

USER MANUAL
AGILENT ACQIRIS
SIGNAL ANALYZERS
INCLUDING
BASE TEST
AND
FFT SPECTROMETER FIRMWARE

Models covered:

AC210 [U1091AAC1](#)

AC240 [U1080A-001](#)

Manual Part Number

U1092-90006

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1. Out of the Box

1.1. Message to the User

Congratulations on having purchased an Agilent Technologies Acqiris data conversion product. Acqiris Digitizers /Analyzers are high-speed data acquisition modules designed for capturing high frequency electronic signals. To get the most out of the products we recommend that you read this manual carefully. We trust the product you have purchased will meet with your expectations and provide you with a high quality solution to your data conversion applications.

1.2. Using this Manual

The AC240 and AC210 analyzer platforms are dual- and single-channel 6U CompactPCI®/PXI™ 1 GS/s per channel digitizers with on-board real-time data processing.

The Analyzer functions are dependent on the firmware code loaded into the Data Processing Unit. They are described in this manual throughout the Firmware sections.

This guide assumes you are familiar with the operation of a personal computer (PC) running a Windows 2000/XP/Vista/7 (32/64) or other supported operating system. It also assumes you have a basic understanding of the principles of data acquisition using either a waveform digitizer or a digital oscilloscope.

The manual is divided into 6 separate sections. To understand the elements of operation for the module it is essential that you carefully read chapters 1-5.

- | | |
|-----------|--|
| Chapter 1 | <i>OUT OF THE BOX</i> , describes what to do when you first receive your new Acqiris product. Special attention should be paid to sections on safety, packaging, and product handling. Before installing your product please ensure that your system configuration matches or exceeds the requirements specified. |
| Chapter 2 | <i>INSTALLATION</i> , covers elements of installation. Before attempting to use your Acqiris product for actual measurements we strongly recommend that you read all sections of this chapter. |
| Chapter 3 | <i>PRODUCT DESCRIPTION</i> , provides a short description of all the functional elements of the AC210 and AC240 Modules. |
| Chapter 4 | <i>FIRMWARE</i> , provides a detailed description of the Base Test and FFT Spectrometer firmware including algorithm and implementation aspects. |
| Chapter 5 | <i>RUNNING THE ANALYZER DEMO</i> Application describes the operation of the Analyzer Demo application. Analyzer Demo allows interactive operation of the AC240/AC210 from a PC running a Windows 2000/XP/Vista/7 (32/64) operating system. |
| Chapter 6 | <i>PROGRAMMING THE FIRMWARE</i> , provides all the information required to write a software application based on the FFT Spectrometer firmware. |

For information necessary for writing your own software to control Acqiris products you should also refer to the **Programmer's Guide** and the **Programmer's Reference Manual**.

1.3. Conventions Used in This Manual

The following conventions are used in this manual:



This icon to the left of text warns that an important point must be observed.

WARNING Denotes a warning, which advises you of precautions to take to avoid being electrically shocked.

CAUTION Denotes a caution, which advises you of precautions to take to avoid electrical, mechanical, or operational damages.

NOTE Denotes a note, which alerts you to important information.

Italic text denotes a warning, caution, or note.

Bold Italic text is used to emphasize an important point in the text or a note

`mono` text is used for sections of code, programming examples, and operating system commands.

B,KB,MB,GB is for Byte, KiloByte = 1024 bytes, MegaByte = 1024*1024 bytes, GigaByte = 1024*1024*1024 bytes

b,Kb,Mb is for bit with multipliers as above.

Triggered Denotes a VHDL object. It could be a single- or multi-bit signal or a component or a library name.

0xn..n Denotes a hexadecimal value.

1.4. Model Names

Agilent Technologies Inc. acquired Acqiris SA and its product lines in December 2006. Use the tables below to cross reference the legacy model name and new Agilent numbers

Agilent Model Number	Acqiris Model Name
U1091AAC1	AC210
U1080A-001	AC240

1.5. Disclaimer and Safety

The AC Series CompactPCI Analyzer cards have been designed to operate in a CompactPCI/PXI crate. Power for the modules is provided by plugging them into a free slot (refer to the installation procedure). Agilent Technologies does not recommend operation of the AC210 or AC240 outside of a CompactPCI/PXI crate



CAUTION: *Do not exceed the maximum input voltage rating! The maximum input voltage for 50 Ω input impedance is ± 5 V.*

1.6. Warning Regarding Medical Use

The AC Series CompactPCI Analyzer cards are not designed with components and testing intended to ensure a level of reliability suitable for use in treatment and diagnosis of humans. Applications of these cards involving medical or clinical treatment can create a potential for accidental injury caused by product failure, or by errors on the part of the user. These cards are *not* intended to be a substitute for any form of established process or equipment used to monitor or safeguard human health and safety in medical treatment.



WARNING: *The modules discussed in this manual have not been designed for making direct measurements on the human body. Users who connect an Acqiris module to a human body do so at their own risk.*

1.7. Packaging and Handling

Your Analyzer is shipped with the following components:

- A compact disc in an Agilent Technologies paper CD envelope that includes
 - 10 product User Manuals in electronic form (8-bit Digitizers, 10-bit Digitizers, 12-bit Digitizers, Averagers, Analyzers, Signal Analyzers, Streamer Analyzers, Time-to-Digital Converters, 3-, 5-, and 8-slot CompactPCI Crates, and the 21-slot CompactPCI Crate),
 - 1 Programmer's Guide and 1 Programmer's Reference Manual,
 - device drivers with sample software for different operating systems, environments and languages,
 - the Analyzer Demo application, a demonstration program for the AC/SC Analyzer products,
 - the AcqirisLive application, a demonstration program for our digitizer and averager products,
 - the SSR Demo application, a demonstration program for the Acqiris AP235/AP240 Analyzers,
 - the APX01 Demo application, a demonstration program for the Acqiris AP101/AP201 Analyzers,
 - the TC Demo application, a demonstration program for the Acqiris TC840/TC842/TC890 Time-to-Digital Converters,
 - full installation procedures for use with Microsoft Windows, National Instruments LabVIEW RT, Wind River VxWorks, IVI-COM/C, and Linux software.
- A declaration of conformity
- Optional documentation such as a model-dependent document giving Specifications & Characteristics, a Calibration Certificate, or a Performance Verification

After carefully unpacking all items, inspect each to ensure there are no signs of visible damage. Also check that all the components received match those listed on the enclosed packing list. Agilent cannot accept responsibility for missing items unless we are notified promptly of any discrepancies. If any items are found to be missing or are received in a damaged condition please contact the Agilent service center or your local supplier immediately. Retain the box and packing materials for possible inspection and/or reshipment.

1.8. Warranty

All Agilent Acqiris Digitizer products are warranted to operate within specification, assuming normal use, for a period of at least one year from the date of shipment. Units sold before April 2008 had three year warranties, as do some more recent ones; in case of doubt examine your invoice. It is recommended that yearly factory calibration be made in order to verify product performance. All repairs, replacement and spare parts are warranted for a period of 3 months. Warranty extensions are available as an option.

Agilent endeavors to provide leading edge technology that includes the latest concepts in hardware and software design. As such software and firmware used with the products is under continual refinement and improvement. All software and instrument firmware is supplied "as is" with no warranty of any kind. Software and firmware is thoroughly tested and thought to be functional at the time of shipment. At Agilent's discretion software and firmware may be revised if a significant operational malfunction is detected.

In exercising this warranty, Agilent will repair or replace any product returned to the Agilent service center, within the warranty period. The warranty covers all defects that are a result of workmanship or materials. This excludes defects that are caused by accident, misuse, neglect, or abnormal operation.

The purchaser is responsible for returning the goods to the nearest Agilent service center. This includes transportation costs and insurance. Agilent will return all warranty repairs with transportation prepaid.

1.9. Warranty and Repair Return Procedure, Assistance and Support

Agilent acquired Acqiris SA and its product lines in December 2006. Please contact your nearest Agilent Service Center before returning any product for repair.

You can find information about technical and professional services, product support, and equipment repair and service on the Web, see <http://www.agilent.com/find/service> (or <http://www.agilent.com/> and after selecting your country click on **Contact Us**). The service center will ask for your name, company, phone number and address, the model and serial numbers of the unit to be repaired, and a brief description of the problem.

Before issuing a Service Order the service center may ask you to communicate with us by phone or eMail so that we can learn as much as needed about the problems observed. If a unit returned under guarantee is found to be working normally and this procedure was not followed we reserve the right to charge you for the work done.

For your nearest customer support center please contact Acqiris Technical Support (ACQIRIS_SUPPORT@agilent.com) or come visit our web site at <http://www.agilent.com/find/acqiris>. Alternatively, contact Acqiris at 1-800-829-4444 in the USA, +41 22 884 32 90 in Europe or +61 3 9210 2890 in the Asia-Pacific region. The Agilent Support Centers can also help redirect you for any questions concerning the installation and operation of your equipment.

1.10. System Requirements

Acqiris products need the following minimum PC System Requirements in order to obtain reasonable performance from your digitizer.

Processor: 500 MHz Pentium (higher recommended). Some PowerPC systems running Wind River VxWorks are supported; please contact us for details.

Memory: 256 MB RAM. The previous number is a very rough estimate. Assuming that you are using AcqirisLive or an application of your own that operates on the acquired data it seems reasonable to ask for 10 times the total acquisition memory that you will be using at the same time in the application. Performance is likely to be degraded if less memory is available.

Display resolution: At least 1024 x 768 pixels and 256 colors for use of AcqirisLive or Demo applications.

Operating System: Microsoft Windows Vista, Vista 64-bit, 7, 7 64-bit, 2000/XP including 2003 Server, Wind River VxWorks 5.5.1 and 6.4, and Linux with kernels 2.4 and 2.6. Users with previous Windows OS versions (i.e. NT4) can download AcqirisSoftware 3.1 from the Agilent WEB site.



NOTE: Windows 2000 will not be supported in future releases.



NOTE: Wind River VxWorks 5.5.1 will not be supported in future releases.

Hard Drive Space: 300 MB for the Complete installation.

CD Drive (or any method to copy the software installation files from CD to the hard drive such as LAN, floppy drive, etc.)

LabVIEW: Full driver implementations are available for National Instruments LabVIEW versions 8.5, 8.2.1, and 8.0. LabVIEW 7.1 is frozen at the level of Acqiris Software 3.2 with support for all instruments.

LabVIEW RT: National Instruments LabVIEW RT is supported for the same versions as shown above. The VISA driver must be version 3.0 or higher.

MATLAB: The MEX interface can be used with MathWorks MATLAB 7.3 or a newer version.

Visual BASIC: The interface files and examples are available for Microsoft Visual Basic .NET.

Tornado: The example files are useable with Wind River Tornado 2.2.1.

1.11. Transport & Shipping



CAUTION: Cards can be safely transported in their original shipping packages. DC cards can be transported when properly mounted in a CompactPCI crate.

To package the instrument for shipping:

Step

Notes

1. Place the instrument in its original packaging materials.
 - If the original packaging materials are not available, use a professional packaging service. Contact your Agilent Service Center for more information.
2. Surround the instrument with at least 3 to 4 inches (8 to 10 cm) of its original packing material or bubble-pack to prevent the instrument from moving in its shipping container.
3. After wrapping it with packing material, place the instrument in its original shipping container or a strong shipping container that is made of double-walled corrugated cardboard with 159 kg (350 lb) bursting strength.
 - The shipping container must be large and strong enough to accommodate your instrument and allow at least 3 to 4 inches (8 to 10 cm) on all sides for packing material.
4. Seal the shipping container securely with strong nylon adhesive tape.
5. Mark the shipping container “FRAGILE, HANDLE WITH CARE” to help ensure careful handling.
6. Use the address obtained from your Agilent Technologies Service Center.
7. Retain copies of all shipping papers.



CAUTION: Damage can result if the original packaging materials are not used. Packaging materials should be anti-static and cushion the instrument on all sides. NEVER USE STYRENE PELLETS IN ANY SHAPE AS PACKAGING MATERIALS. They do not adequately cushion the instrument or prevent it from moving in the shipping container. Styrene pellets can also cause equipment damage by generating static electricity or by lodging in fan motors.

1.12. Maintenance

The cards do not require any maintenance. There are no user serviceable parts inside. A periodic factory calibration can be obtained on request.

1.13. Cleaning

Cleaning procedures consist only of exterior cleaning.

Clean the exterior surfaces of the module with a dry lint-free cloth or a soft-bristle brush. If any dirt remains, wipe with a cloth moistened in a mild soap solution. Remove any soap residue by wiping with a cloth moistened with clear water. Do not use abrasive compounds on any parts.

1.14. Disposal and Recycling



Electronic equipment should be properly disposed of. Acqiris Digitizers and their accessories must not be thrown out as normal waste. Separate collection is appropriate and may be required by law.

2. Installation

This chapter describes how to install the Acqiris hardware and software for Windows, National Instruments LabVIEW RT, Linux, or Wind River VxWorks.

 **NOTE:** For a first time installation we strongly recommend installing the software *before* inserting the hardware into the PC.

2.1. U1091AK02 IC414 Installation

 **NOTE:** If you are going to install an IC414 interface for the first time and are running Windows 2000/XP you should follow the procedure below *before* installing the Acqiris hardware.

2.2. IC414 Hardware installation hints

The PCI-8570/PXI-8570 User's Manual (Rev. 1.00) section 2.5 gives Hardware Installation instructions.

 **CAUTION:** Turn off the power of the PC; the PC may have to be unplugged to ensure that the PCI bus has no power available. Please ignore the PCI-8570 instruction to leave the power cord plugged in; Ground the chassis differently!

 **CAUTION:** Touch the antistatic package to a grounded object before removing the card from the package. Electrostatic discharge can damage the card.

The standard cable pair provided each have a red connector on one end and a black connector on the other. Therefore the correct connection can be made by plugging the Red connector into the LORx socket and the Black connector into the LOTx socket on the PXI module and the other Red connector into the PCI module socket furthest from the PCI card internal base connector and the Black connector into the next socket.

If you intend to use 64-bit 66 MHz transfer to maximize data transfer speed you should cable a “bundled link” using two standard cable pairs and both the L0 and L1 pairs of connectors. You should also make sure that you configure the PXI-8570 M66EN Jumper correctly.

2.2.1. IC414 Windows software installation

Windows Vista/7 (32/64), XP64 and Linux users do not need to read any further since there is no special software installation.

Windows users should have the hardware installed as noted above. This software installation should be done before any Acqiris modules are placed in the CompactPCI crates. This may mean that you have to remove the module from the crate *as delivered*.

The crate should be turned on first followed by the PC. If the cabling and start-up sequence is done correctly there will be no LED illuminated on the PCI unit connected pair and the LED's of the PXI connected pair will be lit.

For **Windows XP** installation, Select the **Control Panel** under Settings in the **Start** menu. Then, if you are using the **Category View** select **Printers and Other Hardware**. After this, for both **Category** and **Classic** views, go to **System** and then display the **Hardware** tab to get access to the **Driver Signing** menu. Since neither the AdLink nor the Acqiris driver has been submitted for **Windows Logo testing** you must select either the **Ignore** or **Warn** action. The resulting menu looks as shown:



The PCI-8570/PXI-8570 User's Manual (Rev. 1.00) section 2.4 contains the software installation instructions. These should be executed before allowing the hardware installation process to look for the driver. If you have an AdLink CD Version 2004A4 or later you can use it; if not you should download the latest driver from the WEB site (<http://www.adlinktech.com/>). You can then continue with the Hardware Installation. A reboot will then be necessary. At this point the Stargen Fabric PCI Adapter and the Stargen Aruba Fabric-to-PCI Bridge should appear correctly installed under System Devices in the Device Manager.



NOTE: If you have an AdLink CD Version 2005A3 or later you can find 8570install.exe in the folder X:\Driver Installation\PXI Platform\PXI Extension\PCI_PXI-8570\Wnt2kxp and the starfab1.inf file in the folder X:\Driver Installation\PXI Platform\PXI Extension\PCI_PXI-8570\Win98.

2.3. Installing the Software under Windows

2.3.1. Warnings

If Setup detects a previous installation of Acqiris software on your system, a warning screen will be displayed. It is recommended to exit Setup and uninstall older versions.

The installer from software releases prior to **Acqiris Software 2.0** installed the Digitizer Driver DLL files into the System directory. These will be removed by Setup. If you wish to keep the old installation on your system, you should exit Setup, and move all Acqiris driver files (acqiris*, acqrs* and acqir*) to some archive directory.

The DLL files will be installed into the bin subdirectory of the Acqiris software root, and the corresponding path will be added to the PATH environment variable.

2.3.2. Multiple Versions

With the software installation from **Acqiris Software 2.0** or above, it is possible to keep multiple versions on the same system, but you must specify a different root directory (i.e. Install Folder). If you keep the same directory, Setup will overwrite your previous installation.

To go back to a previous version, you must change the PATH environment variable and reinstall the Kernel driver.

1. Copy the SYS file from <old_AcqirisSoftware_root>\bin\kernel to the windows\System32\drivers directory.
2. Change the AcqirisDxRoot, AcqirisDxDir and PATH environment variables to the old root.
3. Reboot the computer.

2.4. Installation

For a first time installation on your computer Agilent recommends that you install the software BEFORE installing the hardware on your system. When upgrading to a new version, you should leave your modules installed and powered during installation.



NOTE: You will need administrator privileges to complete the software installation under Windows.

Complete the following steps to install the software for Windows 2000/XP/Vista/7 (32/64).

1. Insert the *Acqiris Software CD* into the CD-ROM drive of your computer. If the Autorun program does not start automatically (Autoplay disabled), you can start it manually, or navigate to the *AcqirisSoftware\Windows* folder in order to display the files included.
2. Choose **Install AcqirisSoftware for Windows 2000/XP/2003 Server/Vista/7** (or run Setup.exe from the *AcqirisSoftware\Windows* folder). After several seconds for initialization the first of many screens will appear.

Please note the following points:

- It is good practice to remove any previously installed version of Acqiris software. If the program finds that there is still Acqiris software installed on your machine a warning panel will appear.
- When upgrading from an old version under Windows 2000, you must uninstall all Agilent Acqiris devices manually prior to running the installer.
- In the **Select Install Type** window selecting **Custom** installation will let you select individual packages for loading. The space indicated for LabVIEW, Firmware and UserManual packages is incorrect. The correct values are 7 MB, 40 MB, and 30 MB respectively. A full installation requires just under 300 MB of disk.

- If MATLAB is installed on your machine, you will be asked to point the installer to the MATLAB root directory. You should do this if you want the installer to modify the standard startup.m file to add the paths for the MEX interface.
- In the **Installation Folder** window you will give the name of the root directory of the Acqiris software installation. If User Manuals (30 MB) and Firmware (40 MB) are loaded more space than indicated here will be required on the drive. For the case of a Tornado 2.2 installation the folder name should not contain any spaces.
- AcqirisLive needs the LabWindows/CVI 8.0 Run-Time Engine to run. If Setup has detected that a LabWindows/CVI Run-Time Engine is already installed on your system, it will ask you if you would like to install it locally for AcqirisLive anyway. If you are not sure about the version of the CVI Run-Time Engine on your system, it is recommended to install it locally.
- The **Installation Summary** window will be shown to allow you to check what you have asked for. At this point it is not too late to go back and make changes. The actual installation will only be started after clicking on “Install” in the next window.
- Please read the **IMPORTANT Information** window text. It could help you avoid serious problems.
- Registration of your installation will help Agilent provide you with better support. You will also be notified of updates and upgrades. All information submitted will be treated in accordance with Agilent’s privacy policy. Setup will prepare a registration e-mail in your e-mail client application upon termination of the setup procedure. You can then decide whether or not you wish to send it. You may also add comments. Uncheck the box if you do not want to register your installation.
- After the software installation is complete you can either accept the suggestion to restart the computer or you should shutdown your computer and proceed with the hardware installation.

2.5. Installing the Software for Linux

Please refer to the README file on the Acqiris Software CD-ROM (AcqirisSoftware/linux folder) for detailed instructions to install the software for linux systems.

2.6. Installing the Hardware

1. Turn off the power of the PC and the crate in the case of a CompactPCI module.



CAUTION: *For PCI modules the PC may have to be unplugged to ensure that the PCI bus has no power available. However, CompactPCI crates can be left plugged in since this ensures proper grounding.*



CAUTION: *Touch the antistatic package to a grounded object before removing the card from the package. Electrostatic discharge can damage the card.*

2. Module in a PC: open the PC, identify a free PCI slot and carefully insert the DP Series card into it. Be sure to ground yourself by touching the grounded PC frame and avoid touching any components on the DP Series card. Make sure that the grounding of the card’s mounting bracket to the back panel rail of the computer is done correctly. If present make sure that the fan’s adjustable retainer is correctly positioned and tightened for mechanical support. Close the PC.

Module in a CompactPCI crate: Follow the instructions of the crate manufacturer to insert the DC Series card into a free CompactPCI peripheral slot. Be sure to ground yourself by touching the grounded crate and avoid touching any components on the DC Series card. Be sure to tighten both front panel mounting screws to lock the module into place and insure proper grounding of the frame.



NOTE: *To ensure the best possible performance, users of Acqiris CC121 Crates with AS bus systems should respect the module placement rules to be found in the Agilent Acqiris 21-slot CompactPCI Crate User Manual.*

PCI Bus extension module in a PC: Consult the manufacturer's documentation for any special instructions. Open the PC, identify a free PCI slot and carefully insert the card into it. Be sure to ground yourself by touching the grounded PC frame and avoid touching any components on the card. Make sure that the grounding of the card’s mounting bracket to the back panel rail of the computer is done correctly. Close the PC. Connect the module to the CompactPCI crate controller.

3. Turn on the power of the crate(s), if present, and then the PC and start the operating system.



NOTE: Acqiris digitizers are equipped with a LED. If this LED is not glowing orange or red when the power is applied there is a severe problem. Either the module is broken or the necessary voltages for its use are not available.



NOTE: For proper system operation when using the IC200, IC414, or other PCI extension interface to connect a CC10X crate to a remote PC, the crate **must** be powered on before the PC in order for the PC BIOS to recognize the presence of the CompactPCI crate.

4. If devices were installed using a previous version of Acqiris software the instruments in these logical positions may still appear as *Unknown Devices*. This can be changed to the new Acqiris type category with the Grey Diamond icon by **Uninstalling** the device and then **Installing** again. Instructions on this procedure can be found in the **ReadMe.txt** file in the **manuals** folder of your Acqiris software installation.

2.7. After Restarting

2.7.1. Windows 2000

Under Windows 2000, you *must* login with administrator privileges after the first boot following the hardware installation; the Plug&Play system must have the appropriate privileges to be able to complete your hardware installation successfully. After a successful hardware installation, you will be able to use your Acqiris digitizer(s) with normal privileges.

At the first boot following the hardware installation, Windows will detect the new hardware and will install the devices automatically. The following image will appear.



NOTE: In some systems an application program (such as AcqirisLive) will not yet work correctly at this point. One additional boot cycle may be needed if this is the first time that a hardware board is being installed.

2.7.2. Windows XP/Vista/7

Under these Windows versions, you *must* login with administrator privileges after the first boot following the hardware installation; the Plug&Play system must have the appropriate privileges to be able to complete your hardware installation successfully. After a successful hardware installation, you will be able to use your Acqiris digitizer(s) with normal privileges.

If you login with administrator privileges after the first boot following the hardware installation, Windows will detect the new hardware and start the “Found New Hardware Wizard” after a few seconds. There is no need to use Windows Update to search for the software. You can “Install the software automatically”.



NOTE: In some systems an application program (such as AcqirisLive) will not yet work correctly at this point. One additional boot cycle may be needed if this is the first time that a hardware board is being installed.

2.8. LabVIEW RT

During program development you can choose whether you use LabVIEW or LabVIEW RT compatible libraries by switching the version present in **National Instruments\LabVIEW m.n\instr.lib**. This swap can be facilitated by using the **Install VI library for LabVIEW or LabVIEW Real-Time** shortcut available in the Shortcut folder under **Start** → **Programs**.

There is only one Acqiris Driver. It supports all Acqiris Instruments. The instructions below concern LabVIEW RT as used in NI PXI processors.

The Aq_RT.inf and AqRT_4.ini files must be uploaded to the target. To do this,

- start the MAX application,
- right click on the target
- select file transfer
- select the Aq_RT.inf file on your host machine and upload ('To Remote') to the LabVIEW RT working directory (/NI-RT/system) on the target

- select the Aq_RT.ini file on your host machine and upload ('To Remote') to the LabVIEW RT working directory (/NI-RT/system) giving it the name AqDrv4.ini

For Acqiris modules which need FPGA files you should,

- create the folder \firmware in the /NI-RT/system directory using the file transfer application
- select the FPGA files (from <AcqirisDxRoot>\Firmware) you want to copy to the target and upload them into the firmware directory

Restart the target after finishing the file transfers.

Restart the MAX Explorer and you should have Acqiris digitizers detected in your PXI system.

2.9. Installing the IVI-COM/C Driver

Please install Acqiris software for Windows first. Then the *Acqiris Software CD* Autorun program gives access to two installers:

- IVI Shared Components 1.4
- IVI-COM/C IviAqD1 driver

These must be installed in the order shown above. For more information you can then consult the Readme.txt file in the IVI\Drivers\IviAqD1 folder or the documentation through the program shortcut present under Ivi/IviAqD1.

2.10. Uninstalling devices under Windows

In the Device Manager, select the instrument to be uninstalled. Choose “Uninstall” from the “Action” menu. After all desired instruments have been uninstalled select “Scan for hardware changes” from the “Action” menu, or reboot the computer. Note that only those devices that are actually physically present are visible in the Device Manager.

2.11. Distribution for Windows 2000/XP/Vista/7 (32/64) and Linux

The manuals/ReadMe.txt file contains a list of files to be found after a complete installation of Acqiris software on Windows systems. Similarly the ReadMeLinux file gives the list of files corresponding to that installation.

3. Product Description

3.1. Overview

The AC Series are powerful analyzer platforms based on a single-channel (AC210) or dual channel (AC240) 1 GS/s per channel, 8-bit CompactPCI digitizer with on-board real-time data processing implemented with the aid of a large field-programmable gate array (FPGA).

The AC Series Architecture offers a convenient way to implement user defined data processing algorithms with data rates of up to 2 GS/s. It provides external processing control through dedicated connectors and visual reference by means of two front-panel LEDs.

Key Features

- **Dual-Channel Performance with Interleave** – The AC240 offers dual-channel synchronicity for I/Q acquisitions with up to 1 GS/s sample rate. Interleaved single-channel mode up to 2 GS/s on either input is software selectable. The AC210 provides a single input channel with a sample rate of up to 1 GS/s.
- **On-board reconfigurable Data Processing Unit (DPU)** – The AC Series analyzer platform can easily be reconfigured to perform a variety of user-defined on-board real-time signal processing tasks on the digitized signals. The on-board FPGA is capable of executing multiplications in less than 5 ns and offers more than 74,000 logic cells, up to 7 Mbits of on-chip RAM, and 328 dedicated 18-bit x 18-bit multipliers with 36-bit results.
- **Optional external Processing Memory** – As an option, the AC Series can be equipped with additional memory to extend the processing capability of the DPU. The P512M Memory option offers two independent banks of 256 MB of DDR333-SDRAM and a 1 MB dual-port SRAM.
- **Front-panel I/O Connectors and Controls** for real-time data processing control (DPU Ctrl²) – It provides several front-panel connectors for real-time control of the DPU. The function of these connectors and LEDs is user defined through the implemented firmware.

3.1.1. Coupling & Impedance

Each channel has a 50 Ω signal input BNC connector giving high quality termination with better than ± 1% precision. It is ideally suited for use with 50 Ω transmission lines. Both AC and DC coupling modes are available. The AC mode couples signals capacitively thus removing the input signal's DC component and filtering out any signal component below 32 Hz. The DC mode allows all signal components to be passed through to the digitizer.

3.1.2. Input Protection

The input amplifiers are fully protected against over-voltage signals. Input signals up to ±5 V DC at 50 Ω, can be input without damaging the front-end electronics.

3.1.3. Mezzanine Front-end

The front-end electronics are all mounted on a removable mezzanine card. In the event of accidental damage, or as components fatigue over time (e.g. relays in high duty cycle automated testing applications), the mezzanine card allows for fast and efficient replacement.

3.1.4. Bandwidth and Rise Time

The bandwidth specification indicates the frequency at which an input signal will be attenuated by 3 dB (approximately 30% loss of amplitude). The bandwidth also affects the minimum rise and fall times that can be passed through the front-end electronics. A pulse with a very sharp edge will be observed to have a minimum rise time (τ_{\min}) determined by the front-end electronics. In general a pulse with a given 10-90% rise time ($\tau_{10-90\text{real}}$) will be observed with a slower value given by:

$$\tau_{10-90}^2 = \tau_{10-90\text{real}}^2 + \tau_{\min}^2 \quad \text{where} \quad \tau_{\min} (\text{ns}) \approx 0.35 (\text{GHz-ns}) / \text{BW} (\text{GHz})$$

If desired, a hardware bandwidth limiter at 20, 200 or 700 MHz can be selected.

Model	Agilent #	Bandwidth into 50 Ω	Minimum Rise Time
AC210, AC240 \approx U1080A		1GHz	0.35 ns

3.1.5. Input Voltage and Offset

The input channel provides a fully programmable amplifier with variable input voltage and offset. Full Scale (FS) input voltages are selectable from 50 mV to 5 V in a 1, 2, 5 sequence. Care should be taken to select an input voltage range that will allow the signal to be recorded using as much dynamic range of the digitizer as possible. The variable offset is programmable in the range of ± 2 V, in the FS ranges 50 to 500 mV, and ± 20 V, in the FS ranges 1 V to 5 V. The raw 8 bit ADC data values are in the range [-128,+127] with the first and last values reserved for underflow and overflow respectively. The midpoint value, 0, of the range corresponds to the negative of the offset voltage. Thus the Full Scale Range (FSR) goes from

$$-\text{Offset Voltage} - (\text{FS}/2) \quad \text{to} \quad -\text{Offset Voltage} + (\text{FS}/2)$$

Signals going outside of the FSR will be clipped and data values for the clipped portion of a signal should be regarded as erroneous.

3.1.6. Vertical Resolution

The AC Series Analyzers use an ADC system with 8 bits of vertical resolution (256 levels). The dynamic range of the ADC covers the Full Scale (FS) of the Input Voltage setting. For example, if the Input Voltage is set to 500 mV then the ADC resolution is equivalent to 1.95 mV. Analyzers use low noise front-end electronics in order to ensure a good signal to noise ratio. To obtain the best dynamic range from the ADC care should be taken to ensure that the input signal varies over more than 50% of the Input Voltage Full Scale (FS) setting. The highest and lowest levels of the ADC correspond to underflow and overflow conditions.

3.1.7. DC Accuracy

DC voltage accuracy is better than $\pm 2\%$ ($\pm 1\%$ typical) of the input voltage full-scale. The differential linearity is better than ± 0.9 LSB

3.2. Trigger

In normal digitizer operation, the trigger signal stops the data acquisition, typically somewhat delayed, depending on the user-specified *delay* value.

When the card is operated as a continuous data analyzer, the trigger signal is routed to the DPU where it is used as an indicator of the time region of interest. However, it does not stop the acquisition.

3.2.1. Trigger Source

The trigger source can be a signal applied to either the Input Channel (for internal triggering) or the External Trigger Input.

The modules provide a front panel BNC External Trigger Input. The External Input provides a fully functional trigger circuit with a fixed 50 Ω termination impedance, as well as selectable level and slope. It has the same BW limiter selections as can be found for the input channel. The user can choose the external trigger Full Scale from the set of values 0.5, 1.0, 2.0 or 5.0 V. A ± 5 V limit on trigger signals should be respected, although somewhat higher voltages for short time periods will not damage the unit.

3.2.2. Trigger Coupling

Trigger coupling is used to select the coupling mode applied to the input of the trigger circuitry. Modes available include AC LF Reject, HF Reject, and DC. The AC LF Reject mode couples signals capacitively and removes the input signal's DC component and frequencies below 50 Hz. DC mode allows all signal components to be passed through to the trigger circuit. The HF Reject mode removes signal components above 50 kHz.

3.2.3. Trigger Level

The trigger level specifies the voltage at which the selected trigger source will produce a valid trigger. The trigger level is defined as a set voltage. Using the internal trigger with DC coupling, the level is set with respect to the midpoint voltage ($V_m = -\text{Offset voltage}$) of the digitizer's vertical scale. Internal trigger level settings must be within

$V_m \pm 0.5$ FS, where FS is the channel or External Full Scale. All trigger circuits have sensitivity levels that must be exceeded in order for reliable triggering to occur. Trigger levels are also adjustable when using AC coupling.

AC Series digitizers will trigger on signals with a peak-peak amplitude $> 15\%$ FS from DC to their bandwidth limit.

3.2.4. Trigger Slope

The trigger slope defines the direction of the signal that will be used to initiate the acquisition when it passes through the specified trigger level. Positive slope indicates that the signal is transitioning from a lower voltage to a higher voltage. Negative slope indicates the signal is transitioning from a higher voltage to a lower voltage.

3.2.5. Window Trigger

The AC Series digitizers implement a Window trigger. Two trigger level thresholds are used to define the desired range. The trigger can then be chosen to occur either when the signal exits or enters the window range. This mode can be thought of as the appropriate OR of two edge triggers of opposite slope.

3.2.6. HF Trigger

The AC Series digitizers implement an HF trigger that allows triggers to be reliably accepted at rates above ~ 1 GHz. In this mode, triggers occur on every fourth positive edge. The window trigger is not available in this mode.

3.2.7. Pre- and Post-Trigger Delay

The pre- and post-trigger counters are only used in normal digitizer mode. When the card is operated as a continuous data analyzer, no acquisition trigger is required. The pre- and post-trigger counters are therefore ignored at this level.

To increase trigger flexibility a pre- or post-trigger delay can be applied to the trigger position.

The amount of pre-trigger delay can be adjusted between 0 and 100% of the acquisition time window (i.e. sampling interval x number of samples), whereas the post-trigger delay can be adjusted between 0 and 200 million samples.

Pre- or post-trigger delays are just different aspects of the same trigger positioning parameter:

- The condition of 100% pre-trigger indicates that all data points are acquired prior to the trigger, i.e. the trigger point is at the **end** of the acquired waveform.
- The condition of 0% pre-trigger (which is identical to a post-trigger of 0) indicates that all data points are acquired immediately after the trigger, i.e. the trigger point is at the **beginning** of the acquired waveform.
- The condition of a non-zero post-trigger delay indicates that the data points are acquired after the trigger occurs, at a time that corresponds to the post-trigger delay, i.e. the trigger point is **before** the acquired waveform.

The digitizer hardware accepts pre- and post-trigger adjustments in increments of 16 samples. By definition post-trigger settings are a positive number and pre-trigger settings are a negative number.

Thus it is only natural that the software drivers provided treat pre- and post-trigger delays as a single parameter in seconds that can vary between $-nbrSamples * samplingInterval$ (100% pre-trigger) and $+maxPostTrigSamples * samplingInterval$ (max post-trigger of 200M samples). Since the Acqiris software drivers provide very accurate trigger position information upon waveform readout, the accepted resolution of the user-requested pre-/post-trigger delay is much better than 16 samples. For more details, refer to the **Programmer's Reference Manual**.

3.2.8. Trigger Status

The front panel includes a tri-color LED indicator to show the status of the trigger for normal digitizer mode. This LED indicator is located near the TRIGGER IN BNC Connector.

When the LED is green it indicates the trigger is armed and waiting for a valid trigger to occur. Red indicates that the trigger has occurred, the acquisition is complete, and the data is waiting to be read out. The user can override the default functions and program the LED color in an application-specific manner.

When the AC Series are used as an analyzer this LED is always green when the acquisition is continuously running.

3.3. Sampling Rate

All Acqiris digitizers contain an analog-to-digital conversion (ADC) system that can sample waveforms, in a real time sampling mode, at rates from the maximum allowed rate down to 100 S/s (10 ms per point). The sampling rate can be programmed and is selectable in a 1, 2, 2.5, 4, 5 sequence (i.e. 1 MS/s, 2 MS/s, 2.5 MS/s, 4 MS/s, 5 MS/s, 10 MS/s, etc.). The maximum sampling rate shown above sometimes exploits the possibility of combining channels. The AC210 can sample up to 1 GS/s and the AC240 up to 2 GS/s. The data of all of the active channels is acquired

synchronously; all of the ADC's are acquiring data at the same time, to within a small fraction of the maximum sampling rate.

3.4. Data Acquisition - Digitizer Mode

Data from the ADC are stored in on-board acquisition memory. The amount of memory in use for acquisition can be programmed and is selectable from 2 points to 128 Kpoints, the full amount of acquisition memory available.

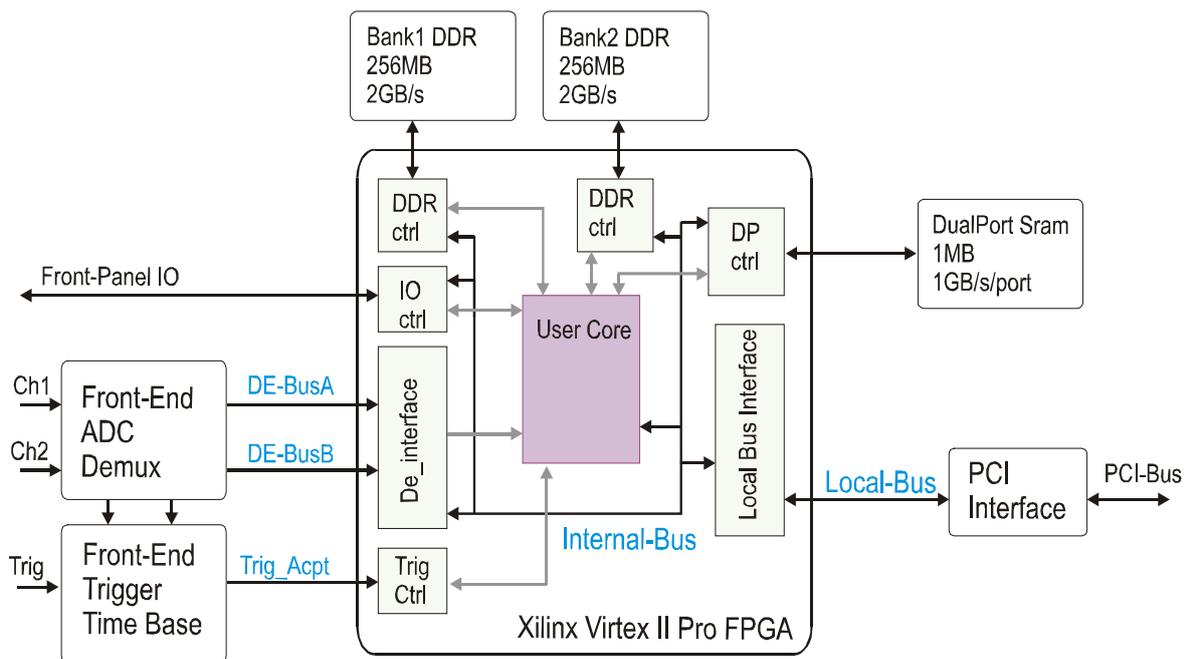
For technical reasons, a certain memory "overhead" is required for each waveform, reducing the available memory by a small amount. In order to simplify programming, an interface function recommends the best sampling rate and the maximum possible number of data points, taking into account the available memory, the requested time window, the number of segments (in Sequence mode), as well as the required memory overhead.

The Time Base Range defines the time period over which data is being acquired. For example, the AC210 in the digitizer mode has an acquisition memory of just under 128 Kpoints and maximum sampling rate of 1 GS/s. Thus, at the maximum sampling rate, the digitizer can record a signal over a time period of up to 130 μ s (128 Kpoints * 1 ns/point). The time base range can be adjusted by varying the amount of acquisition memory or the sampling rate of the digitizer.

3.5. Data Processing – Analyzer Mode

Data from the ADC are continuously streamed through the data demultiplexer (MAC) to the on-board Data Processing Unit. The Data Processing Unit of the AC Series is implemented as a large field-programmable gate array (FPGA) with optional external memory.

The figure below presents the Data Processing hardware environment and some of the basic blocks used to interface them.



3.5.1. Data Processing Unit

The Data Processing Unit is well suited for many data processing schemes. It is implemented as a powerful FPGA, the Xilinx Virtex II Pro XC2VP70-6FF1517 that provides up to two embedded PowerPCs, 328 18-bit x 18-bit multipliers, and 328 block RAMs.

The main features are summarized in the table below. Please refer to <http://www.xilinx.com> for the latest information about this device.

Resources	Qty	Description / Comment
Logic cells	66176	1 Logic cell has 1x (4 Input LUT + Flip-Flop + Carry Logic)
Block RAM	328	Instances of block RAM, 18 kbits each

Multiplier	328	18-bit x18-bit multipliers
DCM	8	Digital clock manager including frequency synthesis and phase shifting features. Frequency up to 420 MHz.
BUFG	16	Global Clock Buffer
Rocket IO	16	Not connected
PowerPC	2	No support from Acqiris

In addition, the P512MB Memory option can improve the processing capability.

The Analyzer functions are completely dependent on the firmware downloaded into the Data Processing Unit. The application-specific firmware available for the AC2x0 Series is described in chapter 4 **FIRMWARE**.

As an option, a FDK (Firmware Development Kit) for the Acqiris AC2x0 series enables users to develop and integrate user-specific data processing algorithms. The FDK reduces the development effort by providing a set of cores that interface to the underlying hardware resources. Please refer to the FDK Reference Manual for more information.

3.5.2. Memory Option

The external DPU memory option consists of:

- 2 banks of 256MB of DDR333 SDRAM with a throughput of up to 2 GB/s per bank.
- 1 MB of dual-port SRAM with a read/write throughput of up to 1 GB/s per port.

Each DDR bank is built around four 512Mbit DDR-SDRAM devices that are organized as 8M x 16bits x 4 banks. Data are synchronously transferred on a 64-bit wide bus on each edge of a 166.67 MHz clock. A DDR Controller that is able to sustain continuous burst reading or writing at a data rate of up to 2 GB/s is provided as part of the FDK.

The Dual Port Memory is built around two 4Mbit Dual Port SRAM organized as 128K x 36 bits. Data are synchronously transferred on a 64-bit wide bus on each port using a clock at up to 166.67 MHz. A Dual Port Controller that is able to sustain continuous simultaneous burst reading and writing at a data rate of up to 1 GB/s is provided as part of the FDK.

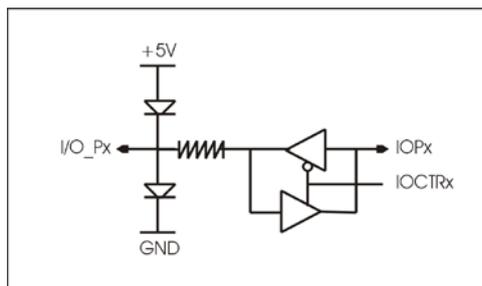
3.5.3. Extended Data Processing Controls

The AC Series provides several front-panel connectors for real-time control of the DPU. The function of these connectors and LEDs is user defined through the implemented firmware.

Two front-panel digital I/O MMCX-type connectors (I/O P1 & P2) are dedicated to the direct control of the data processing unit. These signals are 3.3 V compatible CMOS.

Each digital I/O can independently be configured either as an input or as an output. The figure below shows the equivalent schematic of one I/O Px interface.

The series resistor value is 50 Ω. I/O Px and I/O CTRx signals are connected to the Data Processing Unit.



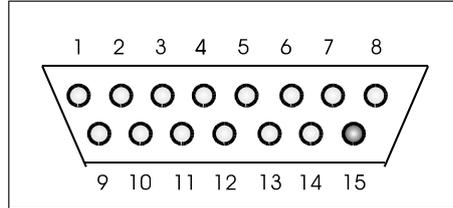
A third MMCX front-panel coaxial connector (ANL Out) is an analog output signal whose voltage is driven by a 16-bit on-board serial DAC. This analog signal can be used in simple control systems. The voltage range of that signal is -5V to +5V. The typical settling time (full scale range) is 1μs.

A front-panel μ DB-15 connector (I/O EXT) provides fourteen bi-directional direct lines to the DPU that can be used as seven differential pairs or as fourteen closely coupled single ended lines. Please note that these lines must use 2.5V signaling logic standards.



CAUTION: *Do not exceed the maximum input voltage rating! The maximum input voltage for μ DB-15 Connector is 2.6 V.*

The figure below shows the pinout of the μ DB-15 connector. Pin 15 is connected to the electrical ground.



The table below shows the pinout allocation of the I/O EXT connector. Each DP_{xn}/DP_{xp} pair refers to two lines routed towards the data processing unit as a differential pair.

Pin	Allocation
1 - 2	DP _{6n} – DP _{6p}
3 - 4	DP _{5p} – DP _{5n}
5 - 6	DP _{4p} – DP _{4n}
7 - 8	DP _{0p} – DP _{0n}
9-10	DP _{3p} – DP _{3n}
11-12	DP _{2p} – DP _{2n}
13-14	DP _{1p} – DP _{1n}
15	GND

Finally, two LEDs (L1 & L2) provide a visual reference. Each LED is independently driven by the Data Processing Unit and displays one of the four following colors (Black or switched off, Red, Green, or Orange).

3.6. External Clock and Reference

For applications where the user wants to replace the internal clock of the digitizer in order to drive the ADC with an external source, an External Clock or Reference signal input is available. The Clock or Reference signals can be entered into the digitizer via the MMCX CLK IN connector on the front panel.

When using an External Clock, the user must ensure that the input signal has a frequency between 10 MHz and 2 GHz, and a minimum amplitude of at least 1 V, peak to peak, into 50 Ω . The External Clock allows the digitizer to make a voltage measurement whenever the clock signal passes through a predefined threshold. However, it should be noted that when 2 channels are being used the maximum Sampling Rate is half of the External Clock Frequency in the Continuous mode and in this case the Start/Stop mode is to be preferred. The threshold range is variable and user selectable between ± 2 V. The signals should not exceed ± 5 V amplitude.

For applications that require greater timing precision and stability than is obtainable from the internal clock, a 10 MHz Reference signal can be used. The amplitude and threshold conditions, for an External Reference, are the same as for the External Clock. If phase synchronization between several digitizers is required, the reference signal should be applied to all of them.

3.7. Internal Calibration

The software drivers supplied include internal calibration functions for the timing, gain, and offset settings, which can be executed upon user request. The digitizers are never calibrated in an “automatic” way, i.e. as a side effect of another operation. This ensures that programmers have full control of all calibrations performed through software in order to maintain proper event synchronization within automated test applications.

The AC2x0 includes a high precision voltage source and a 16-bit DAC, used to determine the input voltage and offset calibration.

For accurate time and voltage measurements it is recommended to perform a calibration once the module has attained a stable operating temperature (usually reached with a few minutes after power on). Further calibration should not be necessary unless temperature variations occur. Calibration can usually be performed with signals present at the channel, external, and clock inputs. However, if the calibration is found to be unreliable, as shown by a calibration failure status, it may be necessary to remove such signals.

3.8. Factory Calibration

Whilst the Internal Calibration function will provide a good degree of confidence that your instrument is operating within its specifications on a day-to-day basis, Agilent recommends that each instrument undergoes a Factory Calibration at least annually to ensure that it remains within the specified performances.

Factory Calibration is the process of measuring the actual performance of an instrument-under-test using lab instruments that in turn have significantly better performance than the instrument-under-test. Lab instrument performance must be traceable to the International System (SI) Units via a national metrology institute (NIST, NPL, NRC, PTB, CENAM, INMETRO, BIPM, etc.)

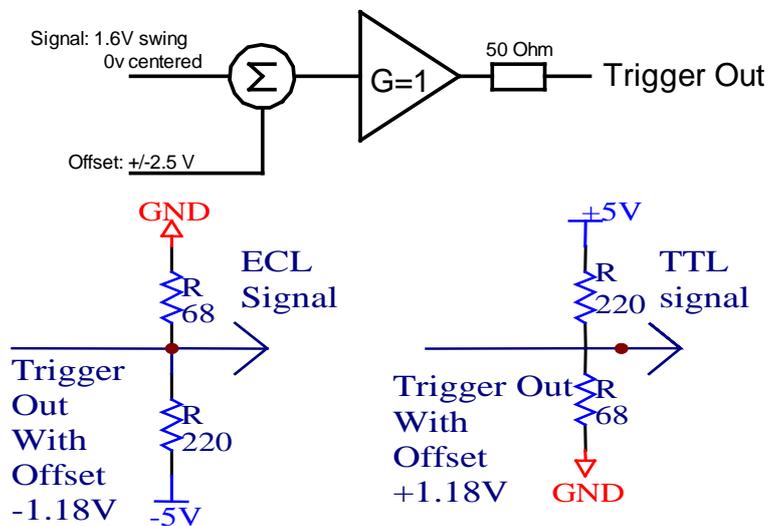
The measured performance is then compared to published data sheet specifications. Agilent Technologies tests the performance corresponding to all data sheet specifications, for all installed options, every time. If we observe your instrument to be out-of-specification, we always adjust it and re-test before we return it to you.

For more information, or to return your instrument to Agilent for calibration please see www.agilent.com/find/calibration

3.9. External Trigger Output

When the module is ready to be triggered and a valid trigger signal occurs, a trigger output is generated for external use. It is always available on the Front Panel Trigger Out MMCX connector. The pulse ends when the data acquisition for the trigger in question is complete.

Trigger Output Block diagram:



The output swing is 1.6 V (± 0.8 V) when unloaded and 0.8 V when terminated on 50 Ω . The rise and fall times are 2.5 ns typical. The offset can be adjusted, by software control in the range [-2.5 V, +2.5 V] unloaded, or [-1.25 V, +1.25 V] into 50 Ω . The maximum output current capability is ± 15 mA. As the output is retro-terminated, it is possible to drive a 50 Ω line unterminated (HiZ) without loss of performance.

For a TTL compatible signal, set the offset to 1.0 V and the swing at destination will be +0.2 to +1.8 V.

For an ECL compatible signal, terminated on 50 Ω to -1.2 V, set the offset to -1.2 V and the output will be in the range [-0.8 V, -1.6 V].

Alternatively, to reduce the current drawn from the digitizer, the terminations shown here can be used:

The trigger out signal can also be routed to the PXI Bus Star Trigger line.

3.10. AC210 Front Panel Inputs and Controls

Name	Generic Function	Comments	Connector
L1	LED Status (*)	Firmware Dependent	LED L1
L2	LED Status (*)	Firmware Dependent	LED L2
ANL Out	Analog Output (*)	Firmware Dependent	MMCX ANL OUT
I/O P1	Input/ Output (*)	Firmware Dependent	MMCX I/O P1
I/O P2	Input/ Output (*)	Firmware Dependent	MMCX I/O P2
I/O Ext	7 differential lines for remote control or 14 closely coupled single ended lines (*)	Firmware Dependent	I/O EXT
INPUT	Signal input	Signal input (Channel 0)	BNC INPUT
CLK IN	Reference clock	50 Ω Input for external clocking	MMCX TR
I/O A	User configurable	–	MMCX I/O A
TRIGOUT	Signal occurs after an accepted TRIGGER. It is synchronous to the acquisition Clock and can be used to trigger events synchronously to the acquisition clock.	Also available when using the AC Series Module as a standard digitizer.	MMCX TRIG OUT
I/O B	User configurable	–	MMCX I/O B
TRIGGER IN	Trigger input	Trigger Input	BNC TRIGGER IN
AS bus	Auto Synchronous Bus System	–	AS bus
Digitizer Status LED	Acquisition Status	Green when data is streamed to DPU	LED next to TRIG OUT

(*) See 3.5.3 **EXTENDED DATA PROCESSING CONTROLS**

The I/O A, I/O B signals are 3.3 V compatible CMOS. This means that, on input, low must be < 0.7 V and high must be in the range [1.7 V, 5.0 V]. An unconnected signal will be high. This definition ensures TTL compatibility. On output, the low level will be in the range [0 V, 0.7 V] and the high level in the range [1.7 V, 3.3 V] for HiZ. The high level output will typically generate 0.8 V into 50 Ω .

For firmware-specific inputs and controls, please refer to the corresponding sections within the firmware description section



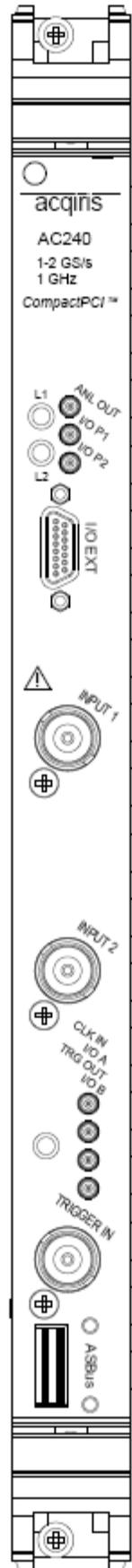
3.11. AC240 Front Panel Inputs and Controls

Name	Generic Function	Comments	Connector
L1	LED Status (*)	Firmware Dependent	LED L1
L2	LED Status (*)	Firmware Dependent	LED L2
ANL Out	Analog Output (*)	Firmware Dependent	MMCX ANL OUT
I/O P1	Input/ Output (*)	Firmware Dependent	MMCX I/O P1
I/O P2	Input/ Output (*)	Firmware Dependent	MMCX I/O P2
I/O Ext	7 differential lines for remote control or 14 closely coupled single ended lines (*)	Firmware Dependent	I/O EXT
INPUT 1	Signal input	Signal input (Channel 1)	BNC INPUT 1
INPUT 2	Signal input	Signal input (Channel 0)	BNC INPUT 2
CLK IN	Reference clock	Input for external clocking	MMCX TR
I/O A	User configurable	–	MMCX I/O A
TRIGOUT	Signal occurs after an accepted TRIGGER. It is synchronous to the acquisition Clock and can be used to trigger events synchronously to the acquisition clock.	Also available when using the AC Series Module as a standard digitizer.	MMCX TRIG OUT
I/O B	User configurable	–	MMCX I/O B
TRIGGER IN	Trigger input	Trigger Input	BNC TRIGGER IN
AS bus	Auto Synchronous Bus System	–	AS bus
Digitizer Status LED	Acquisition Status	Green when data is streamed to DPU	LED next to TRIG OUT

(*) See 3.5.3 **EXTENDED DATA PROCESSING CONTROLS**

The I/O A, I/O B signals are 3.3 V compatible CMOS. This means that, on input, low must be < 0.7 V and high must be in the range [1.7 V, 5.0 V]. An unconnected signal will be high. This definition ensures TTL compatibility. On output, the low level will be in the range [0 V, 0.7 V] and the high level in the range [1.7 V, 3.3 V] for HiZ. The high level output will typically generate 0.8 V into 50 Ω.

For firmware-specific inputs and controls, please refer to the corresponding sections within the firmware description section.



3.12. Physical Specifications

3.12.1. Electrical

			CURRENT REQUIREMENTS (A)			
Model	Firmware	Max. Power Consumption (W)	+12V	+5V	+3.3V	-12 V
AC210	Base Test	22.3	0.7	0.8	2.2	0.05
AC240	Base Test	22.3	0.7	0.8	2.2	0.05
AC240	FFT Spectrometer	42.5	0.8	1.2	6.8	0.05

The Maximum Power Consumption has been increased by 10% over the value calculated with the currents shown to take into account higher allowed values of the crate voltages.

These modules use the PCI Bus at 33 MHz and are compatible for either V I/O = 3.3 V or 5 V. All of these modules are capable of DMA transfers at rates ~100 MB/s.

3.12.2. Environmental and Physical

Operating Temperature

0° to 40°C

The above values are for the ambient temperature of the room (or equivalent) where the AC210/AC240 is located. The temperature as measured on the board may well be significantly higher. On-board temperatures above 60°C should be avoided.

Relative Humidity

5 to 95% (non-condensing)

Dimensions

All AC modules conform to the CompactPCI standard and have a 6U form factor. (233 mm × 160 mm × 20 mm).

EMC Immunity & Emissions

Complies with European EMC Directive 2004/108/EC

- IEC/EN 61326-2-1
- CISPR Pub 11 Group 1, class A
- AS/NZS CISPR 11
- ICES/NMB-001

This ISM device complies with Canadian ICES-001.

Cet appareil ISM est conforme à la norme NMB-001 du Canada.

Required Airflow

> 2 m/s in situ

4. Firmware

The following sections describe the major elements of firmware supplied by Agilent Acqiris, as standard or as an option.

The term 'firmware' refers to the FPGA program. It is contained in a file with the extension '.bit' that must be downloaded by the software driver through the Compact PCI bus to configure the AC240/AC210 Module for a specific operating mode.

Once configured and started through the Compact PCI Bus, the firmware typically needs little or no interaction with the controlling PC, nor further support from the software driver, until explicitly stopped with a software command. In some cases, the user may continuously monitor the data streaming process by capturing some data blocks on the fly, without lowering the data throughput.

This table shows the list of the available firmware for the AC Series.

Firmware Name	Description	Firmware file	Order Information
BASE TEST Firmware - Default FPGA configuration files for initial loading and diagnostic tests	Base Test for AC210	AC210.bit	Standard
	Base Test for AC240	AC240.bit AC240mem.bit	Standard
FFT SPECTROMETER Firmware	32 Kpoint FFT Analyzer firmware for AC240, at 2 GS/s	AC240FFT2GSs.bit	AC240-FFT2G32K-OPT

NOTE: Base Test firmware is always delivered with an AC module. The VHDL source code is only provided with the corresponding FDK option.

Other firmware may be protected by a permission code which prevents their use on unauthorized AC2x0 modules. Protected firmware can only be used on AC2x0 modules that contain the corresponding permission code for the firmware option.

4.1. Base Test Firmware

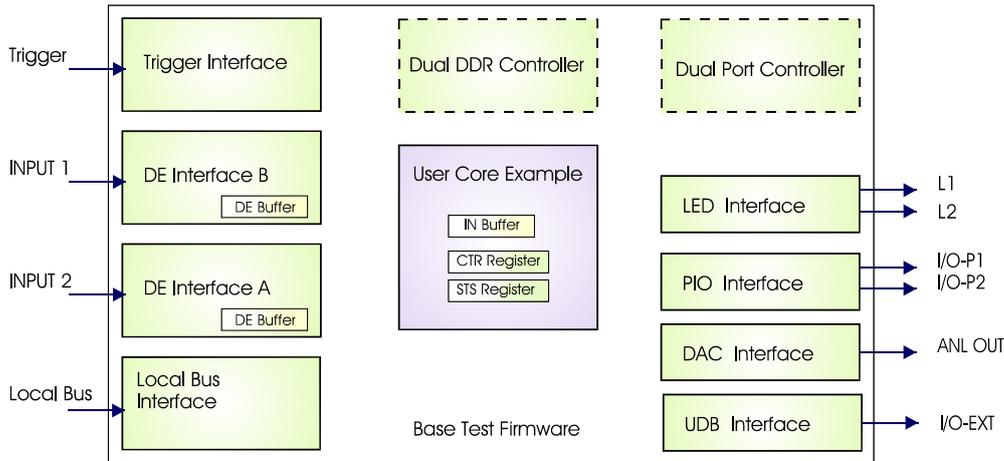
The Base Test firmware is designed to run on any Acqiris AC2x0 or SC2x0 Analyzer platform card. The Base Test firmware is built from a base design database whose purpose is twofold:

- “Design framework”. It is meant to be used as a starting point for any new developments based on the FDK
- Test and Demonstration. It can be used for test/demonstration purposes of either the AC2x0 underlying hardware or the developed firmware itself.

There is one Base Test for the AC210 module and two Base Tests for the AC240 module. The P512MB memory option is only supported for the AC240.

4.1.1. Architecture of Base Test Firmware

The following picture presents the simplified architecture of the Base Test firmware for the AC240 Module.



The Base Design is a collection of Agilent Acqiris-supplied cores (filled in green on the block diagram) that interface with the underlying hardware of the module.

The *Local Bus Interface* core enables the communication between FPGA registers and buffers and the board local bus connected to the PCI bridge. The *DE interface* core provides an interface to the data entry bus to retrieve the demultiplexed data samples. It contains an 8K sample FIFO, named DE buffer, which can be read back through the *Local Bus* interface. There should be one *DE interface* core per acquisition channel used. The *Trigger interface* core is used to place the trigger information within the incoming data stream. The *LED Interface* core controls the color of both L1 and L2 LEDs. The *PIO interface* core is used to manage both I/O-P1 and I/O-P2 MMCX. The *DAC interface* core provides control of the 16-bit DAC connected to the ANL-OUT MMCX. The *UDB Interface* core provides a way to test the μ DB connector. Finally, the *Dual DDR Controller* core and *Dual Port Controller* core can be used to store and read samples or data of interest, provided that the P512MB option is implemented on the board.

The Base Design contains a *User Core* example that implements two registers and one 8K sample buffer named DE-Monitor Buffer. It provides a minimum set of functions to reduce the developer's work when starting a new design and to perform tests on the hardware resources. While using the Analyzer Demo application, the user can capture a waveform on both channels at two levels (in the DE buffers and in the DE-Monitor Buffer) and access the various resources (e.g. LED, PIO, DAC etc...) through the core registers. In addition, it implements some basic tests to check the health of the AC module.

4.1.2. I/O and Controls of Base Test Firmware

The table below presents the functionality of the I/O controls located on the front panel.

Item	Generic Function	Comments	Connector
L1	LED Status	Software LED	LED L1
L2	LED Status	DE-Monitor Buffer Full	LED L2
ANL Out	Analog Output	Driven by DAC core	MMCX ANL OUT
I/O P1	Input/ Output	Driven by FPIO core / Used for Debug	MMCX I/O P1
I/O P2	Input/ Output	Driven by FPIO core / Used for Debug	MMCX I/O P2
I/O Ext	7 differential lines for remote control or 14 closely coupled single ended lines (*)	Used to Test μ DB link, provided specific loop hardware is used.	I/O EXT

The LED colors can be overwritten by the software using the FPGA front-panel LED control register. The ANL-Out value can only be driven by using the FPGA front-panel DAC control register. I/O-P1 and I/O-P2 values are configured as outputs and can multiplex several signals of interest using the FPGA front panel PIO control register. Finally, the I/O Ext connector is only used for production tests with this firmware.

4.2. FFT Spectrometer Firmware

The FFT Spectrometer firmware option implements a real-time Fast Fourier Transform, capable of transforming 32K time-domain data points to 16K spectral lines at 2 GS/s or lower sampling rate, on a **single channel**, without any dead time. It computes the power spectrum and is able to accumulate power spectra in order to obtain a time-averaged power spectrum.

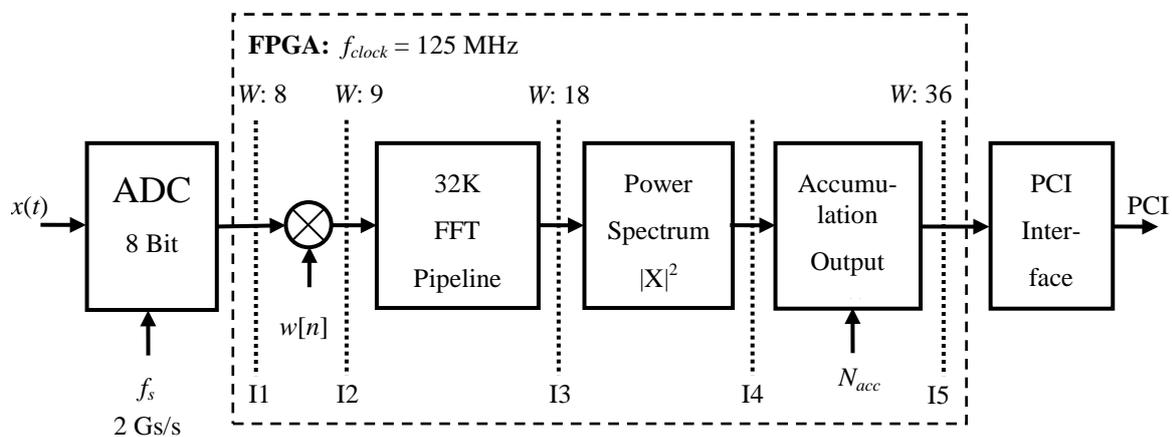
It is designed for the AC240 only. Its primary applications are in radio astronomy and atmospheric physics. It was developed in collaboration with the Astronomy Department of the Swiss Federal Institute of Technology (ETH) in Zürich and the Technical Universities of Aargau and Solothurn, under a grant by the Swiss government.

Key Features

- **High Spectral Resolution** – FFT transforms data blocks of 32,768 values into a power spectrum of 16,384 spectral lines
- **High Speed** – FFT on 32,768 data values is computed in 16.384 μs (at 2 GS/s)
- **Multiple Programmable Sampling Rates** – 2 GS/s, 1 GS/s, 500, and 400 MS/s, or user-defined sampling rates between 400 and 2000 MS/s with an external clock
- **High Bandwidth** of up to 1 GHz (at 2 GS/s), with a resolution of 61 kHz per spectral line
- **No Dead Time** between data blocks
- **Dual Accumulation Buffer** for power spectra, permitting dead-time-less readout of a sum of power spectra while accumulating into the other buffer
- **High Number of Accumulations** – More than 1,000,000 power spectra can be accumulated, only limited by the size of the signal
- **Front Panel Control Signals** – For easy integration of the spectrometer into a variety of systems, a number of programmable front panel control lines are available. Two front panel LEDs show important information on the current status.

4.2.1. Architecture of FFT Spectrometer Firmware

The picture below illustrates the structure of the firmware.



The ADC is shown as a single 8-bit, 2 GS/s converter. In reality, it is implemented with two 1 GS/s ADCs, sampling with a time shift of 500 ps with respect to each other. The 2 ADCs are matched in amplitude, offset, and relative delay through calibration that can be requested with a software command.

The converted data are continuously transferred from the ADC into an input buffer of 32K data points (not shown). Whenever the buffer is full, data are automatically sent to a second, parallel input buffer while the first one is used as a source to the FFT computation. The FFT is computed on the first buffer in a time that is (slightly) shorter than the accumulation time into the second buffer. Thus, when the 2nd buffer is full, the ADC data are sent again to the first buffer, while the FFT firmware is already available for computation on the second buffer. The result is a seamless acquisition of blocks of 32K data points, and their conversion into frequency spectra.

Each 32K point data block is first multiplied by a windowing function $w[n]$ that can be pre-loaded under program control. By default, this function is 1 everywhere, representing a rectangular window. The 32K points are then transformed by the FFT firmware into 16K spectral lines, represented as complex numbers. Both the real and

imaginary parts have 18-bit resolution. The power spectrum is computed by adding the square of the real part to the square of the imaginary part, represented as a 36-bit integer.

Finally, the power spectrum is added to one of the 2 accumulation buffers. This operation is repeated until the programmable number N_{acc} of power spectra is accumulated. At this point, the firmware automatically starts accumulating into the second accumulation buffer and sets a 'buffer full' bit in a status register. In addition it may also generate an interrupt to the host computer.

The host computer is expected to read the accumulation buffer fast enough to liberate it before the other buffer is full. Otherwise, the system cannot switch back to the first accumulation buffer, in which case data may be lost. For this reason, it is recommended to choose N_{acc} large enough. Reading a full accumulated spectrum requires the transfer of 16K spectral lines of 8 bytes each (when reading all 36 bits), requiring approximately 1.5 ms. However, the real amount of time taken to read the buffer depends on the host computer and the operating system. To avoid data loss, it is better to choose accumulation periods of the order of 100 ms or more.

4.2.2. I/O and Controls of FFT Spectrometer Firmware

The table below shows the functionality of the I/O controls located on the front panel.

Item	Generic Function	Comments	Connector
L1	LED Status	Turns red on ADC overflow	LED L1
L2	LED Status	Turns red on Accumulator overflow	LED L2
ANL Out	Analog Output	Not Used	MMCX ANL OUT
I/O P1	Input/ Output	Not Used	MMCX I/O P1
I/O P2	Input/ Output	Not Used	MMCX I/O P2
I/O Ext	14 single ended I/O signals	6 inputs, 8 outputs	I/O EXT

The signals on the I/O Ext μ DB connector are defined as 6 inputs and 8 outputs. One input is defined as 'FFT enable' and one output is defined as 'buffer full'. The other signals can be accessed through the Front-Panel μ DB IO Control Register.

The exact relationship between this register and the physical contacts on the I/O Ext μ DB connector is described below:

Pin	Signal Name	Control Bit in I/O Register	Use
1	DP6_p	0	FFT Enable Input
2	DP6_n	1	Input
3	DP5_p	2	Input
4	DP5_n	3	Input
5	DP4_p	4	Input
6	DP4_n	5	Input
7	DP0_p	8	Output
8	DP0_n	9	Output
9	DP3_p	10	Output
10	DP3_n	11	Output
11	DP2_p	12	Output
12	DP2_n	13	Output
13	DP1_p	14	Output
14	DP1_n	15	Buffer Full Output
15	GND	-	Ground

The physical position of the I/O pins was shown in section 3.5.3 **EXTENDED DATA PROCESSING CONTROLS**.

5. Running the Analyzer Demo Application

The **Analyzer Demo** application is an interactive program running under Windows. It permits the operation of some Agilent Acqiris-supplied applications on the AC2x0 Analyzers and on the SC2x0 Streaming Analyzers.

5.1. Getting Started with Analyzer Demo

Once the software and hardware installation described in Section 2 of this manual is complete, you can start **Analyzer Demo** from the start menu of your computer.

During startup, **Analyzer Demo** searches for all Acqiris Digitizers/Analyzers on the PCI/CompactPCI bus. If none are found, **Analyzer Demo** will display an error message indicating this fact and automatically switch to simulation mode with three different simulated instruments. In such a case the solution may be to turn off the computer, install and turn ON the hardware on the CompactPCI bus, and then restart the computer.

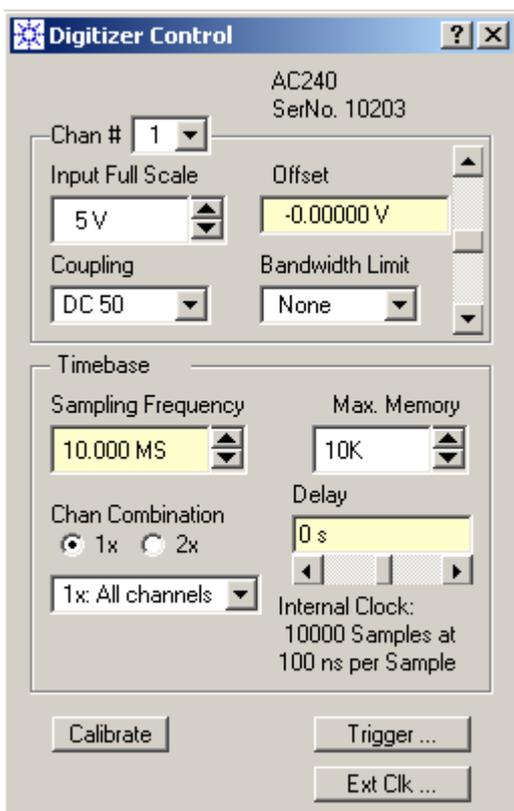
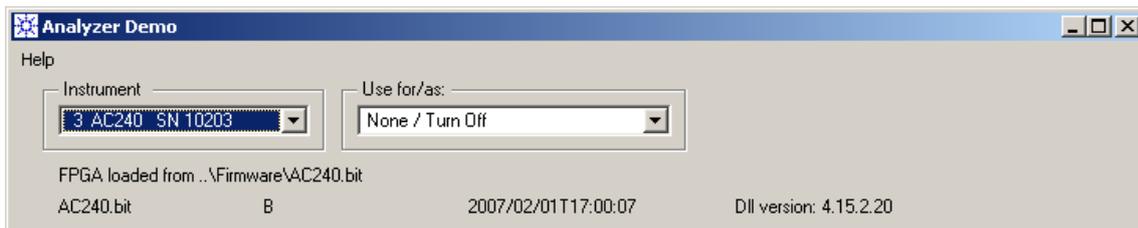
When you start **Analyzer Demo**, it displays three independent windows, an instrument control window, a digitizer control window, and an application window. If several instruments are present, there may be an application window for each of them.

5.2. Instrument Control Window

The instrument control window is common to all available instruments. The drop-down list on the left-hand side allows the selection of the *current* instrument, if several are available.

The drop-down list on the right-hand side “Use for/as” allows choosing an application for the *current* instrument. The application window below automatically opens/adapts and the appropriate FPGA configuration file is automatically loaded. On AC2x0 Analyzers, at least the **Base Test** application is present, as described further below. Other applications might be available, depending on the options installed in the card.

The digitizer control window always acts on the *current* instrument, see next section.



At the bottom of the instrument control window, the name of the loaded FPGA configuration file is shown, together with its directory path. A few additional pieces of information, such as the firmware name and version, are also shown (subject to change).

5.3. Digitizer Control Window

The Digitizer Control panel provides all the features needed to use the AC240/210 module in the Digitizer mode. It allows setting up the appropriate instrument configuration for the signal(s) applied to the front panel input(s), e.g. input coupling, full scale, and sampling rate.

The Digitizer Control panel always acts on the currently selected instrument in the instrument control window.

The model and serial numbers of the currently controlled instrument are shown at the top right of the control window.

If several channels are available in the instrument, the **Chan #** field shows which channel is currently being displayed and/or modified.

The adjustment of the digitizer parameters is best done while running in the Acquisition mode **Auto** in one of the application windows (see next section). Any modifications will then immediately be visible in the waveform display.

The **Offset** can be modified either by selecting the display of the offset field and typing the desired value followed by a <CR>(Enter), or by clicking on the various fields in the scroll bar. It is also possible to drag the scrollbar.

The **Max Memory** field limits the amount of memory being used to acquire a waveform during the setup phase. 5K, or 10K, points is a reasonable amount to obtain a fast display update rate.

The **Chan Combination** field is of interest for the AC240 only. When **1x** is selected, both inputs are converted independently as 2 separate channels. Clicking on **2x** switches the instrument to the mode where the 2 ADCs are interleaved on a single channel, allowing sampling rates up to 2 GS/s. The drop-down box just underneath the Chan Combination field gives a choice of whether the signal at input 1 or 2 is converted.

The **Delay** can be controlled in the same way as the **Offset**. However, it will be ignored when running in the continuous acquisition mode.

The field at the right bottom corner of the Timebase section indicates the settings of the timebase parameters: clock source (internal or external), the number of samples, and the sampling interval. The number of samples is only of interest for the **Single** and **Auto** acquisition modes; it will be ignored when running in the **Continuous** mode.

The pushbutton **Calibrate** requests an internal recalibration of the current instrument. The buttons **Trigger...** and **Ext Clk...** open additional dialog boxes for the setup of the trigger conditions and the selection of an External Clock option.

5.4. Application Windows & Acquisition Mode

The Base Test application will automatically be opened on the first instrument and **Base Test** will be displayed in the Instrument Control Window. If additional instruments are present, they can be activated by selecting another instrument in the left-hand drop-down list of the Instrument Control Window and selecting the appropriate “Use for/as” value.

You may switch from one application to another at any time, with the “Use for/as” drop-down list. There is a time lag of a couple of seconds because a new firmware file is automatically loaded into the FPGA.

The **Acquisition** control field is present both in the **Base Test** (if the Test Category is set to *Data Transfers*) and other optional applications. Thus, it is described in this common section.

The acquisition mode of the analyzer in its digitizer mode is selected using one of four acquisition mode buttons in the **Acquisition** section of the control panel. Available acquisition modes are Stop, Single, Auto, and Continuous. The use of each of the acquisition modes is described below. The Single and Auto modes require a valid trigger, i.e. a trigger signal meeting the trigger conditions at a time when the digitizer is armed and ready to acquire data.

- **Stop** will stop the acquisition and hold the latest complete acquisition on the display.
- **Single** mode is used in order to capture one event at the first valid trigger. It freezes the acquisition in the digitizer’s memory, and on the display, until the user requests another acquisition. After an acquisition is taken in Single mode, the digitizer will ignore subsequent trigger events until the Single button is pressed again or another acquisition mode is selected. Pressing the single button re-arms the trigger and captures a new acquisition.
- **Auto** mode will acquire and display waveforms according to the trigger settings if a valid trigger is present within a timeout interval. If a valid trigger is not available within this interval, the digitizer generates its own trigger in order to digitize and display whatever signal is at the input at that time. Auto mode is typically used to aid in setup when the input signal must be quickly characterized in order to determine proper vertical and trigger settings for Continuous or Single mode acquisitions.
- **Continuous** mode is used to continuously transfer converted data to the on-board DPU (FPGA). In this mode, some digitizer settings such as trigger or delay are ignored. The operation continues until another mode is explicitly selected. This mode must be chosen for any operations that are implemented in the FPGA.

5.5. Base Test Application

The operation of the Base Test is selected by choosing “Base Test” in the “Use for/as” field of the Instrument Control Window.

The Base Test application offers a number of tests for the on-board FPGA of an Analyzer or Streaming Analyzer instrument.

There are 3 test categories available:

- **Registers** tests the connection between the host computer and the FPGA, as well as the connection between the FPGA and the front panel LEDs
- **Data Transfers** tests the connection between the ADC(s) and the FPGA
- **Memory** tests the (optional) dual port static memory and the 2 dynamic memory banks on the AC240. AC210s with memory option are currently not supported by AcqirisAnalyzer.

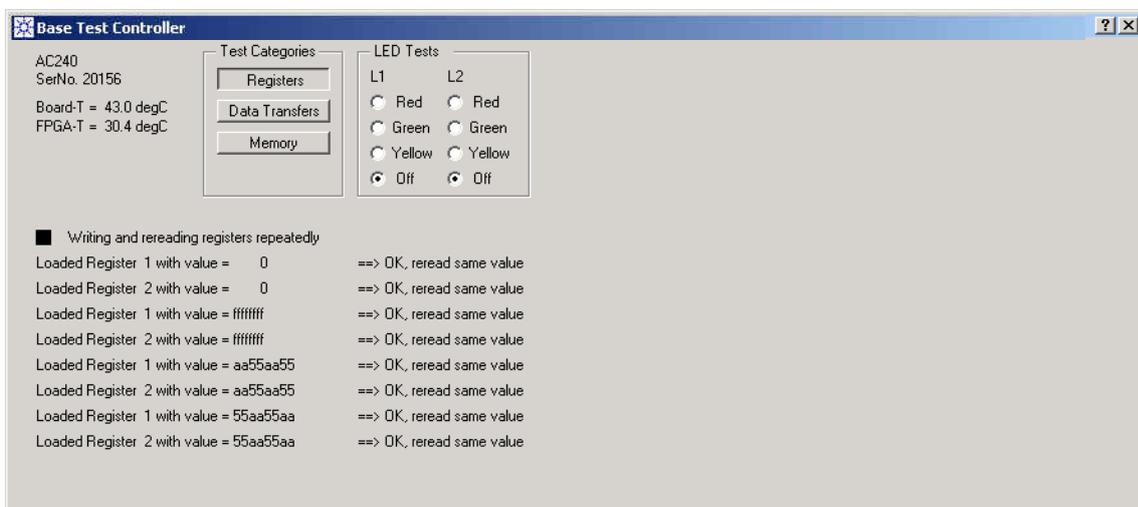
5.5.1. Register Tests

This test writes a number of different data values to a few registers, reads them back, and compares them with the originally written values. The result of the comparison is reported on the screen.

The test is automatically repeated once a second, as indicated by a ‘blinking’ square next to the text ‘Writing and rereading registers repeatedly’.

Since the User registers 1 and 2 are present in all AC2x0.bit and SC2x0.bit default firmware files, this test can be executed for all (streaming) analyzer instruments.

Successful tests show “= => OK, reread same value” while a failed tests displays “= => Error, reread xx”.

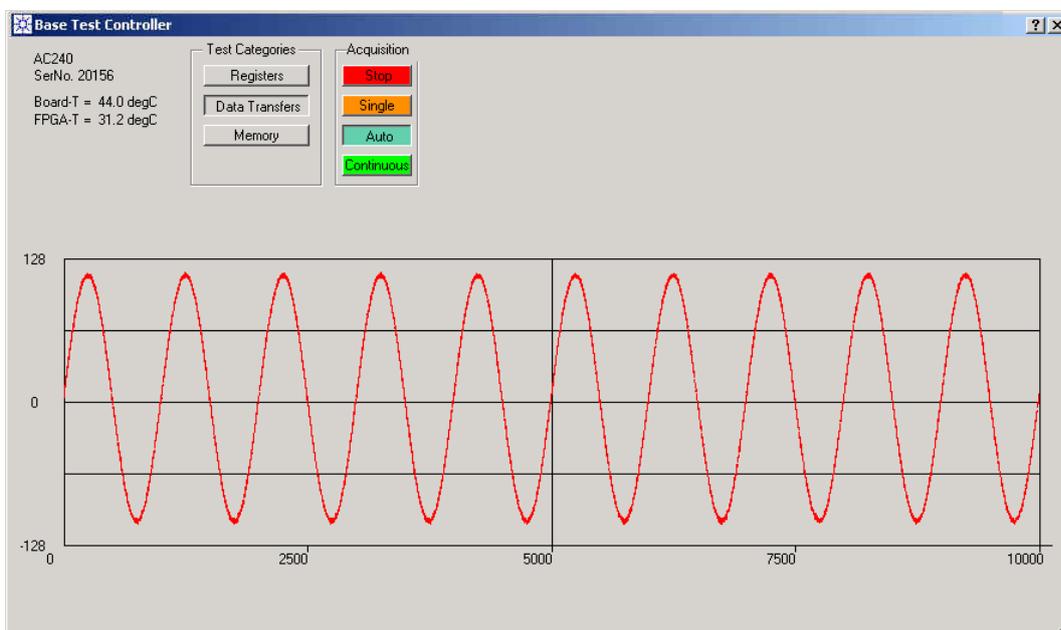


In addition, the control of the front panel LEDs by the FPGA can be tested by pushing the radio buttons in the **LED Tests** section. The LEDs L1 and L2 should show the requested color.

5.5.2. Data Transfer Tests

This test verifies that ADC data arrive correctly to the FPGA input buffers.

First use the **Auto** mode to set the digitizer to an appropriate state for acquiring your input signal.



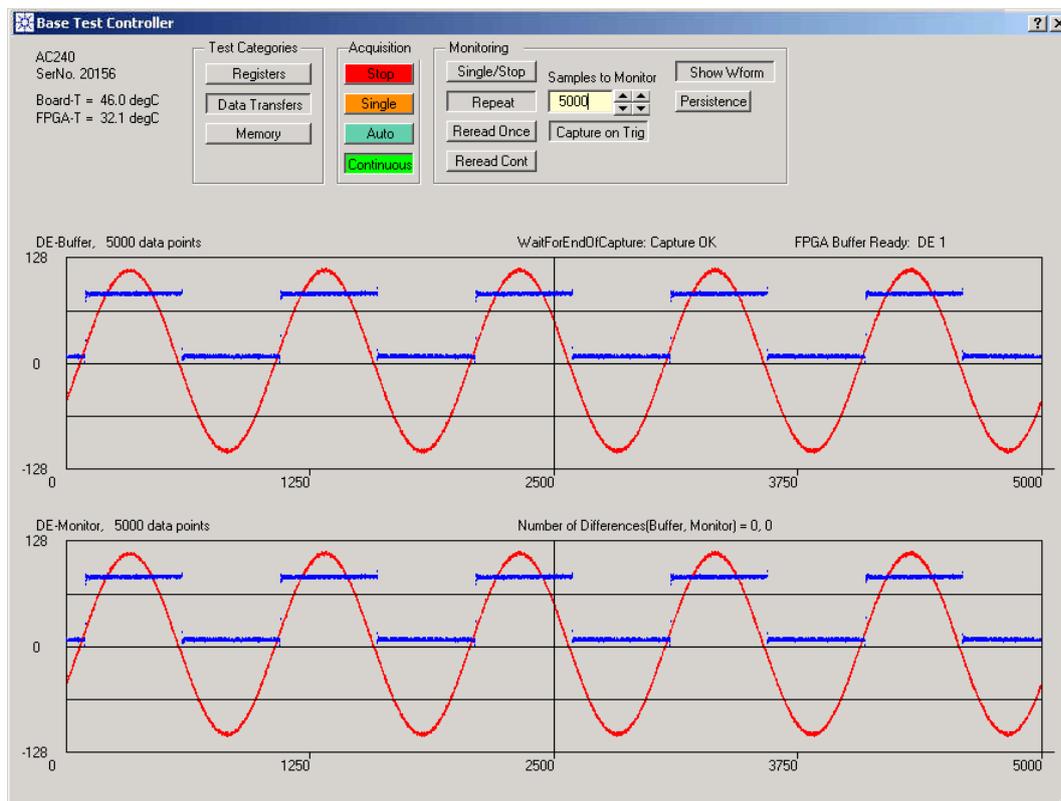
The **Auto** mode does not test the connection to the FPGA, since its waveforms are acquired in the digitizer memory, as opposed to through the FPGA.

For the real test, switch to the **Continuous** mode. A **Monitoring** section appears. The digitizer transfers data continuously to the FPGA, independently of the choices in this section. Select a number of samples to monitor, typically between 1000 and 10000 samples.

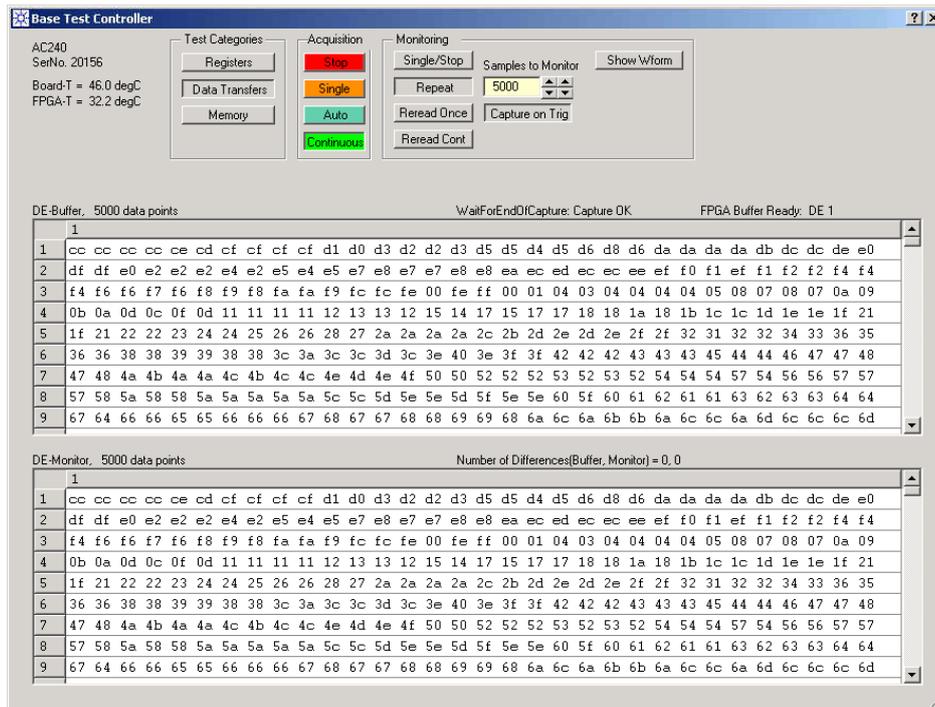
- The **Single/Stop** pushbutton is used to capture a single set of waveforms from the FPGA. It then displays the captured waveforms and freezes the display. It can be pushed again for another capture sequence.
- The **Repeat** pushbutton repeats the capture/display sequence indefinitely until the **Single/Stop** pushbutton is depressed.
- The **Reread Once** and **Reread Cont** buttons are for advanced failure diagnostics. They should be ignored unless asked for by the Acqiris Support team.
- The **Capture on Trig** pushbutton directs the monitoring process to capture data only upon the occurrence of a trigger signal at the FPGA input. This mode works when valid trigger conditions for a signal at the Channel input or the External Trigger input have been defined.

The figure below shows both waveforms of an AC240. The red curve is Channel 1 and the blue one is Channel 2. The upper display shows the waveforms as captured in the Data Entry (DE) section of the FPGA, while the lower one displays a secondary monitoring buffer that monitors the data stream at the Input of the user Core. A text line above the lower waveform display indicates the number(s) of differences between the DE-buffer and the DE-Monitor buffer, for each channel. This should always be zero; a non-zero value indicates that the data have not been successfully transferred.

When displaying the waveforms graphically (i.e. with **Show Wform** depressed), the user has the option of selecting **Persistence**. This mode permits the accumulation of many waveforms on the display in the search for intermittent transfer problems. When using persistence display, you must use the **Capture on Trig** in order to stabilize the horizontal position of the waveform.



For a detailed analysis of an observed data transfer problem, the **Show Wform** button can be released as shown in the following figure, to display the data in hexadecimal format. When 2 channels are available, the numeric data of Channel 1 are displayed first, followed by those of Channel 2.



5.5.3. Memory Tests

The memory tests verify the operation of the (optional) external memories of the FPGA. There are, 2 dynamic memory banks, DDRA and DDRB, and one dual port static memory bank, SRAM.

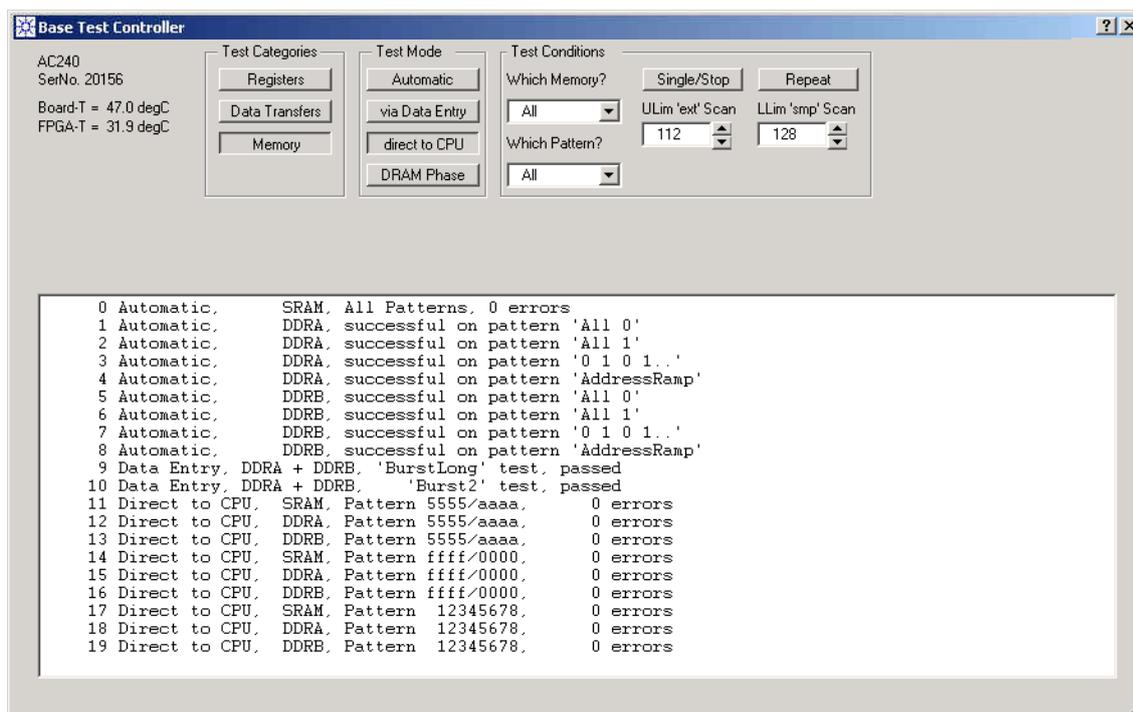
Each memory bank can be accessed through two ports. The *internal bus port* permits the user program to directly access the memory from the host computer and the *user port* gives access to the memory from (user-defined) cores within the FPGA. Each memory interface also offers a built-in memory tester for automatic verification.

The purpose of the memory tests is to exercise different memory accesses with various data patterns, through both access ports.

There are three test modes:

- **Automatic:** exercises the memories with the built-in test generator. It is the most ‘stressful’ test, exercising the memory at full speed with a ‘stressful’ data pattern.
- **Via Data Entry:** exercises the memory with read and write accesses through the user port.
- **Direct to CPU:** exercises the memory with read and write accesses through the internal bus port, i.e. with direct access from the host computer. This test is the slowest and is entirely implemented within AcqirisAnalyzer.

Each test can be run once or in repeat mode. The results of the tests are displayed in the log window of the Base Test Controller window.



5.6. FFT Spectrometer Application

The FFT Spectrometer firmware option implements a single channel real-time Fast Fourier Transform, capable of transforming 32K time-domain data points to 16K spectral lines at 2 GS/s or lower sampling rates, without any dead time. It computes the power spectrum and is able to accumulate power spectra in order to obtain a time-averaged power spectrum.

The operation of the FFT spectrometer is selected by choosing “2GS/s FFT” in the “Use for/as” field of the Instrument Control Window. It allows operation with sampling rates between 400 MS/s and 2 GS/s.

The control panel for this application consists of a control section (upper part of window) and a waveform display section in the lower part of the window.

5.6.1. Digitizer Setup

The control section should be operated left to right. First, you should use the **Acquisition** mode **Auto** (see figure below) to set the digitizer to the desired state, using the Digitizer Control window, as described in the section 5.3, ***DIGITIZER CONTROL WINDOW***. In particular, you should set the appropriate input full scale range and offset, and the sampling rate. Use **Ext Clk...** to set to an external clock source, if required. The display section shows the acquired waveform interactively.

Note: When the 2 GS/s FFT Spectrometer application is selected, the digitizer is automatically set to internal clock, 2 GS/s sampling rate, Chan Combination 2x, and input channel 1. You can change these values with the Digitizer Control Window, but you must keep Channel Combination at 2x.

5.6.2. Operation of the FFT Spectrometer

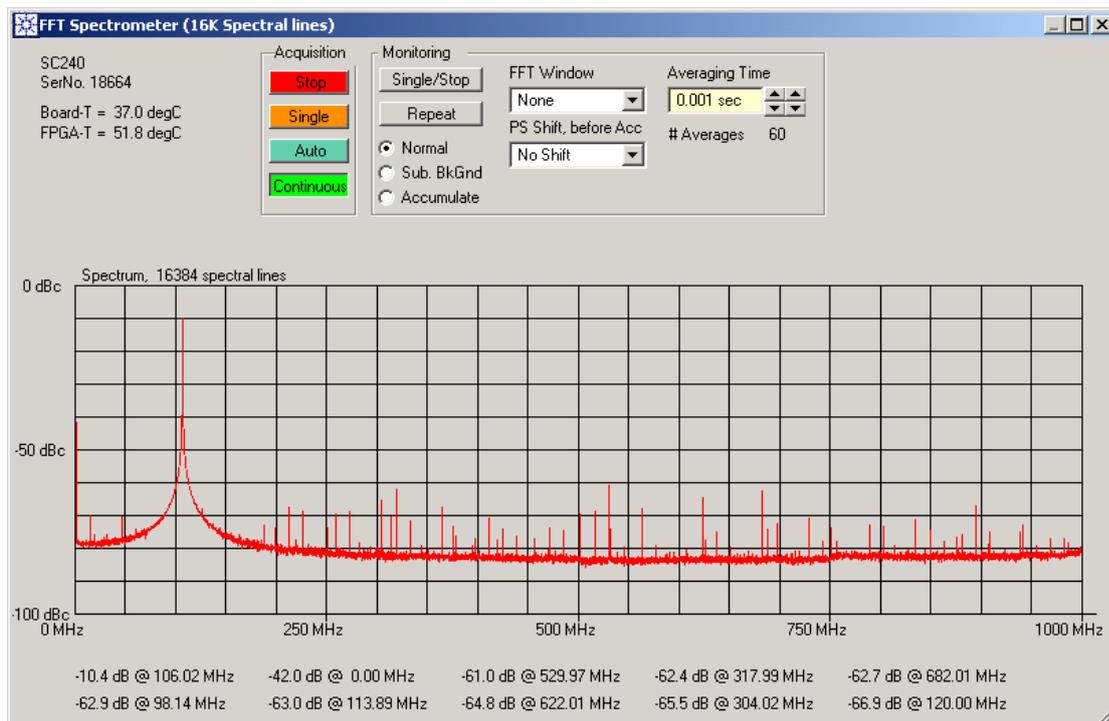
For the operation of the FFT spectrometer, set the **Acquisition** mode to **Continuous**. The status LED at the bottom of the module should turn green and stay that way. In this mode, changes to the Digitizer Control are still possible. They automatically provoke a stop of the FFT operation and the acquisition, loading of the new digitizer configuration, and a restart of the acquisition and the FFT operation.

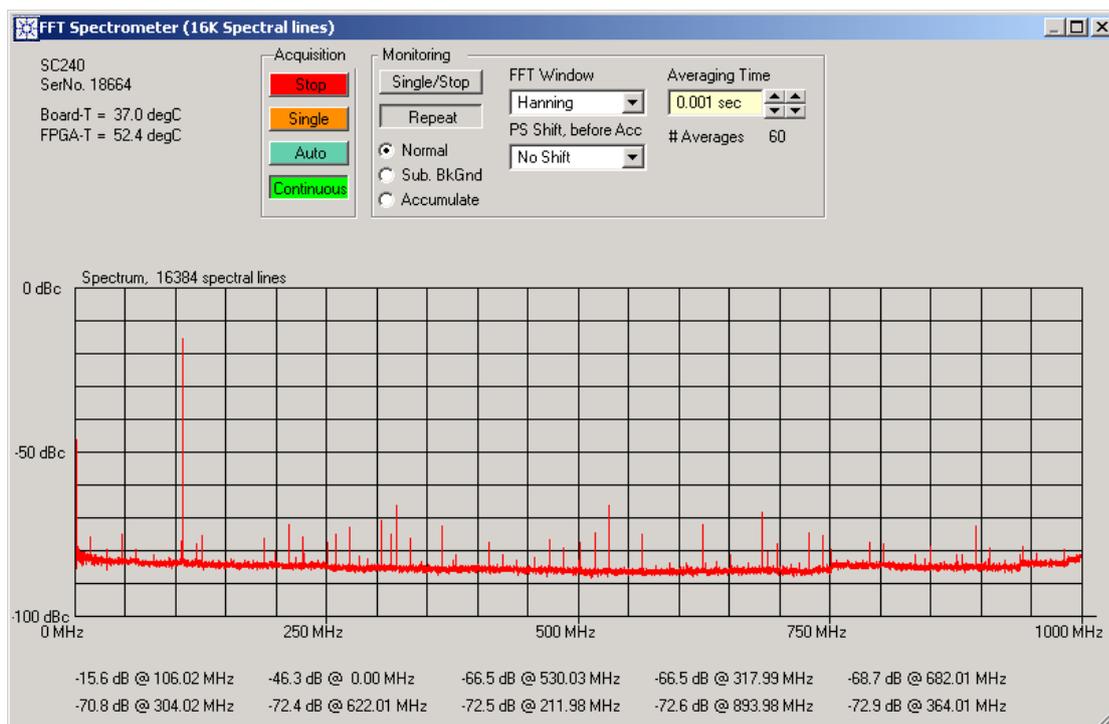
The FFT can be computed with different conditions, as defined by the fields **FFT Window**, **Averaging Time** and **PS Shift, before Acc**. These values may be changed at any time. If **Acquisition** is in **Continuous**, the FFT operation is interrupted, the new conditions are loaded, and the FFT computation is resumed. In the **Stop**, **Single**, and **Auto** modes, these FFT setup parameters are retained but not used.

The **FFT Window** allows the choice between **None** (rectangular window) and a number of simple windows, such as **Hamming**, **Hanning**, and **Blackman-Harris**.

The **Averaging Time** can be chosen in increments of 1 ms up to many seconds. The right-most control buttons increment/decrement the averaging time by 50 ms. The corresponding number of summed spectra is displayed below the field as *# Averages*. You can also set the requested time by selecting the display field, typing the desired numeric value, in ms, followed by a <CR>(Enter).

The value of **PS Shift, before Acc** permits shifting the power spectrum down by a number of bits, resulting in a larger range in the Power Spectrum Accumulator at the expense of a corresponding loss of accuracy. Choices are 0 (No Shift), 4, 8, and 12 bits. Shifting the power spectrum is of interest if the signal has a large peak in the spectrum, as may occur for large (almost) single-frequency signals, because the power spectrum overflows in a very short time. E.g. a full-scale sine wave will overflow the accumulator in less than 10 ms. If the signal has a wide frequency spread, as is the case in noise dominated signals, the accumulator will take longer before overflowing. Note that shifting the PS should be avoided whenever possible, in order to maintain the best accuracy.





5.6.3. Power Spectrum Display

The averaged power spectrum is displayed in the lower part of the application window.

The horizontal scale is from DC to the Nyquist frequency, i.e. $f_s/2$, where f_s is the sampling frequency. At 2 GS/s, the spectrum extends from DC to 1 GHz. There are always 16,384 spectral lines. Thus, the line width is $(f_s/2)/16,384$. At 2 GS/s, the line width is 61.035 kHz.

The vertical scale is logarithmic. The value 0 dBc corresponds to the largest possible power value, i.e. the power of a DC signal with a value of $+1/2$ or $-1/2$ of the input full-scale. A full-scale sine wave would give a peak at -3dBc, provided that a rectangular window is used and the frequency is an integer multiple of the line width.

The 10 largest peaks in the spectrum are identified and displayed below the power spectrum.

5.6.4. Monitoring

During the operation of the FFT spectrometer (**Acquisition** mode set to **Continuous**), the application reads the averaged power spectrum and displays it.

By default, the button **Repeat** should be depressed. In this mode, the power spectrum is read and displayed whenever it is ready. If the **Averaging Time** is short, the computer may be too slow to keep up with the data rate, so that only a fraction of the possible power spectra are read.

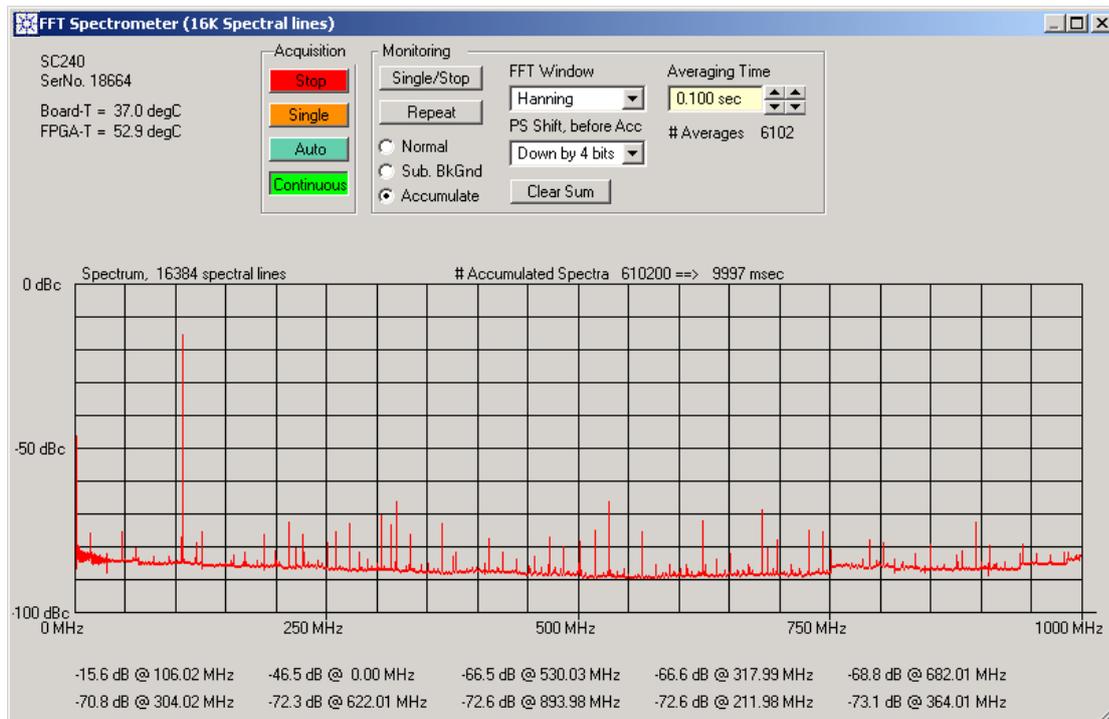
The button **Single/Stop** may be used to freeze the display or to obtain a single new power spectrum.

The **Normal**, **Sub. BkGnd**, and **Accumulate** radio buttons select the way in which the power spectrum is displayed.

- **Normal** selects the display of a single power spectrum, accumulated and averaged over the requested **Averaging Time**.

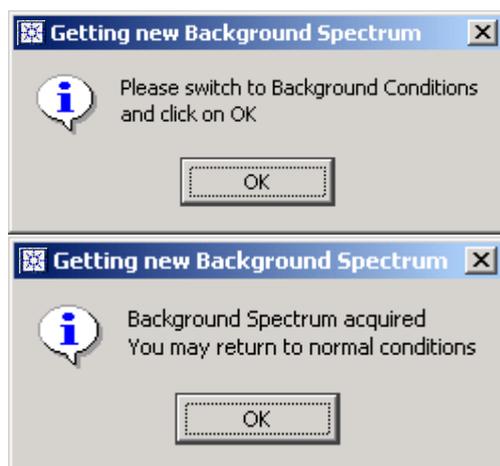
- **Accumulate** permits the addition/averaging of multiple power spectra by the computer. It effectively increases the averaging time, without running the risk of overflowing the accumulation buffer in the module. It also permits a livelier update rate. E.g. setting the **Averaging Time** to 1 sec and choosing **Accumulate** will allow averaging over many seconds and minutes, while updating the display every second. If the **Averaging Time** was set to 100 s instead, the display would not update until that time is elapsed.

The accumulation may be reset by pressing **Clear Sum** at the bottom of the **Monitoring** region, as shown in the figure below. The number of accumulated spectra and the corresponding data acquisition time are shown just above the power spectrum display.



- **Sub BkGnd** displays the power spectrum after the subtraction of a *Background Spectrum*. By default the background spectrum is set to zero.

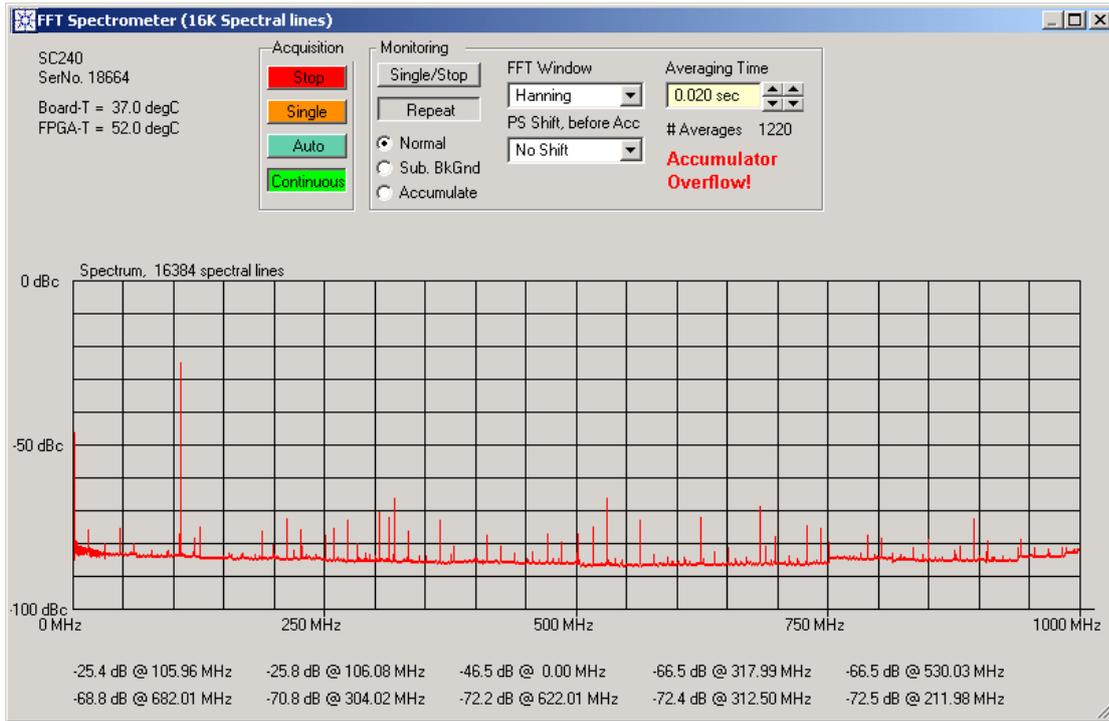
When selecting **Sub BkGnd**, an additional **New BkGnd** button is displayed, at the position of the **Clear Sum** button shown in the figure above. When this button is depressed, the FFT operation is interrupted and the small window shown below appears.



When you press OK after setting your system to a “Background” configuration, the application restarts the FFT operation and acquires a power spectrum in the new conditions. It will be used as the *Background Spectrum*.

Finally, the window shown below appears, informing you that you may switch back to normal operation.

Whenever the input signal exceeds the full range of the digitizer, its digital representation is clipped at the upper and/or lower range of the ADC and does not completely correspond to the signal. The power spectrum is therefore not entirely valid anymore. Similarly, if the accumulated power spectrum exceeds the storage capacity of the 36-bit accumulator buffer, the sum overflows and ‘wraps’ around to zero. The FFT Spectrometer firmware contains both an ADC Overflow and an Accumulator Overflow detector. In the case of overflow, the application shows this fact as a red warning text in the lower right hand corner of the **Monitoring** section as shown below.



6. Programming the Firmware

Please refer to the **Programmer's Guide** for explanations on which programming environments are supported and how to use them. The **Programmer's Reference Manual** lists all available functions and explains the use of their parameters. The function `Acqrs_logicDeviceIO`, essential for the control of the registers in the data processing unit (DPU), is described there.

The sample code lines below assume a C/C++ environment. They do not check the return value of the `AcqrsXX_...` functions. In real applications, you should always check the return values of functions.

The first part of this chapter describes programming aspects that are common to all applications. The second part contains sections that are specific to a particular firmware. They are marked as such.

6.1. Programming Aspects Common to All AC2x0 Applications

6.1.1. Accessing the DPU Registers

All operations in the data processing unit are controlled through registers that are implemented in the FPGA firmware. They are accessed through the function `Acqrs_logicDeviceIO`. One of its arguments is the *registerID* that identifies which register is written to or read from. The AC2x0 and SC2x0 analyzers accept values between 0 and 127, for a total of 128 user-accessible registers. Each application typically uses only a small subset of them. To find which ones, please refer to the appropriate section later in this chapter.

In order to make the code samples in the subsequent section more readable, the following 2 functions are defined:

```
long FPGARead(long regID, long nbrValues, long* dataArrayP)
{
    return Acqrs_logicDeviceIO(instrumentID, "Block1Dev1", regID,
                               nbrValues, dataArrayP, 0, 0);
}

long FPGAWrite(long regID, long nbrValues, long* dataArrayP)
{
    return Acqrs_logicDeviceIO(instrumentID, "Block1Dev1", regID, nbrValues,
                               dataArrayP, 1, 0);
}
```

6.1.2. Register Definitions

A number of variable and register definitions that may apply to Agilent Acqiris-supplied applications are used in this chapter:

```
// General Buffer Identifiers
enum FPGAioConstants
{
    DDR0BufAddress    = 0x00, // Address of dynamic RAM bank 0
    DDR1BufAddress    = 0x01, // Address of dynamic RAM bank 1
    SRAMBufAddress    = 0x04, // Address of static RAM
    DeFrBufAddress    = 0x08, // Address of data entry monitoring buffer
    InFrBufAddress    = 0x0c, // Address of input data monitoring buffer
    TxFrBufAddress    = 0x10, // Address of transmit monitoring buffer
    RxFrBufAddress    = 0x20, // Address of receive monitoring buffer
};

// Register addresses in Acqiris FPGA-firmware
enum FPGAregisters
{
    // Registers in region "Acqiris reserved"
    // These registers are common to all Acqiris-supplied FPGA designs
    ReadAddrReg       = 0, // Indirect Access Port
    StartAddrReg      = 1, // Start address within block
    BufferIDReg        = 2, // Buffer Identifier Register

    FPGActrlReg       = 3, // FPGA control register
    CodeProtectReg    = 4, // FPGA code protection register
    FPGAstatusReg     = 6, // FPGA status register
    TemperatureReg    = 7, // Temperature register
    DEctrlReg         = 8, // Data Entry (DE) control register
};
```

```

RangeGateCtrlReg    = 12, // Range Gate control register
RangeGateStatusLo  = 13, // Range Gate status register (lo part)
RangeGateStatusHi  = 14, // Range Gate status register (hi part)

FPIOlinkReg        = 32, // Control for 2 Front Panel PIO lines
DACctrlReg         = 33, // Control of 16-bit DAC
LEDReg             = 34, // LED control register

uDB_IOctrlReg      = 36, // Control register for microDB link
uDB_IOoutReg       = 37, // Output register for microDB link
uDB_IOinReg        = 38, // Input register for microDB link

// Registers in region "User"
// The registers defined below are used in some Acqiris FPGA designs,
// e.g. Base Test and Data Streamer firmware. In user-defined firmware
// designs, these registers may be assigned in an arbitrary way.
MainCtrlReg        = 64, // Main Control for ALL Acqiris applications
InBufCtrl          = 65, // Control for input monitoring buffer
TxBufCtrl          = 66, // Control for transmit monitoring buffer
RxBufCtrl          = 67, // Control for receive monitoring buffer
};

```

6.1.3. Device Initialization

Before any device can be used, each device must be initialized with a call to the function `Acqrs_InitWithOptions`, typically with *Identification by Order Found*. For details, please refer to section 3.2, *DEVICE INITIALIZATION* in the **Programmer's Guide**.

The function returns the **instrumentID**, (whose value will be different for each device), which must be subsequently used in any other function call to the device.

The sample code below assumes that there is a single instrument attached to the computer. The **Programmer's Guide** shows sample code for more complex situations.

```

ViSession instrumentID;
Acqrs_InitWithOptions("PCI::INSTR0", VI_FALSE, VI_FALSE, "", &instrumentID);

```

This initialization function will automatically recognize an AC240/210 and load the default FPGA configuration file *AC240.bit* or *AC210.bit*. This FPGA configuration supports some device tests, but not the FFT Spectrometer or any other specialized application. A special configuration file must be loaded, as shown in the next section.

6.1.4. Loading an FPGA Configuration File

The data streaming firmware is contained in configuration files with the extension **.bit**.

The appropriate file must be loaded explicitly with the following function call:

```

Acqrs_configLogicDevice(instrumentID, "Block1Dev1",
    "AC240FFT2GSs.bit", 1);

```

The string "Block1Dev1" identifies the FPGA on the board. For all AC2x0 and SC2x0, it is constant. The configuration files should be situated in the same directory as the application or the one indicated by the `fpgaPath` in the `AqDrv4.ini` file. The loading process typically takes several seconds.

6.1.5. Sequence of Data Processing Operations

Start: The data processing operations must be configured and started in a well-defined way:

- a) Configure the digitizer
- b) Start data conversion and start streaming data to DPU
- c) Configure the DPU for the requested operation
- d) Start the DPU operation
- e) Optionally: Interact with the DPU during operation, or monitor a continuous data transfer operation.

Stop: The data processing operations should be stopped in the opposite order:

- a) Stop the DPU operation

- b) Stop the data conversion and data streaming to the DPU

Each step is described in the subsequent sections in more detail.

6.1.5.1. Digitizer Configuration

The digitizer section must be appropriately configured for the expected input signal, as shown in the sample code below:

```
long fullScale = 1.0;           // full scale range = 1.0 V
long use = 1;                  // channel to be used (1 or 2)
double sampInterval = 1.0e-9; // 1 GS/s sampling rate
double delay = 0.0;           // delay ignored, set it to zero
AcqrsD1_configMode(instrumentID, 1, 0, 0);
AcqrsD1_configVertical(instrumentID, 1, fullScale, 0.0, 3, 0);
// the line below is needed for the AC240 ONLY!!
AcqrsD1_configChannelCombination( instrumentID, 2, use);
AcqrsD1_configHorizontal(instrumentID, sampInterval, delay);
```

Comments:

- The function **AcqrsD1_configMode** must be used to set the instrument to the mode *stream data to DPU* (mode = 1), whereby the instrument will **not** stop upon the receipt of a trigger, but continue data acquisition until explicitly stopped by a software command.
- The function **AcqrsD1_configVertical** configures the signal input channel (1). The full scale is set to 1.0 (V), the offset to 0.0 (V), the input coupling to 50 Ohms (3), and the bandwidth limit to *no limit* (0). These values should be appropriately modified for the actual input signal.
- The function **AcqrsD1_configChannelCombination** sets the AC240 for single channel operation.
- The function **AcqrsD1_configHorizontal** sets the sampling rate. The value *delay* is ignored, since it is only needed in the *normal* mode, where the trigger stops an acquisition.

For further details, please refer to the section 3.3 *DEVICE CONFIGURATION* in the **Programmer's Guide**.

6.1.5.2. Starting Data Conversion

The function **AcqrsD1_configMode** shown in the previous section, with mode = 1, together with the load of the appropriate firmware already has configured the AC240/210 to the mode where data is streamed into the DPU. The following code starts the operation of the data converters:

```
AcqrsD1_acquire(instrumentID);
```

Upon receipt of this command, the driver software translates the previously received configuration parameters into the appropriate register values and loads them into the AC240/210. Finally, it transmits a start command, whereupon the AC240/210 starts digitizing the signal at the input channel and transferring the converted data stream to the DPU. This will continue until an explicit software command stops it.

The data streaming firmware in the DPU will stay idle until further configuration commands are received, as shown in the next sections.

6.1.5.3. Configuring the Data Processing Unit

The DPU configuration is application specific. Typical configuration steps may be:

- Enabling the internal clocks of the FPGA (DCMs)


```
long fpgaCtrl = 0;
WriteFPGA(FPGA CtrlReg, 1, &fpgaCtrl); // First disable everything
fpgaCtrl |= 0x00ff0000; // Enable all DCMs
// fpgaCtrl |= 0x00000100; // Enable readout in Big-Endian format
WriteFPGA(FPGA CtrlReg, 1, &fpgaCtrl);
Sleep(10); // Wait some time
```
- Starting the Data Entry (DE) interface of the FPGA

```

long deCtrl = 0x00000000;
WriteFPGA(DECtrlReg, 1, &deCtrl);
deCtrl = 0x80000000;
WriteFPGA(DECtrlReg, 1, &deCtrl);

```

- Configuring the internal operation of the DPU; code for the FFT Spectrometer firmware can be seen in section 6.3.1.

The order of the function calls may be important.

6.1.5.4. Monitoring

If the firmware runs completely autonomously, e.g. if the result of the calculations is transmitted through the front-panel DAC, the host processor need not interact with the digitizer anymore, until the on-going processing is explicitly stopped. It may be useful to monitor the card by occasionally reading some monitoring data from it. This is entirely application-specific. Please refer to the appropriate section later in this chapter.

If the firmware does not run autonomously, it generates data that must periodically be read by the host computer. The data read operation is similar to a monitoring read operation. Again, it is application-specific. The FFT Spectrometer application does not have this functionality.

6.1.5.5. Stopping the DPU Operation

The operations required to stop the DPU is application specific. The FFT Spectrometer is stopped by clearing the following register:

```

long mainCtrl = 0x0; // Disable FFT computation
FPGAWrite(MainCtrlReg, 1, &mainCtrl);
Sleep(10); // Wait 10 ms, until FFT terminated

long deCtrl = 0x0; // Stop the DE interface in the FPGA
FPGAWrite(DECtrlReg, 1, &deCtrl);

long fpgaCtrl = 0x0; // Reset FPGA control
FPGAWrite(FPGA CtrlReg, 1, &fpgaCtrl);

```

6.1.5.6. Stopping the Data Conversion

The data acquisition is stopped with the function `AcqrsD1_stopAcquisition`.

```
AcqrsD1_stopAcquisition(instrumentID);
```

6.1.6. Reading the FPGA Temperature

The temperature of the FPGA is of interest, since failures might occur at high temperatures. If the firmware implements temperature monitoring (which is the case for all Agilent Acqiris-supplied firmware), then use the following code to read the temperature whenever required:

```

long tReg;
FPGARead(TemperatureReg, 1, &tReg);

if ((tReg & 0x8000) == 0)
{
    // Monitoring is NOT enabled, enable it first
    tReg |= 0x8000;
    FPGAWrite(TemperatureReg, 1, &tReg);
    FPGARead( TemperatureReg, 1, &tReg);
}
long tValue = (tReg & 0x1fff); // signed value
if(tValue > 0x1000)tValue = tValue - 0x2000;
double temperature = tValue * 0.0625;

```

The temperature value is in centigrade. Temperatures up to 85° C are acceptable. If the temperature exceeds this value, the cooling should be improved (or if possible, the dissipation of the operating firmware reduced by design changes). In extreme cases, contact Agilent Acqiris.

6.2. Registers in the Base Test and FFT firmware

Many of the registers listed below are only used in tests. For a description of the FFT Spectrometer registers, refer to section 6.3.8, **REGISTERS IN FFT SPECTROMETER FIRMWARE**. For FDK users needing a detailed description of the Base Test registers please consult the FDK Reference Manual.

6.2.1. Register List

Register Number	Register Address	Access Right	Comment			
			Base Test	Base Test with memory	FFT Spectrometer	
Customer Register Space – Reserved for Definition by Acqiris						
0	0x2200	RW	✓		✓	Indirect Access Port
1	0x2204	RW	✓		✓	FPGA Indirect Address
2	0x2208	RW	✓		✓	FPGA Buffer Identifier
3	0x220C	RW	✓		✓	FPGA Main Control
4	0x2210	R	✓		✓	FPGA Code Protection
5	0x2214					Reserved
6	0x2218	R	✓		✓	FPGA Main Status
7	0x221C					FPGA Temperature
8	0x2220	RW	✓		✓	FPGA DE-Bus Control
9	0x2224	RW	✓		✓	FPGA Direct Access Block
10-11	--					Reserved
12	0x2230	RW				Trigger Control
13	0x2234	RW				Trigger Control Status Lo
14	0x2238	RW				Trigger Control Status Hi
15	0x223C	RW				Trigger Control Delay
16-31	--					Reserved
32	0x2280	RW	✓		✓	Front Panel PIO Control
33	0x2284	RW	✓		✓	Front Panel DAC Control
34	0x2288	RW	✓		✓	Front Panel LED Control
35	0x228C					Reserved
36	0x2290	RW	✓		✓	Front Panel μ DB-IO Control
37	0x2294	RW	✓		✓	Front Panel μ DB-IO Output
38	0x2298	R	✓		✓	Front Panel μ DB-IO Input
39	0x229C	RW		✓		Dual Port Memory Control
40	0x22A0	RW		✓		Dual Port Memory Test Pattern
41	0x22A4	R		✓		Dual Port Memory Status
42	0x22A8	R		✓		Dual Port Memory Test Value
43	0x22AC	R		✓		Dual Port Memory Test Result
44	0x22B0	RW		✓		DDR A Control / Status
45	0x22B4	RW		✓		DDR A Self-test Control / Status
46	0x22B8	R		✓		DDR A Self-test Status 1
47	0x22BC	R		✓		DDR A Self-test Status 2
48	0x22C0	R		✓		DDR A Self-test Status 3
49	0x22C4	R		✓		DDR A Self-test Status 4
50	0x22C8	R		✓		DDR A Self-test Error Count
51	0x22CC					Reserved
52	0x22D0	RW		✓		DDR B Control / Status
53	0x22D4	RW		✓		DDR B Self-test Control / Status
54	0x22D8	R		✓		DDR B Self-test Status 1
55	0x22DC	R		✓		DDR B Self-test Status 2

56	0x22E0	R		✓		DDR B Self-test Status 3
57	0x22E4	R		✓		DDR B Self-test Status 4
58	0x22E8	R		✓		DDR B Self-test Error Count
59	0x22EC					Reserved
60	0x22F0	RW		✓		DCM Phase Shift Control / Status
61-63						Reserved
Customer Register Space for the Base Test firmware						
64	0x2300	RW	✓			Base Tests Control
65	0x2304	R	✓			Base Tests Status
66	0x2308	RW		✓		Example of SRAM Interfacing (1)
67	0x230C	RW		✓		Example of DRAM Interfacing (1)
66-127	--		✓			Free
Customer Register Space for the FFT Spectrometer firmware						
64	0x2300	RW			✓	FFT Main Control
65	0x2304	R			✓	FFT Status
66	0x2308	RW			✓	FFT Number of Accumulation
67	0x230C	RW			✓	FFT Configuration
68	0x2310	RW			✓	Front-Panel udB IO Control
69	0x2314	RW			✓	FFT Accumulator Clear
70	0x2318	R			✓	FFT Overflow Status
71	0x231C	R			✓	FFT ADC overflow counter
72 - 127						Unused

6.2.2. Indirect Addressing

In order to conserve address space, large data buffers within the FPGA are not directly mapped to User Registers, but are accessed with an indirect addressing method.

Indirect addressing access must be performed as follows:

1. Write the Start Address within the buffer of interest into the FPGA Indirect Address Register.
2. Write the Buffer Identifier value into the FPGA Buffer Identifier Register.
3. Perform as many read or write operations as required on the Indirect Access Port.

The Indirect Address mapping is listed below.

For FDK users needing a detailed description of the Indirect Addressing please consult the FDK Reference Manual.

6.2.2.1. Buffer Identifiers in Base Test Firmware

Buffer Identifier	Start Address Range	Access Right	Comment
Customer Register Space – Acqiris Reserved			
0x00	0x0 - 0x3FFFFFFC	RW	DDR BANK A
0x01	0x0 - 0x3FFFFFFC	RW	DDR BANK B
0x04	0x0 - 0xFFFFC	RW	Dual Port Memory 1MB
0x08	0x0 - 0x1FFC	RW	DE-Buffer, 8K samples per channel
0x0C	0x0 - 0x1FFC	RW	DE-Monitor, 8K samples per channel
0x00- 0x7F			Reserved for Acqiris
Customer Register Space – Customer Reserved			
0x80-0xFF			Reserved for Customer

6.2.2.2. Buffer Identifiers in FFT Spectrometer Firmware

Buffer Identifier	Start Address	Access Right	Comment
Customer Register Space – Acqiris Reserved			
0x08	0x0 - 0x1FFC	RW	DE-Buffer, 8K samples per channel
0x00- 0x7F			Reserved for Acqiris
Customer Register Space – Customer Reserved			
0x80			Unused
0x81	0x0	R	FFT Output Buffer
0x82	0x0	RW	Window Buffer
0x83 – 0xFF			Unused

6.3. Programming the FFT Spectrometer Application

The FFT Spectrometer application uses the registers 0 to 8 in the “Acqiris Reserved” region (see definitions in section 6.1.2 **REGISTER DEFINITIONS**) and a few registers in the “User” region (see below) for its operation:

```
// Registers in region "User"
// The registers defined below are used by the FFT Spectrometer Application
enum FPGAFFTregisters
{
    FFTMainCtrlReg      = 64, // Main Control for FFT-Analyzer
    StatusReg           = 65, // Status Register for FFT-Analyzer
    NbrAccReg           = 66, // Number of Accumulations
    FFTConfReg          = 67, // FFT Configuration register
    FPuDB_IOCtrlReg     = 68, // Control register for FP uDB IO-Connector
    FFTclearBufReg      = 69, // Accumulator Buffer clear (uses only bit 0)
    FFToverflowReg      = 70, // Overflow Status register

    SumOfPowerSpectrum = 0x81, // Buf Identifier, summed FFT power spectrum
    FFTWindowBuffer    = 0x82, // Buf Identifier, FFT Window
};
```

After having,

configured the FPGA, according to section **6.1.4, LOADING AN FPGA CONFIGURATION FILE**,
 configured the digitizer, according to section **6.1.5.1, DIGITIZER CONFIGURATION**,
 and starting the digitizer as shown in section **6.1.5.2, STARTING DATA CONVERSION**
 you can continue as follows in the next sections.

6.3.1. Configuring the Data Processing Unit

For the FFT Spectrometer, the DPU must be configured with the following commands:

```
// Load the FFT Window
LoadFFTWindow();           // See code in next section!

// Start the DE interface
long deCtrl = 0;
FPGAwrite(DECtrlReg, 1, &deCtrl);           // Clear first
deCtrl = 0x80000000;
FPGAwrite(DECtrlReg, 1, &deCtrl);           // Start DE

// Clear the FPGA control register
long fpgaCtrl = 0;
FPGAwrite(FPGA CtrlReg, 1, &fpgaCtrl);       // Clear FPGA Control

// If the 'End of Accumulation' is expected to arrive by interrupt
// rather than by polling, set the 'Interrupt enable' bit
// fpgaCtrl = 1;
// FPGAwrite(FPGA CtrlReg, 1, &fpgaCtrl); // Enable interrupts

// Initialize # of spectra to accumulate
// The register expects the number of 'pairs' of accumulations, since
// only an even number of accumulations is possible
long nbrDoubleAccumulations = nbrSpectra/2;
FPGAwrite(NbrAccReg, 1, &nbrDoubleAccumulations);

// Configure the FFT Computation and Read modes
long fftConf = 0;
// Set the read mode to 'high 32 bits, followed by low 32 bits'
// Instead of 0x2, use 0x1 for 'high 32 bits', or 0x0 for 'low 32 bits'
fftConf |= (0x2 << 18);
// If desired, set a shift of the power spectrum before accumulating
// Use 0x0 for 0 bits, 0x1 for 4 bits, 0x2 for 8 bits, 0x3 for 12 bits
fftConf |= (0x1 << 2);
FPGAwrite(FFTConfReg, 1, &fftConf);

Sleep(200);

// Enable the FPGA-core to operate
long mainCtrl = 1;
FPGAwrite(MainCtrlReg, 1, &mainCtrl);
```

Comments:

- The computation and loading of the (optional) FFT-window is shown in the next section
- The value *nbrSpectra* is user-defined
- The *Sleep(200)* function waits for 200 ms before starting the actual operation. This time is probably longer than needed, but no tests were made to find the lowest possible value.

6.3.2. Computing and Loading the (optional) FFT Window

This section describes the code needed to compute and load the FFT Window, shown in the previous section as the function call `LoadFFTWindow()`.

The default state of the firmware is for no window, equivalent to a rectangular window. If no other window is requested, simply ignore the code in this section.

The FFT window must be defined as an array of 32K values, for a 32K data input FFT. Due to the required symmetry of the window, it is sufficient to load the first 16K values of the window.

First, the window is computed as an array of floating point values in the range [-1.0, 1.0].

```
enum Window { None, Hanning, Hamming, Flattop, Blackmann };
Window wantedWindow;
const long nbrSpectralLines = 16384;
const double pi = 3.1415926535;
double window[nbrSpectralLines];

for (long i = 0; i < nbrSpectralLines; i++)
{
    double w = pi*i/nbrSpectralLines;
    double cos1 = cos(w);
    double cos2 = cos(2.0*w);
    switch(wantedWindow)
    {
        case None:      window[i] = 1.0;
                        break;
        case Hanning:   window[i] = 0.50 - 0.50*cos1;
                        break;
        case Hamming:   window[i] = 0.54 - 0.46*cos1;
                        break;
        case Flattop:   window[i] = 0.28106 - 0.5209*cos1 + 0.19804*cos2;
                        break;
        case Blackman:  window[i] = 0.42 - 0.50*cos1 + 0.08*cos2;
                        break;
    }
}
```

Comments:

- The windows types are defined as enumerated values, to which other values can be added
- The value `wantedWindow` is user-defined

The floating point representation must be translated into a 9-bit integer representation and coded by pairs into an array of 32-bit integers. The values with even indices must go into bits 9 – 17, those with odd indices into bits 0 – 8.

```
long windowArray[nbrSpectralLines/2];
for (long j = 0; j < nbrSpectralLines/2; j++)
{
    long winEven = 256*window[j*2];
    long winOdd  = 256*window[j*2+1];

    // Limit the values to 9 bits
    if (winEven < -256) winEven = -256;
    if (winEven > 255) winEven = 255;
    if (winOdd < -256) winOdd = -256;
    if (winOdd > 255) winOdd = 255;
    windowArray[j] = ((winEven&0x1fff) << 9) + winOdd&0x1fff;
}
```

Finally, the array is loaded into the FPGA with the following code:

```
long startAddr = 0x0;
long bufAddress = FFTWindowBuffer;
FPGAWrite(StartAddrReg, 1, &startAddr);
FPGAWrite(BufferIDReg, 1, &bufAddress);
FPGAWrite(ReadAddrReg, nbrSpectralLines/2, windowArray);
```

6.3.3. Reading Accumulated Spectra

The FFT Spectrometer firmware contains 2 accumulator buffers. After the first one is filled with the user-requested number of accumulated power spectra, it automatically starts accumulating into the second accumulator and sets a 'buffer full' bit in the status register. In addition, it may generate an interrupt to the host computer.

The host computer is expected to read the accumulator buffer fast enough to release it before the other buffer is full. Otherwise, the system cannot switch back to the first buffer, in which case data may be lost. The host computer should either continuously monitor the 'buffer full' bit, or request and wait for an interrupt.

A complete cycle consists of waiting for a full accumulation buffer, reading it, and releasing the vacated buffer, as shown in the following sample code. Both types of waiting for 'Buffer Full' are shown under (1a) and (1b).

(1a) Waiting for 'Buffer Full' (Polling)

```
bool full = false;
long mainStatus;
while (!full)
{
    FPGARead(StatusReg, 1, &mainStatus);
    if ((mainStatus & 0x80000000) != 0)
        full = true;
}
```

The code above does not have a timeout variable and is therefore not protected against situations where the buffer never becomes full (e.g. if the DPU initialization was not done correctly). This code will occupy the computer completely with waiting, which is not a very efficient use of the machine. The code under (1b) is potentially much more interesting, since it can liberate the CPU for other tasks, provided that 2 execution threads are used.

(1b) Waiting for 'Buffer Full' (Interrupt)

```
bool full;
long timeout = 100;
ViStatus = AcqrsD1_waitForEndOfProcessing(instrumentID, timeout);
if (status == ACQIRIS_ERROR_PROC_TIMEOUT)
    full = false;
else
    full = true;
```

The function **AcqrsD1_waitForEndOfProcessing** does not return before the buffer is full, unless the time *timeout* (in ms) has elapsed. In fact, this function will put the current execution thread to sleep until the interrupt from the AC240 arrives. If you want to do other operations while the power spectrum accumulation proceeds, consider putting the code above into a separate thread. Your main thread will then be free to execute whatever code you choose, e.g. computations on the already acquired spectra.

The interrupt only works if the interrupt was enabled, see section 6.3.1, **CONFIGURING THE DATA PROCESSING UNIT**. Otherwise, **AcqrsD1_waitForEndOfProcessing** will always return a timeout.

(2) Reading the Accumulation Buffer

```
const long nbrSpectralLines = 16384;
long spectrum[nbrSpectralLines];
long startAddr = 0x0;
long bufAddress = SumOfPowerSpectra;
FPGAWrite(StartAddrReg, 1, &startAddr);
FPGAWrite(BufferIDReg, 1, &bufAddress);
FPGARead(ReadAddrReg, nbrSpectralLines, spectrum);
```

The code above works if the read-modes 0x0 or 0x1 were chosen when configuring the DPU, see section 6.3.1, **CONFIGURING THE DATA PROCESSING UNIT**. If the mode 0x2, reading both upper and lower 32 bits, was

chosen, you must replace *nbrSpectralLines* with $2*nbrSpectralLines$ both in the dimension of the array *spectrum* and in the function *FPGARead*.

You don't have to specify which one of the 2 accumulation buffers to read, since the system will always use the 'oldest' (not yet released) buffer, i.e. the one that was filled first.

(3) Releasing the Accumulation Buffer

Reading the accumulator buffer with the *exact* number of accumulated spectral lines **automatically** releases the buffer for a new accumulation (after the other buffer has been filled).

You can prevent the automatic release of the buffer by setting the *bufClear* bit in the *FFTConfReg* (register 67). In this case, you must explicitly release the buffer by writing 1 into the *FFTclearBufReg* (register 69).

6.3.4. Normalizing the Accumulated Spectra

6.3.4.1. Normalizing the Raw Data

The different read modes and the possibility of shifting the power spectrum before accumulation, result in a fair number of different data ranges. It is attractive, but not required, to translate the raw data into a normalized floating-point format that will have the same interpretation, independent of read modes and data shifts.

The following code proposes to normalize to a data range that corresponds to that of the power spectrum of a single FFT, as it leaves the FFT-core, before data shifting and accumulation.

```
enum DShift { ShiftNone, ShiftBy4, ShiftBy8, ShiftBy12 };
enum ReadMode { ReadLo32, ReadHi32, Read36 };

double factor;
double nrmSpectrum[nbrSpectralLines]; // normalized spectrum

switch(dShift)
{
case ShiftNone: factor = 1.0/nbrSpectra; break;
case ShiftBy4: factor = 16.0/nbrSpectra; break;
case ShiftBy8: factor = 256.0/nbrSpectra; break;
case ShiftBy12: factor = 4096.0/nbrSpectra; break;
}

for (long i = 0; i < nbrSpectralLines; i++)
{
switch(readMode)
{
case ReadLo32:
nrmSpectrum[i] = (double)((unsigned long)spectrum[i])*factor;
break;

case ReadHi32:
nrmSpectrum[i] = (double)((unsigned long)
spectrum[i])*factor*16.0;
break;

case Read36:
{
// 'hi' corresponds to the upper 32 of the accumulated 36 bits
// 'lo' corresponds to the lower 32 of the accumulated 36 bits

double hi = 16.0*(double)((unsigned long)spectrum[2*i]);
double lo = (double)(spectrum[2*i + 1]&0xf); // use only 4 LSBs
nrmSpectrum[i] = (hi + lo)*factor;
}
}
}
}
```

The normalized spectrum is therefore an *averaged* spectrum. Of course, when further accumulating such averaged spectra with *different* N_{acc} , the sum would have to be weighted according to the different values of N_{acc} .

6.3.4.2. Converting the Spectra to dBc Full Scale

It may be useful, although not required, to convert the normalized power spectrum to a logarithmic representation.

The **Analyzer Demo** application displays the data in units of "dBc Full Scale", where 0 dBc corresponds to the largest possible power value, i.e. the power of a DC signal at of $+1/2$ or $-1/2$ input full-scale. A full-scale sine wave

would give a peak at -3dBc, provided that a rectangular window is used and the frequency is an integer multiple of the line width.

```
double nrmSpectrum[nbrSpectralLines];    // normalized spectrum, input
double logSpectrum[nbrSpectralLines];    // normalized spectrum, output

for (long i = 0; i < nbrSpectralLines; i++)
{
    if (nrmSpectrum[i] > 0.0)
        logSpectrum [i] = -96.28055 + 10.0*log10(nrmSpectrum[i] +1.0));
    else
        logSpectrum[i] = -100.0;
}
```

Note that the value 96.28055 comes from $10\log_{10}(4,246,732,800)$, the value 4,246,732,800 being the largest raw (DC) spectral value of a single power spectrum when the input is at $-\frac{1}{2}$ input full-scale and no window is used.

6.3.5. Checking on Overflows

It is recommended to check on ADC and/or Accumulator overflows whenever reading an accumulated spectrum. This is easily done by checking the bits 0 and 1 of the overflow register:

```
long ovflows;
FPGARead(FFToverflowReg, 1, &ovflows);
bool adcOvfl = (ovflows & 0x1) != 0;    // ADC
bool accOvfl = (ovflows & 0x2) != 0;    // Accumulator
```

There is no need to clear these bits, since they represent a state.

6.3.6. Stopping the FFT Spectrometer

The FFT Spectrometer is stopped by clearing the following register:

```
long mainCtrl = 0x0;    // Disable FFT computation
FPGAWrite(MainCtrlReg, 1, &mainCtrl);
Sleep(10);    // Wait 10 ms, until FFT terminated

long deCtrl = 0x0;    // Stop the DE interface in the FPGA
FPGAWrite(DECtrlReg, 1, &deCtrl);

long fpgaCtrl = 0x0;    // Reset FPGA control
FPGAWrite(FPGA CtrlReg, 1, &fpgaCtrl);
```

You may also want to stop the data acquisition (e.g. when changing channel input conditions), as described in the section 6.1.5.6, **STOPPING THE DATA CONVERSION**.

6.3.7. Synchronizing with other Equipment (e.g. Telescope)

In typical applications, the FFT Spectrometer is operated together with other equipment such as a telescope. Since background subtraction improves the performance considerably by eliminating some systematic effects, background spectra might be acquired quite regularly. This requires synchronization between the computation of the spectra and the movement of the telescope.

There are basically 2 ways of doing this:

- a) Make the FFT Spectrometer a slave to external electronics that determines the telescope movements
- b) Make the telescope movement a slave to the FFT Spectrometer, i.e. of the host computer of the spectrometer.

a1) Spectrometer as Slave: “Data Ready”

In this mode, the user would define an observation period of N seconds (e.g. 30). The observation might be ‘On Target’ for 1 period and ‘Off Target’ for another period, under control of external electronics. The FFT spectrometer would be working in ‘free-running’ mode, with an accumulation period of M seconds, for example 1 second.

One of the front panel μ DB-15 connector input pins can be defined as “Data Ready” with maybe another one as “Off Target”. They would be set/reset by the external electronics. The host computer would continuously monitor these lines, e.g. every millisecond. If “Data Ready” is low anywhere within the M second accumulation period, the accumulated sum would be considered bad and thrown away. If it is not low during the accumulation period, it would be retained and either used as ‘real’ data or background, depending on the value of “Off Target”.

Thus, any transition between “On” and “Off” would result in the loss of the approximately M seconds out of the nominal N seconds of observation time.

a2) Spectrometer as Slave: “Enable/Disable FFT”

This mode is very similar to a1) described above, except that it uses the front panel μ DB-15 connector input “Enable/disable FFT” (pin 1 of the connector) to actually veto the accumulation of power spectra. The transition to ‘enable’ clears the already accumulated spectrum since one typically wants to avoid mixing spectra before and after the transition.

There would still be some loss of accumulated spectra, but this method would have the advantage that the host computer would not have to continuously monitor the “Data Ready” line. It would be sufficient to read the value of the “Off Target” signal whenever an interrupt occurs.

b) Spectrometer as Master

In this mode, the 8 output lines of the front panel μ DB-15 connector can be used to initiate the movement of the telescope or other actions. The host computer software would typically wait until an accumulation is terminated before reading and resetting the FFT accumulation and setting the appropriate lines on the μ DB-15 connector to get the telescope to move. It probably would have to monitor one of the input lines to detect when the telescope is in a stable position.

6.3.8. Registers in FFT Spectrometer Firmware

This section describes in detail all registers that are specific to the FFT Spectrometer Firmware.

6.3.8.1. Indirect Access Port

This register gives access to large data blocks, together with the Indirect Address Register and the Buffer Identifier Register. As seen by the control software, it acts like a FIFO data port.

Register Space	Register Number	Register Address
Customer	0	0x2200
31..0		
IndirData		

[31..0] IndirData RW Indirect Data value. Every read or write access uses the indirect address defined by the Indirect Address and Buffer Identifier registers.

6.3.8.2. Indirect Address Register

Register Space	Register Number	Register Address
Customer	1	0x2204
31..0		
IndirAddr		

[31..0] IndirAddr RW This register defines the address for Indirect Access. It is used when accessing the Indirect Access Port. This address is defined in bytes and is auto incremented by 4 for each read or written word from / to the Indirect Access Port.

6.3.8.3. Buffer Identifier Register

Register Space	Register Number	Register Address
Customer	2	0x2208
31..0		
IndirCtr		

[31..0] IndirCtr RW This register defines the target for indirect Access. Available Buffers are listed in the section 6.2.2.

6.3.8.4. Acqiris Control Register

Register Space	Register Number	Register Address
Customer	3	0x220C

31..24	23..16	15..1	0
--	Enb_Dcm	--	IntEn

- [0] **IntEn** **RW** Must be set to '1' to enable the interrupt.
- [23..16] **Enb_Dcm** **RW** Enable the FPGA internal clocks. Shall be set to hexa 0xFF.

6.3.8.5. Code Protection Register

Register Space	Register Number	Register Address
Customer	4	0x2210

31..16	15..4	3..0
	DeveloperID	FirmwareID

- [3..0] **FirmwareID** **R** Fixed to 0x1
- [15..4] **DeveloperID** **R** Fixed to 0x10

6.3.8.6. Acqiris Status Register

Register Space	Register Number	Register Address
Customer	6	0x2218

31	30..28	27..25	24	23..16
				Lck_Dcm

15	14..8	7..3	2	1	0

- [23..16] **Lck_Dcm** **R** Status of the FPGA internal clocks. Two clocks shall be locked for correct operation. The clocks shall be first enabled by writing the bit 16 to 23 of the **Acqiris Control Register** to hexa 0xFF.
 - (0) '1' => **Lbclk** locked
 - (1) Not relevant
 - (2) Not relevant
 - (3) Not relevant
 - (4) '1' => **Declk** locked
 - (5) Not relevant
 - (6) Not relevant
 - (7) Not relevant

6.3.8.7. TempMonitor

Register Space	Register Number	Register Address
Customer	7	0x221C

31..16	15	14..13	12..0
--	TMPE	--	TMP_Monitor

[12..0] **TMP_Monitor** **R** FPGA temperature when monitoring is enabled

[15] **TMPE** **RW** FPGA Temperature Monitoring Enable

6.3.8.8. DEControl Register

Register Space	Register Number	Register Address
Customer	8	0x2220

31	30..0
DeStart	--

[30..0] **Reserved** **R** Must be set to 0.

[31] **DeStart** When set to '1', enables the data stream from the DE-Buffers to the FFT core.

6.3.8.9. Main Control Register

This register defines the main control of the FFT Spectrometer Application.

Register Space	Register Number	Register Address
Customer	64	0x2300

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
											EnT				EnF

[0] **EnF** **RW** FFT Enable

0 Disable operation

1 Enable operation

[4] **EnT** **RW** Trigger Enable

0 No trigger. Data are continuously read and computed on (after FFT operation is enabled)

1 Enable Trigger. Data are read and computed after the occurrence of a trigger

6.3.8.10. Status Register

This register contains the status of the FFT Spectrometer firmware.

Register Space	Register Number	Register Address
Customer	65	0x2304

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Full															Busy

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
				Tgd	Tm	DB	DA				Run		Enb	uDB	SftE

- [0] **SftE** **R** Software Enable Status
 - 0 FFT operation was not enabled by software, via Main Control Register
 - 1 FFT operation was enabled by software, via Main Control Register
- [1] **uDB** **R** uDB Enable status
 - 0 FFT operation was not enabled by front panel uDB connector
 - 1 FFT operation was enabled by front panel uDB connector
- [2] **Enb** **R** OR'd Software and uDB Enable Status
 - 0 FFT operation is not enabled
 - 1 FFT operation is enabled = OR of bits 0 and 1 of this register
- [4] **Run** **R** FFT Acquisition Status
 - 0 FFT-core is not acquiring data
 - 1 FFT-core is acquiring data = AND of bits 2, 8, and 9 of this register
- [8] **DA** **R** Input Channel A Status
 - 0 Input channel A off
 - 1 Input channel A delivering data
- [9] **DB** **R** Input Channel B Status
 - 0 Input channel B off
 - 1 Input channel B delivering data
- [10] **Tm** **R** Trigger mode Status
 - 0 Trigger mode off
 - 1 Trigger mode on
- [11] **Tgd** **R** Trigger received Status
 - 0 No trigger received
 - 1 Trigger received
- [16] **Busy** **R** FFT-core busy Status
 - 0 FFT-core is not busy
 - 1 FFT-core is busy computing
- [31] **Full** **R** Accumulator Buffer Full Status
 - 0 Accumulator buffer is not full
 - 1 Accumulator buffer is full

6.3.8.11. Accumulator Counter

This register controls the number of power spectra that are accumulated before generating the 'full' flag.

Register Space	Register Number	Register Address
Customer	66	0x2308

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
Nacc (31..16)															

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Nacc (15..0)															

- [31..0] Nacc RW** Number of (pairs of) power spectra to accumulate
 Due to the internal structure of the firmware, Nacc corresponds to ½ of the actually accumulated spectra.
- 0 Accumulate 2 spectra (special case)
 - 1 Accumulate 2 spectra
 - 2 Accumulate 4 spectra
 - etc...

6.3.8.12. FFT Configuration Register

This register controls the configuration of the FFT computation and readout.

Register Space	Register Number	Register Address
Customer	67	0x230C

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
												RdMode			

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
										OvW	Bclr	Ashift			

- [1..0] Test1 RW** Shall be set to 00
- [3..2] Ashift RW** Power Spectrum Down Shift before accumulation
- 0 No shift
 - 1 4 bits
 - 2 8 bits
 - 3 12 bits
- [4] Bclr RW** Reset mode of the accumulation
- 0 Accumulator buffer is cleared automatically on full read-out (requires that all 16'384 values are read)
 - 1 Accumulator buffer is cleared by writing 1 into bit 0 of register 69
- [5] OvW RW** Overwrite Buffer
- 0 Don't overwrite accumulation buffer until read or cleared
 - 1 Overwrite accumulation buffer when second buffer is full (not recommended)
- [8] Test2 RW** Shall be set to 0
- [17..16] Test3 RW** Shall be set to 00
- [19..18] RdMode RW** Read Mode
 When reading in modes 2 and 3, the data have an overlap of 28 bits, which must be eliminated when converting to a 64-bit integer or floating point number. See section 6.3.4, **NORMALIZING THE ACCUMULATED SPECTRA**.
- 0 Read lower 32 bits

- 1 Read upper 32 bits
- 2, 3 Read upper 32 bits, followed by lower 32 bits

6.3.8.13. Front Panel μ DB I/O Register

This register controls the I/O through the front panel μ DB connector.

Register Space	Register Number	Register Address
Customer	68	0x2310

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Full	uDBout									uDBIn					uDB

- [0]** **uDB** **RW** FFT operation enable input on μ DB pin 1.
 - 0 Disable FFT operation (if not also enabled by software via register 64)
 - 1 Enable FFT operation
- [5..1]** **uDBin** **RW** uDB input on pin 2 to 6.
 - Pin 2 corresponds to bit 1, pin 3 to bit 2 etc.
 - 0 Accumulator buffer is cleared automatically on full read-out (requires that all 16'384 values are read)
 - 1 Accumulator buffer is cleared by writing 1 into bit 0 of register 69
- [14..8]** **uDBout** **RW** uDB output on pin 7 to 13.
 - Pin 7 corresponds to bit 8, pin 8 to bit 9 etc.
- [15]** **Full** **RW** Accumulator Buffer Full output on μ DB pin14
 - 0 Accumulator buffer is not full
 - 1 Accumulator buffer is full

For a list of the I/O pins, see section 4.2.2 *I/O AND CONTROLS OF FFT SPECTROMETER FIRMWARE*.

6.3.8.14. Clear Buffer Register

This register controls the release of the accumulator buffer, when in non-automatic mode.

Register Space	Register Number	Register Address
Customer	69	0x2314

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
															ClrB

- [0]** **ClrB** **RW** Clear the accumulator buffer.
 - This bit is automatically reset to zero.
 - 1 Clear the buffer (only if it is full)

6.3.8.15. Overflow Status Register

This register indicates the number of ADC overflow ADC during the accumulation.

Register Space	Register Number	Register Address
Customer	70	0x2318

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
														Acc	ADC

- [0] **ADC** **R** ADC overflow
 - 1 ADC overflow detected, i.e. the (signed) codes 0x7f or 0x80 were detected
- [1] **Acc** **R** Accumulator overflow
 - 1 Overflow detected, i.e. at least one accumulator value exceeded the 36-bit (unsigned) range.

6.3.8.16. ADC Overflow Counter Register

This register indicates the overflow status of the ADC and the accumulator.

Register Space	Register Number	Register Address
Customer	71	0x231C

31..0															
ADCOvfCnt															

- [31..0] **ADCOvfCnt** **R** Indicates how many times an ADC-overflow has been detected during the accumulation of the programmed number of spectra (0x2308).

7. Appendix A: U1093A (U1056A) AS bus for CompactPCI/PXI Digitizers

Specification and User Instructions for AS1, AS2, AS3, AS4

SPECIFICATIONS :

ASBus Bridges allow the connection and synchronization of up to 7 DC series digitizers (up to 28 channels). Acqiris recommends having the master digitizer in the center of the group in order to obtain the best performance possible.

Clock synchronization accuracy : $\pm 100\text{ps}$.

Install/Remove the ASBus Bridges as follows :

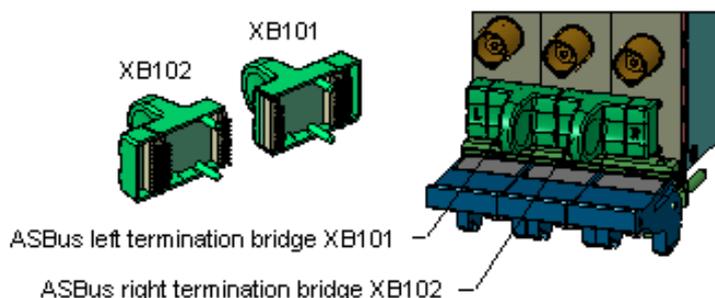
- Make sure that the screws securing each of the modules to the crate are not completely tightened.
- Before inserting the ASBus bridge, manually adjust any difference in the module separation, such that the distance between the center of the connectors corresponds to that of the ASBus bridge.
- Carefully insert/remove the ASBus bridges between each pair of modules. Force should not be required.
- Tighten the screws of each of the modules.
- If desired, add an XB104 Retainer on each handle to secure the ASBus bridge against vibration. It clicks into place when fully inserted. It can be easily removed by pulling on the lower half of the Retainer while sliding it off.

User Instructions for 2 digitizers :

ASBus dual termination bridge XB103



User Instructions for 3 digitizers :



User Instructions for 4 to 7 digitizers :

