, the status bit will be cleared. Writing a 0 to this bit will have no effect.
1	failbsts	If the Fail B output from the H.100 switch chip is enabled, this bit is set to a 1 when the CTbus B clock has failed. It is set to a 0 if the CTbus B clock is normal. If the Fail B output from the H.100 switch chip is disabled, read-back of this bit is undefined and should be disabled as an interrupt source. If a 1 is written to this bit and the interrupt source is cleared, the status bit will be cleared. Writing a 0 to this bit will have no effect.
2	lref0sts	If the clock reference control logic is initialized and enabled, this bit is set to a 1 indicates that the local reference 0 clock source has failed. A 0 on this bit indicates that the reference input clock is functional. If the clock reference control logic is un-initialized or disabled, read-back of this bit is invalid. If a 1 is written to this bit and the interrupt source is cleared, the status bit will be cleared. Writing a 0 to this bit will have no effect.
3	lref1sts	If the clock reference control logic is initialized and enabled, this bit is set to a 1 indicates that the local reference 1 clock source has failed. A 0 on this bit indicates that the reference input clock is functional. If the clock reference control logic is un-initialized or disabled, read-back of this bit is invalid. If a 1 is written to this bit and the interrupt source is cleared, the status bit will be cleared. Writing a 0 to this bit will have no effect.

Miscellaneous Status and Control Register

The Miscellaneous Status and Control Register contains a number of functions that deal with various unrelated parts of the board. The register contains status bits for the factory defaults jumper, the control bit for the H.100 switch chip's output enable and the flash write protect bit. The bit definitions are as follows.

See Table 9-14 and Table 9-15.

Table 9-14: Miscellaneous Status and Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	factdflt	RSVD	RSVD	RSVD	RSVD	h100ode	flashwp	RSVD
Reset Value	Current Value	1	1	1	1	0	0	1
R/W	READ ONLY					R/W		RO
Address	Base+46H							

Table 9-15: Miscellaneous Status and Control Register Bit Field Definitions

Bits	Name	Description
0	RSVD	Reserved always set to a 1.
1	flashwp	When set to a 0, the flashwp bit will prevent a write cycle to the Application or Boot Flash devices from being effected. The write cycle will terminate normally for the MPC8260, but the addressed flash device will not have its write strobe toggled. When this bit is set to 1, write operations will be allowed to the flash device.
2	h100ode	This bit controls the Output Drive Enable to the H.100 switch chip. When the bit is set to 0, the STo0 to Sto27 serial outputs and STio0 to STio31 serial bidirectional outputs on the chip are tri-stated. The C1M5o, 1.544 MHz clockout, C32o, 32 MHz clockout, ST_Cko, 8 MHz TDM clock, and ST_Fpo 8 KHz TDM Frame Pulse are tri-stated as well. When set to a 1, all of these signals are enabled to drive their respective busses.
3	RSVD	Reserved, always set to a 1.
4	RSVD	Reserved, always set to a 1.
5	RSVD	Reserved, always set to a 1.
6	RSVD	Reserved, always set to a 1.
7	factdflt	This bit reflects that state of the P3-7 to P3-8 jumper. When the jumper is connected, the bit will indicate a 0. This indicates that user wants the PTI default software configuration loaded into all devices at bootup. When the P3-7 to P3-8 jumper is removed, the bit indicates a 1. This setting means that the customer settings will be loaded at bootup.

Clock Timer Timeout Period Register

The Clock Timer Timeout Period Register contains a clock count value that is used to check for a valid time period for the local reference clocks fed from the PSM521 select multiplexer to the H.100 switch chip. The timer value is the maximum number of 20 MHz clocks that are allowed to elapse during the reference clock period. If the 20 MHz count is exceeded without seeing at least one valid local reference clock the lrefxsts bit will be set which causes an MPC8260 interrupt on $\overline{\text{IRQ5}}$. The appropriate lrefxmsk bit must be unmasked in order for the interrupt to be generated.

See Table 9-16 and Table 9-17.

Table 9-16: Clock Timer Timeout Period Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	timer[7:0]							
Reset Value	20H							
R/W	R/W							
Address	Base+47H							

Table 9-17: Clock Timer Timeout Register Bit Field Definitions

Bits	Name	Description
7-0	timer[7:0]	This eight-bit field represents the count value for the reference clock timeout timer. The default initial count value is 20H 20 MHz clocks or 1.6 us.

General Purpose Latch Register

The General Purpose Latch Register is an eight-bit Read/Write storage location for any general purpose use. The location is volatile.

See Table 9-18 and Table 9-19.

Table 9-18: General Purpose Latch Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	latch[7:0]							
Reset Value	0							
R/W	R/W							
Address	Base+48H							

Table 9-19: General Purpose Latch Register Bit Field Definitions

Bits	Name	Description
7:0	latch[7:0]	This register is a general purpose, eight-bit storage latch. This location is volatile.

General Purpose Switch and P1 Reset Register

The General Purpose Switch and P1 Reset Register is an eight-bit, read-only register. It contains the bit values of the four-bit general purpose switch (SW1) and the current status of the Compact PCI reset bit.

See Table 9-20 and Table 9-21.

Table 9-20: General Purpose Switch and P1 Reset Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	p1rst	RESERVED			gpsw[4:1]			
Reset Value	Current Value	0			Current Value			
R/W	READ ONLY							
Address	Base+49H							

Table 9-21: General Purpose Switch and P1 Reset Register Bit Field Definitions

Bits	Name	Description
3:0	gpsw[4:1]	Each bit of this field represents the position of one of the four switches of SW1. Bit 3 represents switch 1, bit 2 switch 2, bit 1 switch 3 and bit 0 switch 4. A 1 represents an open switch condition. A 0 represents a closed switch position.
7	p1rst	This bit represents the current state of the CompactPCI reset signal. A 0 indicates that the CompactPCI bus is reset. A 1 indicates that the CompactPCI bus is activated.

LREF3 Select Register

The LREF3 Select Register, controls the selection of the LREF3 clock input for the H.100 TDM switch. This clock input is intended to be used as the source for the CTbus Netreference signal. The H.100 TDM switch can connect its LREF3 input internally to the CTbus Netref signal and drive it. This logic resides in the PSM521 PAL. The outputted clock is H100_LREF3.

See Table 9-22 and Table 9-23.

Table 9-22: LREF3 Select Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	Lref3_sel[3:0]				RESERVED			
Reset Value	00H							
R/W	R/W				READ ONLY			
Address	Base+4AH							

Table 9-23: LREF3 Select Register Bit Field Definitions

Bits	Name	Description
7-4	lref1_sel	Setting these bits to the value indicated, selects the indicated reference clock output on the H100_LREF1 output line. The field is mapped as follows: 0 QuadFALC1 receive clock 1 1 QuadFALC1 receive clock 2 2 QuadFALC1 receive clock 3 3 QuadFALC1 receive clock 4 4-A Reserved B External LVDS Clock Source 1 C External LVDS Clock Source 2 D-F Reserved
3-0	RSVD	Reserved, always set to 0.

Board ID Register

The Board ID Register reflects the current board ID value. It is currently set to 03h.

See Table 9-24 and Table 9-25.

Table 9-24: Board ID Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	brdid[7:0]							
Reset Value	3							
R/W	READ ONLY							
Address	Base+4CH							

Table 9-25: Board ID Register Bit Field Definitions

Bits	Name	Description
7:0	brdid[7:0]	0 h Initial Board revision

Board Option and Revision Register

The Board Option and Revision Register reflects the current Board Options and the Hardware Revision. It is currently set to 00h.

See Table 9-26 and Table 9-27.

Table 9-26: Board ID Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	Optioned				Revid			
Reset Value	Optioned				Revid			
R/W	READ ONLY							
Address	Base+4DH							

Table 9-27: Board ID Register Bit Field Definitions

Bits	Name	Description
7:4	Optioned	0 - Installed Options 0 means no options.
3:0	Revid	X - Board Hardware version revision. Initial value 0.

General Purpose Switch and H.100 Termination Register

The General Purpose Switch and H.100 Termination Register is a four-bit, read-only register and one-bit read/write. It contains the bit values of the four-bit general purpose switch (SW1) and the current status of the H.100 Termination State.

See Table 9-28 and Table 9-29.

Table 9-28: General Purpose Switch and P1 Reset Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	Gpsw[8:5]				Reserved			Term
Reset Value	Current Value				0			0
R/W	READ ONLY							R/W
Address	Base+4EH							

Table 9-29: General Purpose Switch and P1 Reset Register Bit Field Definitions

Bits	Name	Description
7:4	gpsw[8:5]	Each bit of this field represents the position of one of the four switches of SW1. Bit 7 represents switch 5, bit 6 switch 6, bit 5 switch 7 and bit 4 switch 8. A 1 represents an open switch condition. A 0 represents a closed switch position.
0	Term	When this bit is set to a 0, the CTbus termination is disabled. When this bit is set to a 1, the CTbus termination is enabled. The termination is only enabled when the PCI384 is the last device on either end of the CTbus.

H.100 Digital Switch Clock Control

Version 2.0 of the board includes circuitry to supply external bypass clocks due to a timing issue of the present version of the digital switch. This register controls the selection of Slave mode where the external clock is used or Master which it is not. Also a bit selects which clock on the TDM bus is presented to the PLL.

See Table 9-30 and Table 9-31.

Table 9-30: Zarlink Patch Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	A_B	M_S
Reset Value	0	0	0	0	0	0	0	0
R/W	RO	RO	RO	RO	RO	RO	R/W	R/W
Address	Base+4FH							

Table 9-31: Zarlink Patch Control Register Bit Field Definitions

Bits	Name	Description
0	M_S	A 0 in this bit selects the Master mode where the Zarlink switch uses the standard reference clock. A 1 in this bit selects the Slave mode where the Zarlink switch uses the 64 MHz Bypass clock.
1	A_B	A 0 in this bit position selects the A side of the 8 MHz reference clock to the PLL. A 1 in this bit position selects the B side of the 8 MHz clock.
2-7	RSVD	Reserved, this bit is always set to 0.

Board Status LED Control Register

The Board Status LED Control Register controls the Green/Yellow Board status LED. The register controls whether the LED is on or off and what color the bicolor LED will show. Note that the default state on a PQ_SRESET is the LED off.

See Table 9-32 and Table 9-33.

Table 9-32: Board Status LED Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RSVD	RSVD	stsgrn	RSVD	RSVD	RSVD	stsgoff	RSVD
Reset Value	0	0	0	0	1	0	1	0
R/W	RO	RO	R/W	RO	RO	RO	R/W	RO
Address	Base+50H							

Table 9-33: Board Status LED Control Register Bit Definitions

Bits	Name	Description
0	RSVD	Reserved, this bit is always set to 0.
1	Stsgoff	A 1 in this bit position turns off the green General Board Status LED. This bit setting will turn off the LED regardless of the status of the stsgrn bit. A 0 in this bit position will enable the stsgrn bit to turn on the LED and show one of the two LED colors available to the LED.
2	RSVD	Reserved, this bit is always set to 0.
3	RSVD	Reserved, this bit is always set to 1.
4	RSVD	Reserved, this bit is always set to 0.
5	Stsgrn	With the stsgoff bit set to a 0, a 0 in this bit position will turn the green General Board Status led to green and a 1 in this bit position will set the LED to yellow.
6-7	RSVD	Reserved, these bits are always set to 0.

General Purpose Switch

The General Purpose Switch is an eight-bit, read-only register. It contains the bit values of the eight-bit general purpose switch (SW1) switch positions 8-1.

See Table 9-34 and Table 9-35.

Table 9-34: General Purpose Switch and P1 Reset Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	gpsw[8:1]							
Reset Value	Switch Setting							
R/W	READ ONLY							
Address	Base+5DH							

Table 9-35: General Purpose Switch and P1 Reset Register Bit Field Definitions

Bits	Name	Description
7:0	gpsw[8:1]	Each bit of this field represents the position of one of the four switches of SW1. Bit 7 is switch 8, bit 6 switch 7, bit 5 switch 6, bit 4 switch 5, bit 3 switch 4, bit 2 switch 3, bit 1 switch 2 and bit 0 switch 1. A 1 represents an open switch condition. A 0 represents a closed switch position.

Receiver Termination Control Register

The Receiver Termination Control Register sets the receive line termination resistor values for the QuadFALC Line Interface Units. The bits in the register control a set of analog switches that connect the termination resistors in and out of the receiver circuit. The control function groups port 0 and port 1 into a set of termination choices and ports 2 and 3 into another set of termination choices. The tables below outline the various configuration possibilities.

See Table 9-36 and Table 9-37.

Table 9-36: Receiver Termination Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RSVD	RSVD	23rterm1	23rterm0	RSVD	RSVD	01rterm1	01rterm0
Reset Value	0	0	0	0	0	0	0	0
R/W	RO	RO	R/W	R/W	RO	RO	R/W	RW
Address	Base+5EH							

Table 9-37: Receiver Termination Control Register Bit Field Definitions

Bits	Name	Description		
0-1	01rterm0, 01rterm1	01rterm0, 01rterm1	Termination Value	Set these bits to match the T1, J1, E1 interface being used.
		00	100 Ohm T1	
		10	110 Ohm E1	
		01	120 Ohm E1	
		11	75 Ohm J1	
2-3	RSVD	Reserved, this bit is always set to 0.		
4-5	23rterm0, 23rterm1	23rterm0, 23rterm1	Termination Value	Set these bits to match the T1, J1, E1 interface being used.
		00	100 Ohm T1	
		10	110 Ohm E1	
		01	120 Ohm E1	
		11	75 Ohm J1	
6-7	RSVD	Reserved, this bit is always set to 1.		

Transmitter Termination Control Register

The Transmitter Termination Control Register sets the transmit line return loss resistor values for the QuadFALC Line Interface Units. The bits in the register control a set of analog switches that connect the termination resistors in and out of the transmitter circuit. The control function groups port 0 and port 1 into a set of termination choices and ports 2 and 3 into another set of termination choices. The tables below outline the various configuration possibilities.

See Table 9-38 and Table 9-39.

Table 9-38: Transmitter Termination Control Register Bit Fields

MPC8260 Bit	0(MSB)	1	2	3	4	5	6	7(LSB)
Buffered Bus Bit	7(MSB)	6	5	4	3	2	1	0(LSB)
Field	RSVD	RSVD	23tterm1	23tterm0	RSVD	RSVD	01tterm1	01tterm0
Reset Value	0	0	0	0	0	0	0	0
R/W	RO	RO	R/W	R/W	RO	RO	R/W	R/W
Address	Base+5FH							

Table 9-39: Transmitter Termination Control Register Bit Field Definition

Bits	Name	Description		
0-1	01tterm0, 01tterm1	01tterm0, 01tterm1	Termination Value	Set these bits to match the T1, J1, E1 interface being used.
		00	For 100 Ohm T1, 110 Ohm J1, 120 Ohm E1 return loss	
		11	75 Ohm E1 return loss	
2-3	RSVD	Reserved, this bit is always set to 0.		
4-5	23tterm0, 23tterm1	23tterm0, 23tterm1	Termination Value	Set these bits to match the T1, J1, E1 interface being used.
		00	For 100 Ohm T1, 110 Ohm J1, 120 Ohm E1 return loss	
		01	75 Ohm E1 return loss	
6-7	RSVD	Reserved, this bit is always set to 1.		

Chapter

10

Pinouts

Overview

The following tables present the pin assignments for the PCI384.

Key topics in this chapter include:

PCI Pin Assignments

In Table 10-1, “Base P1 PCI Connector,” '#' denotes an active low signal. Signals in braces {} reflect the *PCI Local Bus Specification Revision 2.2* signals that are not connected on the PCI384.

Table 10-1: Base P1 PCI Connector

Pin Number	Side B Signal Name	Side A Signal Name
1	nc {-12 V}	TRST#
2	nc {TCK}	nc {+12 V}
3	GND	TMS
4	TDO	TDI
5	+5 V	+5 V
6	+5 V	INTA#
7	nc {INTB#}	nc {INTC#}
8	nc {INTD#}	+5 V
9	PRSNT1#	reserved
10	reserved	nc {+5 V (I/O)}
11	PRSNT2#	reserved
12	Connector Keyway	Connector Keyway
13	Connector Keyway	Connector Keyway
14	reserved	reserved {3.3 Vaux}
15	GND	RST#
16	CLK	nc {+5 V (I/O)}
17	GND	GNT#
18	REQ#	GND
19	nc {+5 V (I/O)}	reserved {PME#}
20	AD[31]	AD[30]
21	AD[29]	nc {+3.3 V}
22	GND	AD[28]
23	AD[27]	AD[26]
24	AD[25]	GND
25	nc {+3.3 V}	AD[24]
26	C/BE[3]#	IDSEL
27	AD[23]	nc {+3.3 V}
28	GND	AD[22]
29	AD[21]	AD[20]
30	AD[19]	GND
31	nc {+3.3 V}	AD[18]
32	AD[17]	AD[16]
33	C/BE[2]#	nc {+3.3 V}
34	GND	FRAME#
35	IRDY#	GND

Table 10-1: Base P1 PCI Connector (Continued)

Pin Number	Side B Signal Name	Side A Signal Name
36	nc {+3.3 V}	TRDY#
37	DEVSEL#	GND
38	GND	STOP#
39	nc {LOCK#}	nc {+3.3 V}
40	PERR#	nc {SDONE}
41	nc {+3.3 V}	nc {SBO#}
42	SERR#	GND
43	nc {+3.3 V}	PAR
44	C/BE[1]#	AD[15]
45	AD[14]	nc {+3.3 V}
46	GND	AD[13]
47	AD[12]	AD[11]
48	AD[10]	GND
49	M66EN {GND}	AD[09]
50	Connector Keyway	Connector Keyway
51	Connector Keyway	Connector Keyway
52	AD[08]	C/BE[0]#
53	AD[07]	nc {+3.3 V}
54	nc {+3.3 V}	AD[06]
55	AD[05]	AD[04]
56	AD[03]	GND
57	GND	AD[02]
58	AD[01]	AD[00]
59	nc {+5 V (I/O)}	nc {+5 V (I/O)}
60	ACK64#	REQ64#
61	+5 V	+5 V
62	+5 V	+5 V
	Connector Keyway	Connector Keyway
	Connector Keyway	Connector Keyway
63	reserved	GND
64	GND	C/BE[7]#
65	C/BE[6]#	C/BE[5]#
66	C/BE[4]#	nc {+5 V (I/O)}
67	GND	PAR64
68	AD[63]	AD[62]
69	AD[61]	GND
70	nc {+5 V (I/O)}	AD[60]
71	AD[59]	AD[58]
72	AD[57]	GND
73	GND	AD[56]
74	AD[55]	AD[54]

Table 10-1: Base P1 PCI Connector (Continued)

Pin Number	Side B Signal Name	Side A Signal Name
75	AD[53]	nc {+5 V (I/O)}
76	GND	AD[52]
77	AD[51]	AD[50]
78	AD[49]	GND
79	nc {+5 V (I/O)}	AD[48]
80	AD[47]	AD[46]
81	AD[45]	GND
82	GND	AD[44]
83	AD[43]	AD[42]
84	AD[41]	nc {+5 V (I/O)}
85	GND	AD[40]
86	AD[39]	AD[38]
87	AD[37]	GND
88	nc {+5 V (I/O)}	AD[36]
89	AD[35]	AD[34]
90	AD[33]	GND
91	GND	AD[32]
92	reserved	reserved
93	reserved	GND
94	GND	reserved

Modular Jack Pin Assignments

The PCI384 is connected to the media using a shielded four position RJ48C connector on the mounting bracket. The mating plug (not supplied) may be a shielded or unshielded 8 position modular jack. Where maximum noise immunity is desired the shielded version should be used.

The pin assignments for this connector are shown Table 10-2, “Mezzanine J1 Telecom Line Interface,” below.

Table 10-2: Mezzanine J1 Telecom Line Interface

Pin Number	Signal Name
A1	RRING_4
A2	RTIP_4
A3	nc
A4	TRING_4
A5	TTIP_4
A6	nc
A7	nc
A8	nc
B1	RRING_3
B2	RTIP_3
B3	nc
B4	TRING_3
B5	TTIP_3
B6	nc
B7	nc
B8	nc
C1	RRING_2
C2	RTIP_2
C3	nc
C4	TRING_2
C5	TTIP_2
C6	nc
C7	nc
C8	nc
D1	RRING_1
D2	RTIP_1
D3	nc
D4	TRING_1
D5	TTIP_1
D6	nc
D7	nc
D8	nc

Master/Slave PCM Expansion Connector Pin Assignments

Two PCI384 boards may be configured as a Master/Slave pair using the H.100 bus provided at card edge connector P2. Connect the two boards together (in adjacent PCI slots) using ribbon cable assembly using the pin assignments shown in Table 10-3, “Mezzanine P2 Master/Slave PCM Expansion Connector,” below.

Table 10-3: Mezzanine P2 Master/Slave PCM Expansion Connector

Pin Number	Signal Name	Signal Name	Pin Number
1	reserved	reserved	2
3	CT_D31	CT_D30	4
5	CT_D29	CT_D28	6
7	GND	CT_D27	8
9	CT_D26	CT_D25	10
11	CT_D24	GND	12
13	CT_D23	CT_D22	14
15	CT_D21	CT_D20	16
17	GND	CT_D19	18
19	CT_D18	CT_D17	20
21	CT_D16	GND	22
23	CT_D15	CT_D14	24
25	CT_D13	CT_D12	26
27	GND	CT_D11	28
29	CT_D10	CT_D9	30
31	CT_D8	GND	32
33	CT_D7	CT_D6	34
35	CT_D5	CT_D4	36
37	GND	CT_D3	38
39	CT_D2	CT_D1	40
41	CT_D0	GND	42
43	$\overline{\text{CT\_FRAME\_A}}$	GND	44
45	CT_C8_A	GND	46
47	CTNR1	GND	48
49	$\overline{\text{CT\_FRAME\_B}}$	GND	50
51	CT_C8_B	GND	52
53	reserved	GND	54
55	reserved	GND	56
57	reserved	GND	58
59	reserved	GND	60
61	reserved	GND	62
63	reserved	GND	64
65	reserved	reserved	66
67	GND	reserved	68

JTAG/BDM Debug Port

The JTAG testing port on the MPC8260 is used to support the EST Common On chip debug Processor (COP) debugger on the P4 connector. The connector is setup to support the extended 16-pin COP debugger signaling, but the basic ten-pin signaling devices may be used with an interposing adapter. The P4 pinout is found in Table 10-4, “JTAG Pinouts,” below.

Table 10-4: JTAG Pinouts

Pin Number	Signal Name
1	PQ_TDO - JTAG Test Data Out signal
2	$\overline{\text{PQ\_QACK}}$ - Quiescent State Acknowledge, not supported
3	PQ_TDI - JTAG Test Data In signal
4	$\overline{\text{PQ\_TRST}}$ - JTAG Reset and Tri-state signal
5	$\overline{\text{PQ\_QREQ}}$ - Quiescent State Request-
6	V3V
7	PQ_TCK - JTAG Test Clock
8	No Connection
9	PQ_TMS - JTAG Test Mode Select
10	No Connection
11	$\overline{\text{PQ\_SRESET}}$ - MPC8260 Soft Reset
12	Ground
13	$\overline{\text{PQ\_HRESET}}$ - MPC8260 Hard Reset
14	No Connection
15	$\overline{\text{CHKSTPO}}$ - Checkstop output, Not supported
16	Ground

All of the control signals are pulled to V3V with a minimum 10 KB resistor to prevent false actuation when no JTAG controller is connected.

The mechanical requirements 16-pin Motorola connector are specified as follows.

- Vertical, 16 (2 X 8) pin header
- 0.10" between centers of adjacent pins
- 0.025" square pins
- 0.23" height of each post

Analyzer Headers Pinout

The Mictor Analyzer Headers are located on the PCI384 base board. They are typically used to connect a logic analyzer or similar type of device to the PCI384 board for use as engineering debug and diagnostic aids. They are not installed in shippable, standard product.

The pinouts are shown in the following tables:

- Table 10-5, “P5 Analyzer Header Pinout, 60X Bus Address,” on page 110
- Table 10-6, “P8 Analyzer Header, 60X Bus MS Data,” on page 111
- Table 10-7, “P9 Analyzer Header Pinout, 60X Bus LS Data,” on page 111
- Table 10-8, “P7 Mictor Pinout, 60X Bus Control,” on page 112
- Table 10-9, “P6 Mictor Pinout, 60X Bus Miscellaneous Signals,” on page 112.

Table 10-5: P5 Analyzer Header Pinout, 60X Bus Address

Pin Number	Signal Name	Signal Name	Pin Number
1	PQ_A0	PQ_A9	19
2	PQ_A16	PQ_A25	20
3	PQ_A1	PQ_A10	21
4	PQ_A17	PQ_A26	22
5	PQ_A2	PQ_A11	23
6	PQ_A18	PQ_A27	24
7	PQ_A3	PQ_A12	25
8	PQ_A19	PQ_A28	26
9	PQ_A4	PQ_A13	27
10	PQ_A20	PQ_A29	28
11	PQ_A5	PQ_A14	29
12	PQ_A21	PQ_A30	30
13	PQ_A6	PQ_A15	31
14	PQ_A22	PQ_A31	32
15	PQ_A7	GROUND	33
16	PQ_A23	GROUND	34
17	PQ_A8	$\overline{\text{PQ\_TA}}$	35
18	PQ_A24	CLK_1 (66 MHz)	36

Table 10-6: P8 Analyzer Header, 60X Bus MS Data

Pin Number	Signal Name	Signal Name	Pin Number
1	PQ_D0	PQ_D9	19
2	PQ_D16	PQ_D25	20
3	PQ_D1	PQ_D10	21
4	PQ_D17	PQ_D26	22
5	PQ_D2	PQ_D11	23
6	PQ_D18	PQ_D27	24
7	PQ_D3	PQ_D12	25
8	PQ_D19	PQ_D28	26
9	PQ_D4	PQ_D13	27
10	PQ_D20	PQ_D29	28
11	PQ_D5	PQ_D14	29
12	PQ_D21	PQ_D30	30
13	PQ_D6	PQ_D15	31
14	PQ_D22	PQ_D31	32
15	PQ_D7	GROUND	33
16	PQ_D23	GROUND	34
17	PQ_D8	$\overline{\text{PQ\_PSDVAL}}$	35
18	PQ_D24	NC	36

Table 10-7: P9 Analyzer Header Pinout, 60X Bus LS Data

Pin Number	Signal Name	Signal Name	Pin Number
1	PQ_D32	PQ_D41	19
2	PQ_D48	PQ_D57	20
3	PQ_D33	PQ_D42	21
4	PQ_D49	PQ_D58	22
5	PQ_D34	PQ_D43	23
6	PQ_D50	PQ_D59	24
7	PQ_D35	PQ_D44	25
8	PQ_D51	PQ_D60	26
9	PQ_D36	PQ_D45	27
10	PQ_D52	PQ_D261	28
11	PQ_D37	PQ_D46	29
12	PQ_D53	PQ_D62	30
13	PQ_D38	PQ_D47	31
14	PQ_D54	PQ_D63	32
15	PQ_D39	GROUND	33
16	PQ_D55	GROUND	34
17	PQ_D40	$\overline{\text{PQ\_TS}}$	35
18	PQ_D56	NC	36

Table 10-8: P7 Mictor Pinout, 60X Bus Control

Pin Number	Signal Name	Signal Name	Pin Number
1	PQ_TEA	PQ_MODCK3	19
2	PQ_TBST	PQ_TT2	20
3	PQ_ABB	PQ_GBL	21
4	PQ_ARTY	PQ_TT3	22
5	PQ_BG	PQ_L2HIT	23
6	PQ_TS	PQ_TT4	24
7	PQ_PSDVAL	PQ_CPU_BR	25
8	PQ_TSIZ0	PQ_TA	26
9	PQ_CPU_DBG	PQ_HRESET	27
10	PQ_TSIZ1	BADDR31	28
11	PQ_BR	PQ_SRESET	29
12	PQ_TSIZ2	PQ_MODCLK1	30
13	ALE	PORESET	31
14	PQ_TSIZ3	PQ_MODCK2	32
15	PQ_DBG	GROUND	33
16	PQ_TT0	GROUND	34
17	PQ_DBB	PQ_TEA	35
18	PQ_TT1	PA_AACK	36

Table 10-9: P6 Mictor Pinout, 60X Bus Miscellaneous Signals

Pin Number	Signal Name	Signal Name	Pin Number
1	PQ_PGTA	PSDDQM1	19
2	PSDA10	MON_INT	20
3	PSDWE	PSDDQM0	21
4	BUF_ALE	PQ_IRQ5	22
5	CS_INTERFACE	CS_REG	23
6	BUF_DS	PQ_IRQ4	24
7	PSDDQM7	CS_FLASHA	25
8	BUF_RD	PQ_IRQ3	26
9	PSDDQM6	CS_LOC_RAM	27
10	BUF_WR	QFALC2_INT	28
11	PSDDQM5	CS_SDRAM	29
12	BUF_RW#	QFALC1_INT	30
13	PSDDQM4	CS_FLASHB	31
14	TOD_INT	PQ_NMI	32
15	PSDDQM3	GROUND	33
16	SMB_INT	GROUND	34
17	PSDDQM2	PQ_PGTA	35
18	PQ_IRQ7	NC	36

P4 Nine-Pin D Subminiature Connector Pinout

The P4 nine-pin D Subminiature connector is used to carry the console port and the two external LVDS clock inputs. The connector is located on the mezzanine I/O board. The pinout is defined in Table 10-10, “P4 Nine-Pin D Subminiature Pinout,” below.

Table 10-10: P4 Nine-Pin D Subminiature Pinout

Pin Number	Signal Name
1	NC- RESERVED for Chassis Shield
2	Received RS232 Data
3	Transmit RS232 Data
4	NC
5	Signal Ground
6	CLK1IN-
7	CLK1IN+
8	CLK2IN+
9	CLK2IN-

P3 Board Configuration Header

The P3 eight-pin header is used to set board configuration and mode of operation. The pinout is defined in Table 10-11, “P3 Header Pinout,” below.

Table 10-11: P3 Header Pinout

Pin Number	Signal Name	PROM PROG	NORMAL	NORMAL W/BRK DETECT
1	RSTCONF [1]		JUMPER	JUMPER
2	GND	JUMPER		
3	PORESET			
4	NMI			
5	GND			JUMPER
6	BREAK_DET [2]			
7	GND			
8	FACTORY_JMP [3]			

1. RSTCONF is grounded for normal operation. Jumper open uses default hardware configuration word where the reset of the 60x bus components will not be configured.
2. BREAK_DET provides a hardware reset with a serial break detect sequence.
3. FACTORY_JMP for factory default initialization.

Chapter

11

Memory

Overview

This chapter discusses the various memory components and settings on the PCI384. Key topics in this chapter include:

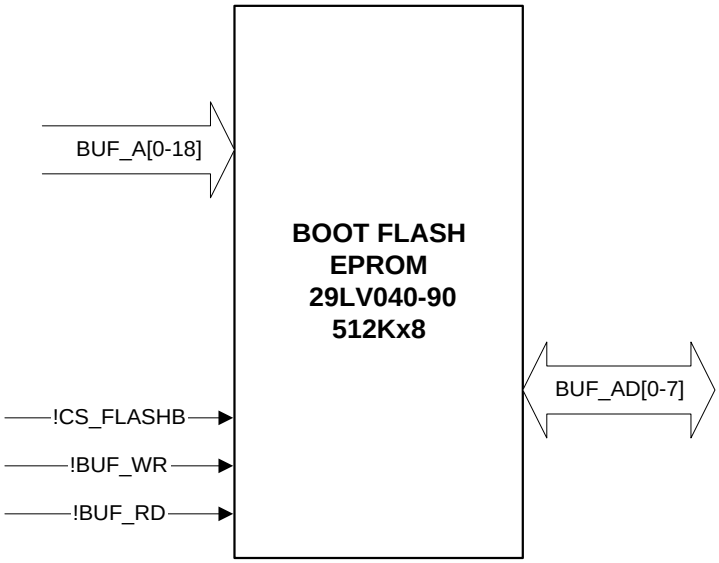
Boot Flash Memory

The Boot Flash EPROM is used to store the initial startup code for the PCI384 MPC8260 CPU. The device is socketed and mounted on the base board at position U9. Figure 11-1, “Boot Flash Block Diagram,” presents a block diagram for the boot flash.

The boot flash is a non-volatile memory that has the following general characteristics:

- The flash is a single power supply, 4 MB, 3.0 Volt-only flash memory device organized as 524,288bytes. The data appears on DQ0-DQ7.
- The device is in a 32-pin PLCC package.
- All read, erase, and program operations are accomplished using only a single power supply. Internally generated and regulated voltages are provided for the program and erase operations.
- The device is entirely command set compatible with the JEDEC single-power-supply flash standard.

Figure 11-1: Boot Flash Block Diagram



The device is accessed by the MPC8260 through the Buffered Data bus. The Buffered data bus supplies a buffered MPC8260 address and a bidirectional 8 or 16 bit data bus. The CS_FLASHB supplied by the MPC8260, acts as the device chip enable. The BUF_RD supplied by PSM517 PAL logic acts as the device output enable and BUF_WR again supplied by the PSM517 logic is the write enable. The PCI384 has a write protect feature that will not allow the BUF_WR signal to activate unless the flash_wp bit is set in the General Purpose registers. The default after reset is to disable writing to the flash. The read and write cycle is timed by the GCPMs wait state programming. There is no cycle termination signal generated by the device.

MPC8260 Related Flash Settings - CS0

The MPC8260 uses its chip select 0 (CS0) as the $\overline{\text{CS\_FLASHB}}$ signal. The CS0 Base Register is set to define the address space as FFF0_0000h to FFF7_FFFF H. Table 11-1, “CS0 Base Register Settings,” shows a list of recommended Base and Option Register field settings other than the base address and address masks:

Table 11-1: CS0 Base Register Settings

Field	Recommended Setting	Description
PS	01	8-bit
DECC	00	Data Errors checking disabled
WP	0	Read and write accesses are allowed
MSEL	000	Machine Select GPCM-60x bus
EMEMC	0	Accesses are handled by the MSEL selected controller. No external memory controller allowed in this space.
ATOM	00	The address space is controlled by the memory controller bank and is not used for atomic operations
DR	0	No data pipe lining is done.
V	1	This bank is valid

The Option Register Settings follow the form outlined for the GPCM Machine, as shown in Table 11-2, “CS0 Option Register Settings,” below.

Table 11-2: CS0 Option Register Settings

Field	Recommended Setting	Description
BCTLD	0	$\overline{\text{BCTLx}}$ is asserted upon access to the current memory bank
CSNT	1	$\overline{\text{CS}}$/WE is negated quarter of a clock earlier
ACS	11	$\overline{\text{CS}}$ is output half a clock after the address lines.
SCY	6	Clock Wait states (2+SCY)*clocks
SETA	0	$\overline{\text{PSDVAL}}$ is generated by the GCPM.
TRLX	1	Relaxed timing is generated by the GCPM for this memory region.
EHTR	00	Normal Idle Timing

General Flash Information

Flash access in the read mode is done as any normal PROM device.

The programming or write operations are entirely command set compatible with the JEDEC single-power-supply flash standard. Commands are written to the command register using standard microprocessor write timings. Register contents serve as input to an internal state-machine that controls the erase and programming circuitry. Write cycles also internally latch addresses and data needed for the programming and erase operations. Device programming occurs by executing the program command sequence. This initiates the Embedded Program algorithm, an internal algorithm that automatically times the program pulse widths and verifies proper cell margin.

Device erasure occurs by executing the erase command sequence. This initiates the Embedded Erase algorithm, an internal algorithm that automatically preprograms the array (if it is not already programmed) before executing the erase operation. During erase, the device automatically times the erase pulse widths and verifies proper cell margin. The host system can detect whether a program or erase operation is complete by reading the DQ7 (Data/Polling) and DQ6 (toggle) status bits. After a program or erase cycle has been completed, the device is ready to read array data or accept another command.

The sector erase architecture allows memory sectors to be erased and reprogrammed without affecting the data contents of other sectors. Device hardware data protection measures include a low VCC detector that automatically inhibits write operations during power transitions. The hardware sector protection feature disables both program and erase operations in any combination of the sectors of memory. This is achieved via programming equipment.

The Erase Suspend feature enables the user to put erase on hold for any period of time to read data from, or program data to, any sector that is not selected for erasure. True background erase can thus be achieved.

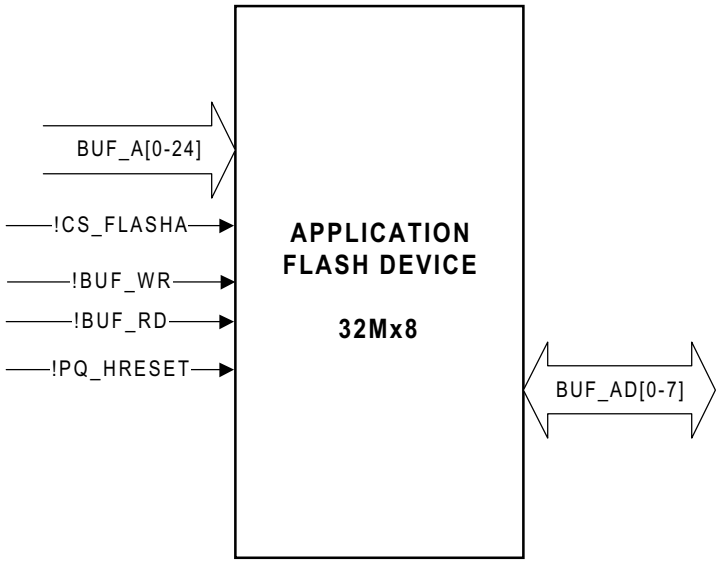
Application Flash Memory

The Application Flash PROM is used to store application code that will not be executed at boot up time. The device is permanently mounted to the base board at position U14. Figure 11-2, “Application Flash Block Diagram,” on page 119 presents a block diagram for the application flash.

The application flash is a non-volatile memory that has the following general characteristics:

- The flash is a single power supply, 32 MB, 3.0 V-only flash memory device.
- Read accesses to the device are the same as any EPROM. The data appears on DQ0-DQ7 on the buffered data bus.
- For writing purposes, the device has 127 blocks of byte wide data storage. The storage blocks have 128 KB bytes of storage each.
- Selecting, writing and erasing the blocks is done by using a Common Flash Interface (CFI) and a Scalable Command Set (SCS).
- All read, erase, and program operations are accomplished using only a single power supply. Internally generated and regulated voltages are provided for the program and erase operations.
- The device package is a 56-pin TSOP.

Figure 11-2: Application Flash Block Diagram



The device is accessed by the MPC8260 through the Buffered Data bus. The Buffered data bus supplies a buffered MPC8260 address and a bidirectional 8 or 16 bit data bus. The CS_FLASHA is supplied by the MPC8260 and acts as the device chip enable. The BUF_RD is supplied by the PSM517 PAL logic and acts as the device output enable. The PSM517 also supplies the BUF_WR strobe which is the write enable. The PCI384 has a write protect feature that will not allow the BUF_WR signal to activate unless the flash_wp bit is set in the General Purpose registers. The default after reset is to disable writing to the flash. The read and write cycle is timed by the GCPMs wait state programming. There is no cycle termination signal generated by the device.

MPC8260 Related Flash Settings - CS3

The MPC8260 uses its chip select 3 (CS3) as the $\overline{\text{CS\_FLASHA}}$ signal. The CS3 Base Register is set to define the address space as 1000_0000h to 10FF_FFFF H. Table 11-3, “CS3 Base Register Settings,” shows a list of recommended Base and Option Register field settings other than the base address and address masks.

Table 11-3: CS3 Base Register Settings

Field	Recommended Setting	Description
PS	01	8-bit
DECC	00	Data Errors checking disabled
WP	0	Read and write accesses are allowed
MSEL	000	Machine Select GPCM-60x bus
EMEMC	0	Accesses are handled by the MSEL selected controller. No external memory controller allowed in this space.
ATOM	00	The address space is controlled by the memory controller bank and is not used for atomic operations.
DR	0	No data pipe lining is done.
V	1	This bank is valid

The Option Register Settings follow the form outlined for the GPCM Machine, as shown in Table 11-4, “CS3 Option Register Settings,” below.

Table 11-4: CS3 Option Register Settings

Field	Recommended Setting	Description
BCTLD	0	$\overline{\text{BCTLx}}$ is asserted upon access to the current memory bank
CSNT	1	$\overline{\text{CS}}$/WE is negated a quarter of a clock early.
ACS	11	$\overline{\text{CS}}$ is output half a clock after the address lines.
SCY	6	Clock Wait states
SETA	0	$\overline{\text{PSDVAL}}$ is generated by the GCPM.
TRLX	1	Relaxed timing is generated by the GCPM for this memory region.
EHTR	00	Normal Idle Timing

General Application Flash Information

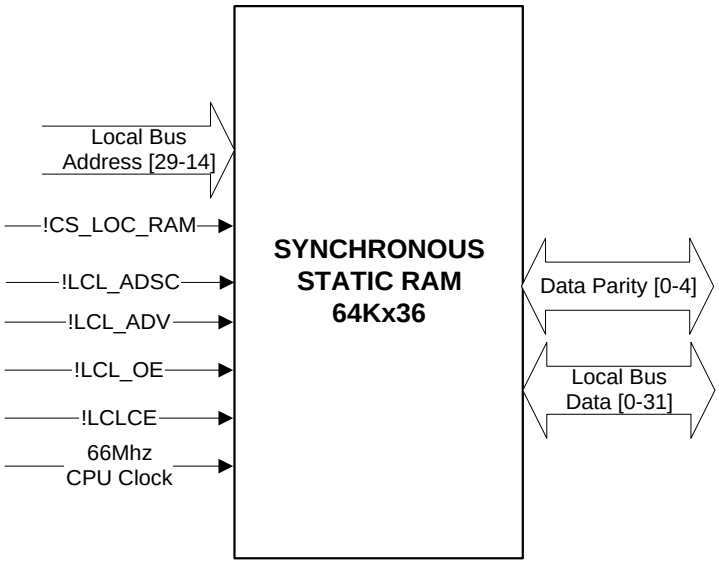
Flash access in the read mode is done as any normal PROM device.

Selecting, writing and erasing the blocks is done by using a Common Flash Interface (CFI) and a Scalable Command Set (SCS). Refer to Intel’s *3 Volt Intel STRATAFLASH Memory Datasheet* for the command set and the programming strategy.

Communications Buffer Memory

The Communications Buffer Memory, is used by the Communications Processor Module (CPM) in the MPC8260. Figure 11-3, “Communications Buffer Memory Block Diagram,” presents a block diagram for the communications buffer memory. The memory array is interfaced directly to the MPC8260s Local Bus connection. The Local Bus interface is tightly coupled to the CPM inside the part making it an ideal resource for use as a high-speed buffer memory for data and commands. The memory buffer is located on the base board assembly.

Figure 11-3: Communications Buffer Memory Block Diagram



The Communications Memory Buffer is implemented with synchronous static random access (SSRAM) memory operating at 3.3 V. 256 KB of data are available and organized as 64 KB x 32-bits of data. The flow-through device is supported, as opposed to the pipelined device. Individual parity for each byte lane is provided in 64 KB x 4-bits of memory (The default controller setting is parity disabled). Both data and parity are implemented in a monolithic component that is wired directly to the MPC8260 local bus.

The device is controlled by the MPC8260's Universal Programmable Memory (UPM) controller. A 66 MHz clock is provided to the memory from the same low-skew clock driver that provides 66 MHz to the MPC8260. The active-high and active low chip selects are permanently enabled via external resistors. Global Write Enable (GWE#) is pulled-up to permanently disable the global override of byte-write enable. Address Status Processor (ADSP#) is pulled-up to permanently disable burst abort with READ override. Snooze (ZZ) is grounded to permanently disable snooze. Burst mode (LBO#) is grounded to permanently select linear burst. FT#/VSS is a ground pin on this device.

The memory interface consists of 32 data (LCL_D[0-31]), 4 byte-parity (LCL_DP[0-4]), 4 byte-write selects (LCL_LBS[0-3]), 1 write enable (CS_LOC_RAM), 1 Address Status Controller (LCL_ADSC), 1 Address Advance (LCL_ADV), 1 output enable (LCL_OE), 1 chip enable (LCL_CE), and 16 address (LCL_A29-A14). The seventeenth address connection to memory is grounded. LCL_OE, LCL_ADSC, LCL_ADV, and LCL_CE are driven from the MPC8260s LGPL[0-3] pins, respectively.

MPC8260 Related Communications Buffer Memory Settings

The MPC8260 uses its chip select 2 (CS2) as the $\overline{\text{CS\_LOC\_RAM}}$ signal. The CS2 Base Register is set to define the address space as 2080_0000h to 2083_FFFF H. Table 11-5, “CS2 Base Register Settings,” shows a list of recommended Base and Option Register field settings other than the base address and address masks:

Table 11-5: CS2 Base Register Settings

Field	Recommended Setting	Description
PS	11	32-bit
DECC	00	Data Errors checking disabled (default)
WP	0	Read and write accesses are allowed
MSEL	101	Machine Select UPMB for Local Bus
EMEMC	0	Accesses are handled by the MSEL selected controller. No external memory controller allowed in this space.
ATOM	00	The address space is controlled by the memory controller bank and is not used for atomic operations
DR	0	No data pipe lining is done.
V	1	This bank is valid

The Option Register Settings follow the form outlined for the UPMB Machine, as shown in Table 11-6, “CS2 Option Register Settings,” below.

Table 11-6: CS2 Option Register Settings

Field	Recommended Setting	Description
BCTLD	1	$\overline{\text{BCTLx}}$ is not asserted upon access to the current memory bank
BI	0	This Bank of memory supports bursting
EHTR	00	Normal Idle Timing

UPMB Table Settings

The UPMB Table settings are as follows:

```
; single read   (offset 0x00 in upm ram)
       .long   0xffe33000, 0xffaff005, 0xfffff000, 0xfffff000
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
; burst read    (offset 0x08 in upm ram)
       .long   0xffe33008, 0xffacf00c, 0xffacf00c, 0xffacf00c
       .long   0xffa33004, 0xffacf004, 0xffacf004, 0xffacf004
       .long   0xffaff005, 0xfffff000, 0xfffff000, 0xfffff000
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
; single write (offset 0x18 in upm ram)
       .long   0x00f33005, 0xfffff000, 0xfffff000, 0xfffff000
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
; burst write   (offset 0x20 in upm ram)
       .long   0x00f3300c, 0x00fcf00c, 0x00fcf00c, 0x00fcf00c
       .long   0x00f33004, 0x00fcf004, 0x00fcf004, 0x00fcf005
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
; refresh       (offset 0x30 in upm ram)
       .long   0xfffff005, 0xfffff000, 0xfffff000, 0xfffff000
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
       .long   0xfffff000, 0xfffff000, 0xfffff000, 0xfffff000
; exception     (offset 0x3C in upm ram)
       .long   0xfffff005, 0xfffff000, 0xfffff000, 0xfffff000
UPMBTableEnd:
```

PCI Interface

The Tundra PowerSpan part number CA91L8260 is a PowerPC-to-PCI bus switch. The Single PCI PowerSpan variant, part number CA91L8260, is a two port device which has one PCI Interfaces and a PowerPC Processor Bus Interface.

PowerSpan Signal Connections and Hardware Configuration

The bus interface signals between PowerSpan and the MPC8260 are directly connected. External pull-up resistors are connected to all bus control signals to ensure that they are held in an inactive state when not accessed. Additionally pull-up resistors are provided on the address and data bus interfaces to minimize current consumption of the buffers when tri-stated. Refer to the schematic documents for connection information.

No provision has been made for parity checking on either of the address or data presented to the Tundra.

This choice was made due to the multiplexed nature of the 8260 pins being used for interrupts and bank selection of the SDRAM memory.

Since either the MPC8260 or PowerSpan can be bus masters and both have internal arbiters one part must be selected as the master. The 8260 is selected as the master and uses its internal arbiter. The PowerSpan connections for address bus or data bus requests are programmed to be outputs and are therefore inputs to the 8260 arbiter.

The MPC8260 sees the Tundra 60x bus registers at 3000_0000 to 3000_0fff H, which is the default 60x bus address decode for the Tundra device.

The CompactPCI P1 connections are designed to be a 64-bit interface and is selectable to be either 33 MHz or 66 MHz configurable by the backplane.

Due to a Tundra component problem a circuit modification was included which adds an additional load on the P1-IDSEL line. Also the present parts have timing issues with the PCI bus running at 66 MHz. Refer to the following document for explanation from Tundra:

- 80A1000_ER001_06.pdf - *PowerSpan (CA91L8260/CA91L8260) Device Errata and Design Notes*

For further information reference the following documents available on the Tundra Web site for circuit connection, description and operation:

http://www.tundra.com

At the Tundra Web site search for the following documents in the PowerSpan Design Support Tools section which are current at the time of this release:

- 80A1000_FB001_04.pdf - *PowerSpan Feature Brief*
- 80A100B_AN001_01.pdf - *PowerSpan (CA91L8260) as a PCI to PCI Bridge*
- 80A100A_AN001_01.pdf - *PowerSpan (CA91L8260) MPC8260 Connection Application Note*
- 80A1000_ER001_06.pdf - *PowerSpan (CA91L8260/CA91L8260) Device Errata and Design Notes*
- 80A1000_MA001_08.pdf - *PowerSpan (CA91L8260, CA91L8260) PowerPC-to-PCI Bus Switch Manual*

PowerSpan Reset Configuration Word

The MPC8260 system provides support for a single Configuration Master and up to seven Configuration Slaves. In the PCI384, the MPC8260 is setup to be the configuration master. During the assertion of HRESET, the Configuration Master, (MPC8260) reads configuration words from memory and writes them to the Configuration Slaves. A total of seven 64-bit words are transferred over the data bus. One of Address[0:6] lines is strobed to transfer each word.

The PowerSpan acts as a Configuration Slave because the following conditions have been preset:

- PB_RSTCONF connected to one of MPC8260 Address[0:6]
- PB_RST connected to MPC8260 HRESET
- PB_D[0-31] connected to MPC8260 D[0-31]

The reset configuration word consists of eight bits set on the PB_D[0-7] bits:

Bit 0 - 0 - disable PB arbiter - PowerSpan is not the 60X arbiter
Bit 1 - 0 - disable PCI-1 arbiter - PowerSpan is not the PCI-1 arbiter
Bit 2 - 0 - disable PCI-2 arbiter - No PCI-2 arbiter present
Bit 3 - 0 - PCI-1 is primary PCI
Bit 4 - 1 - disable PCI-1 REQ64
Bit 5 - 0 - PB Boot - get local configuration from the processor bus.
Bit 6 - 0 - disable debug mode
Bit 7 - 0 - disable PLL bypass mode

Local Processor Connections Between MPC8260 and PowerSpan

The design is similar to the referenced Tundra documentation except in the following signals described in Table 11-7, “Signal Variations Between PowerSpan and the MPC8260,” below. The variation being in the selection of making the MPC8260 the bus arbiter and the inability to select the pin modes due to the multiplexed pins on the MPC8260. Also a different means of controlling the board reset is done either by the MPC8260 processor or PCI bus.

Table 11-7: Signal Variations Between PowerSpan and the MPC8260

Power Span Signal	MPC8260 Signal	Description or comments
$\overline{\text{PB\_BR}}$[1]	BR	Address Bus Request - 8260 arbitration
$\overline{\text{PB\_BG}}$[1]	BG	Address Bus Grant - 8260 arbitration
$\overline{\text{PB\_DBG}}$[1]	DBG	Data Bus Grant - 8260 arbitration
PB_DBB	DBB	Data Bus Busy - 8260 arbitration
PB_AP[0:3]	AP[0:3]	Address bus parity - not used
PB_DP[0:7]	DP[0:7]	Data bus parity - not used
PB_CI	CI	Cache Inhibit - not used

Table 11-8, “PowerSpan Reset Pins,” itemizes the Reset operation of the PCI384 PCI ports.

Table 11-8: PowerSpan Reset Pins

Pin Name	Direction	Description	Comments
PO_RST	Input	Power-On Reset	Voltage Level Sense
HEALTHY	Input	Board Status	Power Good Asserts Signal
PB_RST	Input	Processor Bus Hard Reset	PB_RST_DIR = 0 (Input) HRESET generated
P1_RST#	Input	PCI-1 Bus Reset	P1_RST_DIR = 0 (Input) CPCI bus generated
TRST	Input	JTAG Reset	Voltage Level Sense

PowerSpan Interrupts

Table 11-9, “PowerSpan Interrupt Connections,” documents the interrupt connections from the PowerSpan.

Table 11-9: PowerSpan Interrupt Connections

PowerSpan Signal	MPC8260 Signal	Description or comments
$\overline{\text{INT\_3}}$	$\overline{\text{PQ\_IRQ4}}$	General Purpose PowerSpan Interrupt
$\overline{\text{INT\_5}}$	$\overline{\text{PQ\_SRESET}}$ (via PAL logic)	Mailbox generates SRESETTo MPC8260

Appendix

A

Agency Approvals

Overview

This appendix presents agency approval and certification information for the PCI384 Four-Channel T1/E1/J1 Telecom Adapter.

The PCI384 is certified as indicated in the following sections. If a certification is not listed below, the PCI384 may still comply. Contact Performance Technologies for current product certifications and availability.

Topics covered in this chapter include:

CE Certification

The product(s) described in this manual conform to the *EU 89/336/EEC Electromagnetic Compatibility Directive*, amended by *92/31/EEC* and *93/68/EEC*.

The product described in this manual is the **PCI384**. The product identified above complies with the *EU 89/336/EEC Electromagnetic Compatibility Directive* by meeting the applicable EU standards as outlined in the Declaration of Conformance. The Declaration of Conformance is available from Performance Technologies or from your authorized distributor.

ETSI EN 300 386

Electromagnetic Compatibility and Radio Spectrum Matters (ERM); Telecommunications Network Equipment; Electromagnetic Compatibility (EMC) Requirements.

ISO 9002 Notice

Performance Technologies, Inc., is registered by BVQI as ISO 9002 Compliant.

UL 1950

Safety certification.

FCC (USA) Class A Notice

This equipment has been tested and found to comply with the limits for a Class A digital device, pursuant to Part 15 of the FCC Rules. These limits are designed to provide reasonable protection against harmful interference when the equipment is operated in a commercial environment.

This equipment generates, uses, and can radiate radio frequency energy and, if not installed and used in accordance with the instruction manual, may cause harmful interference to radio communications. Operation of this equipment in a residential area is likely to cause harmful interference in which case the user will be required to correct the interference at his own expense.

This device complies with Part 15 of the FCC Rules. Operation is subject to the following two conditions:

1. This device may not cause harmful interference, and
2. This device must accept any interference received, including interference that may cause undesired operation.

Note: *Modifications made to this device that are not approved by Performance Technologies, Inc. may void the authority granted to the user by the FCC to operate this equipment.*

Industry Canada Class A Notice

This Class A digital apparatus complies with Industry Canada's Equipment Standard for Digital Equipment (ICES-003).

Cet appareil numérique de la classe A est conforme à la norme NMB-003 du Canada.

Safety Information

This section is provided as a summary of the safety recommendations throughout this manual. Performance Technologies, Incorporated (PTI) recommends that all safety precautions are followed to prevent harm to yourself or the equipment. Please follow all warnings marked on the equipment.

Safety Precautions

Caution:
To reduce the risk of fire, use only No. 26 AWG or larger telecommunications line cord for your outside cable connection. Belden ABAM No. 22 AWG is a typical selection.

Caution:
Follow all warnings and instructions marked on the equipment.

Caution:
Ensure that the voltage and frequency of your power source matches the voltage and frequency inscribed on the equipment's electrical rating label.

Caution:
Never push objects of any kind through the openings in the equipment. Dangerous voltages may be present. Conductive foreign objects could produce a short circuit that could cause fire, electrical shock, or damage the equipment.

Caution:
Electronic components on printed circuit boards are extremely sensitive to static electricity.

Ordinary amounts of static electricity generated by your clothing or work environment can damage the electronic equipment. It is recommended that anti-static ground straps and anti-static mats are used when installing the board in a system to help prevent damage due to electrostatic discharge.

Compliance with RoHS and WEEE Directives

In February 2003, the European Union issued *Directive 2002/95/EC* regarding the Restriction of the use of certain Hazardous Substances in electrical and electronic equipment (RoHS) and *Directive 2002/96/EC* on Waste Electrical and Electronic Equipment (WEEE).

This product is compliant with *Directive 2002/95/EC*. It may also fall under the *Directive 2002/96/EC*.

Performance Technologies' complete position statements on the RoHS and WEEE Directives can be viewed on the Web at: http://pt.com/page/about-us/ehsms/.